## $\mu$ PD48288209-A <br> $\mu$ PD48288218-A <br> $\mu$ PD48288236-A

## 288M-BIT Low Latency DRAM Common I/O

## Description

The $\mu \mathrm{PD} 48288209-\mathrm{A}$ is a $33,554,432$-word by 9 bit, the $\mu \mathrm{PD} 48288218-\mathrm{A}$ is a $16,777,216$ word by 18 bit and the $\mu$ PD48288236-A is a $8,388,608$ word by 36 bit synchronous double data rate Low Latency RAM fabricated with advanced CMOS technology using one-transistor memory cell.

The $\mu \mathrm{PD} 48288209-\mathrm{A}, \mu \mathrm{PD} 48288218-\mathrm{A}$ and $\mu \mathrm{PD} 48288236-\mathrm{A}$ integrate unique synchronous peripheral circuitry and a burst counter. All input registers controlled by an input clock pair (CK and CK\#) are latched on the positive edge of CK and CK\#.
These products are suitable for application which require synchronous operation, high speed, low voltage, high density and wide bit configuration.

## Specification

- Density: 288M bit
- Organization
- Common I/O: 4 M words $\times 9$ bits $\times 8$ banks

2 M words $\times 18$ bits $\times 8$ banks
1 M words $\times 36$ bits $\times 8$ banks

- Operating frequency: $400 / 300 \mathrm{MHz}$
- Interface: HSTL I/O
- Package: 144-pin TAPE FBGA
- Package size: $18.5 \times 11$
- Leaded and Lead free
- Power supply
- $\quad 2.5 \mathrm{~V}$ Vext
- 1.8 V Vdo
- 1.5 V or $1.8 \mathrm{~V} \operatorname{VdD} \mathrm{Q}$
- Refresh command
- Auto Refresh
- 8192 cycle / 32 ms for each bank
- 64 K cycle / 32 ms for total
- Operating case temperature : $\mathrm{Tc}=0$ to $95^{\circ} \mathrm{C}$


## Features

- SRAM-type interface
- Double-data-rate architecture
- PLL circuitry
- Cycle time: $2.5 \mathrm{~ns} @$ trc = 20 ns
$3.3 \mathrm{~ns} @ \mathrm{trc}=20 \mathrm{~ns}$
- Non-multiplexed addresses
- Multiplexing option is available.
- Data mask for WRITE commands
- Differential input clocks (CK and CK\#)
- Differential input data clocks (DK and DK\#)
- Data valid signal (QVLD)
- Programmable burst length: 2 / 4 / 8 (x9 / x18)

2/4 (x36)

- User programmable impedance output ( $25 \Omega-60 \Omega$ )
- JTAG boundary scan

Ordering Information

| Part number | Cycle <br> Time <br> ns | Clock Frequency MHz | Random Cycle ns | Organization (word x bit) | Core Supply Voltage (Vext) V | Core Supply Voltage (Vod) V | Output Supply <br> Voltage <br> (VodQ) <br> V | Package |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mu \mathrm{PD} 48288209 \mathrm{FF}-\mathrm{E} 33-\mathrm{DW} 1-\mathrm{A}$ | 3.3 | 300 | 20 | 32 M x 9 | $\begin{aligned} & 2.5+0.13 \\ & 2.5-0.12 \end{aligned}$ | $1.8 \pm 0.1$ | $1.8 \pm 0.1$ | 144-pinTAPE FBGA$(18.5 \times 11)$Lead-free |
| $\mu$ PD48288218FF-E33-DW1-A | 3.3 | 300 | 20 | $16 \mathrm{M} \times 18$ |  |  |  |  |
| $\mu$ PD48288236FF-E33-DW1-A | 3.3 | 300 | 20 | $8 \mathrm{M} \times 36$ |  |  |  |  |
| $\mu$ PD48288209FF-EF25-DW1-A | 2.5 | 400 | 20 | 32 M x 9 |  |  | $1.5 \pm 0.1$ |  |
| $\mu$ PD48288209FF-EF33-DW1-A | 3.3 | 300 | 20 |  |  |  |  |  |
| $\mu$ PD48288218FF-EF25-DW1-A | 2.5 | 400 | 20 | $16 \mathrm{M} \times 18$ |  |  |  |  |
| $\mu$ PD48288218FF-EF33-DW1-A | 3.3 | 300 | 20 |  |  |  |  |  |
| $\mu$ PD48288236FF-EF25-DW1-A | 2.5 | 400 | 20 | $8 \mathrm{M} \times 36$ |  |  |  |  |
| $\mu$ PD48288236FF-EF33-DW1-A | 3.3 | 300 | 20 |  |  |  |  |  |

Remark Products with -A at the end of part number are lead-free products.

## Pin Configurations

\# indicates active LOW signal.
144-pin TAPE FBGA ( $18.5 \times 11$ )
(Top View) [Common I/O x36]

|  | 1 | 2 | 3 | 4 | 5 | 6 | 7 | 8 | 9 | 10 | 11 | 12 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A | Vref | Vss | Vext | Vss |  |  |  |  | Vss | Vext | TMS | TCK |
| B | Vod | DQ8 | DQ9 | VssQ |  |  |  |  | VssQ | DQ1 | DQ0 | Vod |
| C | Vtt | DQ10 | DQ11 | VddQ |  |  |  |  | VdoQ | DQ3 | DQ2 | $V_{\text {tt }}$ |
| D | $\begin{aligned} & \text { Note } \\ & \hline \text { (A22) } \\ & \hline \end{aligned}$ | DQ12 | DQ13 | VssQ |  |  |  |  | VssQ | QK0\# | QK0 | Vss |
| E | $\begin{aligned} & \text { Note } \\ & (\mathbf{A} 21) \\ & \hline \end{aligned}$ | DQ14 | DQ15 | VdoQ |  |  |  |  | VmdQ | DQ5 | DQ4 | $\begin{aligned} & \text { Note } \\ & (\mathbf{A} 20) \\ & \hline \end{aligned}$ |
| F | A5 | DQ16 | DQ17 | VssQ |  |  |  |  | VssQ | DQ7 | DQ6 | QVLD |
| G | A8 | A6 | A7 | Vdo |  |  |  |  | Vdo | A2 | A1 | A0 |
| H | BA2 | A9 | Vss | Vss |  |  |  |  | Vss | Vss | A4 | A3 |
| J | DK0 | DK0\# | Vdo | Vod |  |  |  |  | Vdo | Vdo | BA0 | CK |
| K | DK1 | DK1\# | VdD | Vdo |  |  |  |  | Vod | Vdo | BA1 | CK\# |
| L | REF\# | CS\# | Vss | Vss |  |  |  |  | Vss | Vss | A14 | A13 |
| M | WE\# | A16 | A17 | Vdo |  |  |  |  | VdD | A12 | A11 | A10 |
| N | A18 | DQ24 | DQ25 | VssQ |  |  |  |  | VssQ | DQ35 | DQ34 | $\begin{aligned} & \text { Note } \\ & \hline \text { (A19) } \\ & \hline \end{aligned}$ |
| P | A15 | DQ22 | DQ23 | VddQ |  |  |  |  | VdoQ | DQ33 | DQ32 | DM |
| R | Vss | QK1 | QK1\# | VssQ |  |  |  |  | VssQ | DQ31 | DQ30 | Vss |
| T | $V_{T T}$ | DQ20 | DQ21 | VddQ |  |  |  |  | VdiQ | DQ29 | DQ28 | $V_{\text {TT }}$ |
| U | Vod | DQ18 | DQ19 | VssQ |  |  |  |  | VssQ | DQ27 | DQ26 | Vdo |
| V | Vref | ZQ | Vext | Vss |  |  |  |  | Vss | Vext | TDO | TDI |

Note Reserved for future use. This signal is internally connected and has parasitic characteristics of an address input signal. This may optionally be connected to Vss, or left open.

| CK, CK\# | : Input clock | TMS | : IEEE 1149.1 Test input |
| :---: | :---: | :---: | :---: |
| CS\# | : Chip select | TDI | : IEEE 1149.1 Test input |
| WE\# | : WRITE command | TCK | : IEEE 1149.1 Clock input |
| REF\# | : Refresh command | TDO | IEEE 1149.1 Test output |
| A0-A18 | : Address inputs | Vref | HSTL input reference input |
| A19-A22 | : Reserved for the future | Vext | Power Supply |
| BA0-BA2 | : Bank address input | Vdo | Power Supply |
| DQ0-DQ35 | : Data input/output | VddQ | DQ Power Supply |
| DK0-DK1, DK0\#-DK1\# | : Input data clock | Vss | : Ground |
| DM | : Input data Mask | VssQ | DQ Ground |
| QK0-QK1, QK0\#-QK1\# | : Output data clock | $V_{T T}$ | Power Supply |
| QVLD | : Data Valid |  |  |
| ZQ | : Output impedance matching |  |  |

\# indicates active LOW signal.
144-pin TAPE FBGA ( $18.5 \times 11$ )
(Top View) [Common I/O x18]

|  | 1 | 2 | 3 | 4 | 5 | 6 | 7 | 8 | 9 | 10 | 11 | 12 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A | Vref | Vss | Vext | Vss |  |  |  |  | Vss | Vext | TMS | TCK |
| B | Vdd | $\begin{aligned} & \mathbf{N}^{\text {Note }}{ }^{3} \\ & \hline \end{aligned}$ | DQ4 | VssQ |  |  |  |  | VssQ | DQ0 | $\begin{array}{\|c\|c\|} \hline \text { DNote }^{3} \\ \hline \end{array}$ | Vdd |
| C | Vtт | $\begin{aligned} & \text { Note }^{3} \\ & \text { DNU } \end{aligned}$ | DQ5 | VdoQ |  |  |  |  | VdoQ | DQ1 | Note 3 <br> DNU | $V_{\text {tt }}$ |
| D | $\begin{gathered} { }^{\text {Note }}{ }^{1} \\ (\mathbf{A 2 2 )} \\ \hline \end{gathered}$ | $\begin{gathered} \text { Note }{ }^{3} \\ \text { DNU }^{3} \end{gathered}$ | DQ6 | VssQ |  |  |  |  | VssQ | QK0\# | QK0 | Vss |
| E | $\begin{array}{r} \text { Note } 1 \\ (\text { A21) } \end{array}$ | $\begin{aligned} & \text { Notet }^{3} \\ & \text { DNU }^{3} \end{aligned}$ | DQ7 | VdoQ |  |  |  |  | VdoQ | DQ2 | $\begin{gathered} \text { Note } \\ \hline \end{gathered}$ DNU | $\begin{gathered} \begin{array}{c} \text { Note } \\ \hline \\ \text { (A20) } \end{array} \\ \hline \end{gathered}$ |
| F | A5 | $\begin{aligned} & \text { NNOte }^{3} \\ & \text { DNU }^{3} \end{aligned}$ | DQ8 | VssQ |  |  |  |  | VssQ | DQ3 | $\begin{gathered} \text { Notete }^{3} \\ \text { DNU }^{3} \end{gathered}$ | QVLD |
| G | A8 | A6 | A7 | Vdo |  |  |  |  | Vdo | A2 | A1 | A0 |
| H | BA2 | A9 | Vss | Vss |  |  |  |  | Vss | Vss | A4 | A3 |
| J | $\begin{aligned} & \hline \text { Note } 2 \\ & \text { NF } \\ & \hline \end{aligned}$ | $\text { Note } 2$ NF | Vdd | Vod |  |  |  |  | Vod | Vod | BAO | CK |
| K | DK | DK\# | VdD | Vdd |  |  |  |  | Vdo | Vdo | BA1 | CK\# |
| L | REF\# | CS\# | Vss | Vss |  |  |  |  | Vss | Vss | A14 | A13 |
| M | WE\# | A16 | A17 | Vdd |  |  |  |  | VdD | A12 | A11 | A10 |
| N | A18 | $\begin{gathered} \text { DNU }^{\text {Note }} \\ \hline \end{gathered}$ | DQ14 | VssQ |  |  |  |  | VssQ | DQ9 | $\begin{gathered} \text { DNU }^{\text {Note }} \\ \hline \end{gathered}$ | A19 |
| P | A15 | $\begin{aligned} & \text { Note }{ }^{\text {N }}{ }^{2} \\ & \hline \end{aligned}$ | DQ15 | VdoQ |  |  |  |  | VdoQ | DQ10 | $\begin{gathered} \text { Note } \\ \text { DNU }^{3} \\ \hline \end{gathered}$ | DM |
| R | Vss | QK1 | QK1\# | VssQ |  |  |  |  | VssQ | DQ11 | $\begin{aligned} & \text { Note }{ }^{\text {No }} \end{aligned}$ | Vss |
| T | Vtt | $\begin{aligned} & \text { Note }^{3} \\ & \text { DNU } \end{aligned}$ | DQ16 | VdoQ |  |  |  |  | VdoQ | DQ12 | $\begin{gathered} \text { Note }^{3} \\ \text { DNU } \end{gathered}$ | $V_{\text {tt }}$ |
| U | Vdd | $\begin{aligned} & \text { DNU }^{\text {Note }} \\ & \hline \end{aligned}$ | DQ17 | VssQ |  |  |  |  | VssQ | DQ13 | $\begin{gathered} \text { Notete }^{3} \\ \text { DNU }^{3} \end{gathered}$ | Vod |
| v | Vref | ZQ | Vext | Vss |  |  |  |  | Vss | Vext | TDO | TDI |

Notes 1. Reserved for future use. This signal is internally connected and has parasitic characteristics of an address input signal. This may optionally be connected to Vss, or left open.
2. No function. This signal is internally connected and has parasitic characteristics of a clock input signal. This may optionally be connected to Vss, or left open.
3. Do not use. This signal is internally connected and has parasitic characteristics of a I/O. This may optionally be connected to Vss, or left open.

| CK, CK\# | : Input clock | TMS | : IEEE 1149.1 Test input |
| :--- | :--- | :--- | :--- |
| CS\# | : Chip select | TDI | : IEEE 1149.1 Test input |
| WE\# | : WRITE command | TCK | : IEEE 1149.1 Clock input |
| REF\# | : Refresh command | TDO | : IEEE 1149.1 Test output |
| A0-A19 | : Address inputs | VREF | : HSTL input reference input |
| A20-A22 | : Reserved for the future | VEXT | : Power Supply |
| BA0-BA2 | : Bank address input | VDD | : Power Supply |
| DQ0-DQ17 | : Data input/output | VDDQ | : DQ Power Supply |
| DK, DK\# | : Input data clock | Vss | : Ground |
| DM | : Input data Mask | VssQ | : DQ Ground |
| QK0-QK1, QK0\#-QK1\# | : Output data clock | VTT | : Power Supply |
| QVLD | : Data Valid | NF | : No function |
| ZQ | : Output impedance matching | DNU | : Do not use |

\# indicates active LOW signal.
144-pin TAPE FBGA ( $18.5 \times 11$ )
(Top View) [Common I/O x9]

|  | 1 | 2 | 3 | 4 | 5 | 6 | 7 | 8 | 9 | 10 | 11 | 12 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A | Vref | Vss | Vext | Vss |  |  |  |  | Vss | Vext | TMS | TCK |
| B | Vod | Note 3 <br> DNU | Note 3 DNU | VssQ |  |  |  |  | VssQ | DQ0 | Note 3 DNU | Vod |
| C | Vtt | $\begin{aligned} & \text { Note } \\ & \text { DNU } \end{aligned}$ | $\begin{aligned} & \text { Note }{ }^{\text {DNO }} \end{aligned}$ | VdoQ |  |  |  |  | VdoQ | DQ1 | $\begin{aligned} & \text { Note }{ }^{\text {NNU }} \end{aligned}$ | Vtt |
| D | $\begin{gathered} \begin{array}{c} \text { Note } \\ \hline \\ \text { (A22) } \end{array} \\ \hline \end{gathered}$ | $\begin{gathered} \text { Note }{ }^{3} \end{gathered}$ DNU | $\text { Note } 3$ DNU | VssQ |  |  |  |  | VssQ | QK0\# | QK0 | Vss |
| E | $\begin{gathered} \text { Note } \\ (\mathbf{A} 21) \end{gathered}$ | $\begin{aligned} & \text { Note }{ }^{3} \end{aligned}$ | $\mathrm{DNU}^{\text {Note }}$ | VdoQ |  |  |  |  | VdoQ | DQ2 | $\mathrm{NNO}^{\text {Note }}$ | A20 |
| F | A5 | ${ }^{\text {Note }}{ }^{3}$ DNU | $\begin{gathered} \text { Note } \\ \text { DNU }^{\text {NN }} \end{gathered}$ | VssQ |  |  |  |  | VssQ | DQ3 | $\begin{gathered} \text { Note }^{3} \\ \mathrm{DNU}^{2} \end{gathered}$ | QVLD |
| G | A8 | A6 | A7 | Vod |  |  |  |  | Vod | A2 | A1 | A0 |
| H | BA2 | A9 | Vss | Vss |  |  |  |  | Vss | Vss | A4 | A3 |
| J | $\begin{aligned} & \hline \text { Note }{ }^{2} \\ & \text { NF } \\ & \hline \end{aligned}$ | $\begin{aligned} & \text { Note } 2 \end{aligned}$ $\mathrm{NF}$ | Vod | Vdo |  |  |  |  | Vod | Vdo | BA0 | CK |
| K | DK | DK\# | VdD | Vdo |  |  |  |  | Vod | VdD | BA1 | CK\# |
| L | REF\# | CS\# | Vss | Vss |  |  |  |  | Vss | Vss | A14 | A13 |
| M | WE\# | A16 | A17 | Vdo |  |  |  |  | Vdo | A12 | A11 | A10 |
| N | A18 | Note 3 <br> DNU | Note 3 DNU | VssQ |  |  |  |  | VssQ | DQ4 | Note 3 DNU | A19 |
| P | A15 | Note 3 <br> DNU | $\text { Note } 3$ DNU | VdoQ |  |  |  |  | VdoQ | DQ5 | Note 3 DNU | DM |
| R | Vss | $\begin{gathered} \text { Note }^{3} \\ \text { DNU }^{2} \end{gathered}$ | $\begin{aligned} & \text { Note }^{2} \end{aligned}$ | VssQ |  |  |  |  | VssQ | DQ6 | Note 3 <br> DNU | Vss |
| T | Vtt | $\text { Note } 3$ DNU | $\text { Note } 3$ DNU | VdoQ |  |  |  |  | VdoQ | DQ7 | Note 3 <br> DNU | $V_{\text {tt }}$ |
| U | VdD | $\begin{gathered} \text { Note } \\ \text { DNU } \end{gathered}$ | $\begin{aligned} & \text { Note } \\ & \text { DNU } \end{aligned}$ | VssQ |  |  |  |  | VssQ | DQ8 | $\begin{aligned} & \text { Note } 3^{\text {DNU }} \end{aligned}$ | Vod |
| V | Vref | ZQ | Vext | Vss |  |  |  |  | Vss | Vext | TDO | TDI |

Notes 1. Reserved for future use. This signal is internally connected and has parasitic characteristics of an address input signal. This may optionally be connected to Vss, or left open.
2. No function. This signal is internally connected and has parasitic characteristics of a clock input signal. This may optionally be connected to Vss, or left open.
3. Do not use. This signal is internally connected and has parasitic characteristics of a I/O. This may optionally be connected to Vss, or left open.

| CK, CK\# | : Input clock | TMS | : IEEE 1149.1 Test input |
| :--- | :--- | :--- | :--- |
| CS\# | : Chip select | TDI | : IEEE 1149.1 Test input |
| WE\# | : WRITE command | TCK | : IEEE 1149.1 Clock input |
| REF\# | : Refresh command | TDO | : IEEE 1149.1 Test output |
| A0-A20 | : Address inputs | VREF | : HSTL input reference input |
| A21-A22 | : Reserved for the future | VEXT | : Power Supply |
| BA0-BA2 | : Bank address input | VDD | : Power Supply |
| DQ0-DQ8 | : Data input/output | VDDQ | : DQ Power Supply |
| DK, DK\# | : Input data clock | Vss | : Ground |
| DM | : Input data Mask | VssQ | : DQ Ground |
| QKO, QK0\# | : Output data clock | VTT | : Power Supply |
| QVLD | : Data Valid | NF | : No function |
| ZQ | : Output impedance matching | DNU | : Do not use |

Pin Identification

| Symbol | Type | Description |
| :---: | :---: | :---: |
| CK, CK\# | Input | Clock inputs: <br> CK and CK\# are differential clock inputs. This input clock pair registers address and control inputs on the rising edge of CK. CK\# is ideally 180 degrees out of phase with CK. |
| CS\# | Input | Chip select <br> CS\# enables the commands when CS\# is LOW and disables them when CS\# is HIGH. When the command is disabled, new commands are ignored, but internal operations continue. |
| WE\#, REF\# | Input | WRITE command pin, Refresh command pin: <br> WE\#, REF\# are sampled at the positive edge of CK, WE\#, and REF\# define (together with CS\#) the command to be executed. |
| A0-A20 | Input | Address inputs: <br> A0-A20 define the row and column addresses for READ and WRITE operations. During a MODE REGISTER SET, the address inputs define the register settings. They are sampled at the rising edge of CK. <br> In the x36 configuration, A19-A20 are reserved for address expansion; in the $x 18$ configuration, A20 is reserved for address expansion. These expansion addresses can be treated as address inputs, but they do not affect the operation of the device. |
| A21-A22 | Input | Reserved for future use: <br> These signals should be tied to Vss or leave open. |
| BA0-BA2 | Input | Bank address inputs; <br> Select to which internal bank a command is being applied. |
| DQ0-DQ35 | Input <br> /Output | Data input/output: <br> The DQ signals form the 36 bit data bus. During READ commands, the data is referenced to both edges of QKx. During WRITE commands, the data is sampled at both edges of DKx. |
| QKx, QKx\# | Output | Output data clocks: <br> QKx and QKx\# are opposite polarity, output data clocks. They are always free running and edgealigned with data output from the $\mu$ PD48288209/18/36-A. QKx\# is ideally 180 degrees out of phase with QKx. <br> For the $x 36$ device, QK0 and QK0\# are aligned with DQ0-DQ17. QK1 and QK1\# are aligned with DQ18-DQ35. For the $\times 18$ device, QK0 and QK0\# are aligned with DQ0-DQ8. QK1 and QK1\# are aligned with DQ9-DQ17. For the x 9 device, QK0 and QK0\# are aligned with DQ0-DQ8. |
| DKx, DKx\# | Input | Input data clock; <br> DKx and DKx\# are the differential input data clocks. All input data is referenced to both edges of DK. DK\# is ideally 180 degrees out of phase with DK. <br> For the $x 36$ device, DQ0-DQ17 are referenced to DK0 and DK0\#, and DQ18-DQ35 are referenced to DK1 and DK1\#. For the x 9 and x 18 devices, all DQs are referenced to DK and DK\#. |
| DM | Input | Input data mask; <br> The DM signal is the input mask signal for WRITE data. Input data is masked when DM is sampled HIGH along with the WRITE input data. DM is sampled on both edges of DK (DK1 for the x36 configuration). The signal should be Vss if not used. |
| QVLD | Output | Data valid; <br> The QVLD indicates valid output data. QVLD is edge-aligned with QKx and QKx\#. |


| Symbol | Type | Description |
| :---: | :---: | :---: |
| ZQ | Input <br> /Output | External impedance [25 $\Omega-60 \Omega$ ]; <br> This signal is used to tune the device outputs to the system data bus impedance. DQ output impedance is set to $0.2 \times R Q$, where $R Q$ is a resistor from this signal to Vss. Connecting ZQ to Vss invokes the minimum impedance mode. Connecting $Z Q$ to $V_{D D Q}$ invokes the maximum impedance mode. Refer to Figure 2-5. Mode Register Bit Map to activate this function. |
| TMS , TDI | Input | JTAG function pins: <br> IEEE 1149.1 test inputs: These balls may be left as no connects if the JTAG function is not used in the circuit |
| TCK | Input | JTAG function pin; <br> IEEE 1149.1 clock input: This ball must be tied to Vss if the JTAG function is not used in the circuit. |
| TDO | Output | JTAG function pin; <br> IEEE 1149.1 test output: JTAG output. <br> This ball may be left as no connect if JTAG function is not used. |
| Vref | Input | Input reference voltage; <br> Nominally $V_{D D} Q / 2$. Provides a reference voltage for the input buffers. |
| VExt | Supply | Power supply; <br> 2.5 V nominal. See Recommended DC Operating Conditions for range. |
| VDD | Supply | Power supply; <br> 1.8 V nominal. See Recommended DC Operating Conditions for range. |
| VodQ | Supply | DQ power supply; <br> Nominally, 1.5 V or 1.8 V . Isolated on the device for improved noise immunity. <br> See Recommended DC Operating Conditions for range. |
| Vss | Supply | Ground |
| VssQ | Supply | DQ ground; <br> Isolated on the device for improved noise immunity. |
| $V_{\text {TT }}$ | Supply | Power supply; <br> Isolated termination supply. Nominally, VodQ/2. See Recommended DC Operating Conditions for range. |
| NF |  | No function; <br> These balls may be connected to Vss. |
| DNU |  | Do not use; <br> These balls may be connected to Vss. |

## Block Diagram

8M x 36


Note When the BL=4 setting is used, A18 is "Don't care".

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## 1. Electrical Specifications

Absolute Maximum Ratings

| Parameter | Symbol | Conditions | Rating | Unit | Note |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Supply voltage | Vext |  | -0.3 to +2.8 | V |  |
| Supply voltage | VDD |  | -0.3 to +2.1 | V |  |
| Output supply voltage, <br> Input voltage, Input / Output voltage | VdoQ | 1.8 V nominal | -0.3 to +2.1 | V | 1 |
|  |  | 1.5 V nominal | -0.3 to +1.975 | V | 1 |
| Input / Output voltage | VIH / VIL | 1.8 V nominal | -0.3 to +2.1 | V | 1 |
|  |  | 1.5 V nominal | -0.3 to +1.975 | V | 1 |
| Junction temperature | Tj MAX. |  | 110 | ${ }^{\circ} \mathrm{C}$ |  |
| Storage temperature | $\mathrm{T}_{\text {stg }}$ |  | -55 to +125 | ${ }^{\circ} \mathrm{C}$ |  |

Note 1. The $\mu$ PD48288209/18/36FF-E support 1.8 V VDDQ nominal.
The $\mu$ PD48288209/18/36FF-EF support 1.5 V VDDQ nominal.

Caution Exposing the device to stress above those listed in Absolute Maximum Ratings could cause permanent damage. The device is not meant to be operated under conditions outside the limits described in the operational section of this specification. Exposure to Absolute Maximum Rating conditions for extended periods may affect device reliability.

## Recommended DC Operating Conditions

$0^{\circ} \mathrm{C} \leq \mathrm{Tc} \leq 95^{\circ} \mathrm{C} ; 1.7 \mathrm{~V} \leq \mathrm{VDD} \leq 1.9 \mathrm{~V}$, unless otherwise noted

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit | Note |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Supply voltage | Vext |  | 2.38 | 2.5 | 2.63 | V | 1 |
| Supply voltage | VDD |  | 1.7 | 1.8 | 1.9 | V | 1 |
| Output supply voltage | VodQ |  | 1.7 | 1.8 | 1.9 | V | 1, 2, 3 |
|  |  |  | 1.4 | 1.5 | 1.6 | V | 1, 3 |
| Reference Voltage | $V_{\text {Ref }}$ |  | $0.49 \times \mathrm{VdoQ}$ | $0.5 \times \mathrm{VdoQ}$ | $0.51 \times \mathrm{VdDQ}$ | V | 1, 4, 5 |
| Termination voltage | $V_{\text {TT }}$ |  | $0.95 \times V_{\text {REF }}$ | $V_{\text {Ref }}$ | $1.05 \times \mathrm{V}_{\text {ref }}$ | V | 1,6 |
| Input HIGH voltage | $\mathrm{V}_{\mathrm{IH} \text { ( } \mathrm{DC})}$ |  | $\mathrm{V}_{\text {Ref }}+0.1$ |  |  | V | 1 |
| Input LOW voltage | VIL (DC) |  |  |  | Vref - 0.1 | V | 1 |

Notes 1. All voltage referenced to Vss (GND).
2. During normal operation, VDDQ must not exceed VDD.
3. The $\mu$ PD $48288209 / 18 / 36 F F-E$ support 1.8 V VdoQ nominal The $\mu$ PD48288209/18/36FF-EF support 1.5 V VddQ nominal.
4. Typically the value of $V_{\text {ref }}$ is expect to be $0.5 \times V_{d D Q}$ of the transmitting device. Vref is expected to track variations in VDDQ.
5. Peak-to-peak AC noise on Vref must not exceed $\pm 2 \% V_{\text {ref }}(\mathrm{DC})$.
6. $V_{T T}$ is expected to be set equal to $V_{\text {REF }}$ and must track variations in the $D C$ level of $V_{\text {REF }}$.

## DC Characteristics

$0^{\circ} \mathrm{C} \leq \mathrm{Tc} \leq 95^{\circ} \mathrm{C}$; $1.7 \mathrm{~V} \leq \mathrm{VDD} \leq 1.9 \mathrm{~V}$, unless otherwise noted

| Parameter | Symbol | Test condition | MIN. | MAX. | Unit | Note |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input leakage current | l L |  | -5 | +5 | $\mu \mathrm{A}$ | 1,2 |
| Output leakage current | ILo |  | -5 | +5 | $\mu \mathrm{A}$ | 1,2 |
| Reference voltage current | Iref |  | -5 | +5 | $\mu \mathrm{A}$ | 1,2 |
| Output high current | Іон | $\mathrm{VOH}_{\text {O }}=\mathrm{V}_{\text {doQ }} / 2$ | (VdoQ/2) / (1.15 $\times$ RQ/5) | (VdoQ/2) / (0.85 x RQ/5) | mA | 3,4 |
| Output low current | loL | $\mathrm{V}_{\mathrm{ol}}=\mathrm{V}_{\mathrm{dD}} / 2$ | (VdoQ/2) / (1.15 x RQ/5) | (VdoQ/2) / (0.85 x RQ/5) | mA | 3,4 |

Notes 1. Outputs are impedance-controlled. | Іон | = (VdDQ/2)/(RQ/5) for values of $125 \Omega \leq R Q \leq 300 \Omega$.
2. Outputs are impedance-controlled. lol = (VdDQ/2)/(RQ/5) for values of $125 \Omega \leq R Q \leq 300 \Omega$.
3. Іон and lol are defined as absolute values and are measured at $V_{D D Q} / 2$. Іон flows from the device, lol flows into the device.
4. If $M R S$ bit $A 8$ is 0 , use $R Q=250 \Omega$ in the equation in lieu of presence of an external impedance matched resistor.

Capacitance ( $\mathrm{TA}=\mathbf{2 5}^{\circ} \mathrm{C}, \mathrm{f}=1 \mathrm{MHz}$ )

| Parameter | Symbol | Test conditions | MIN. | MAX. | Unit |
| :--- | :---: | :--- | :---: | :---: | :---: |
| Address / Control Input capacitance | $\mathrm{C}_{\mathrm{IN}}$ | $\mathrm{V}_{1 \mathrm{~N}}=0 \mathrm{~V}$ | 1.5 | 2.5 | pF |
| I/O, Output, Other capacitance <br> (DQ, DM, QK, QVLD) | $\mathrm{Cl}_{/ / \mathrm{O}}$ | $\mathrm{V}_{/ / \mathrm{O}}=0 \mathrm{~V}$ | 3.5 | 5.0 | pF |
| Clock Input capacitance | $\mathrm{C}_{\mathrm{clk}}$ | $\mathrm{V}_{\mathrm{clk}}=0 \mathrm{~V}$ |  |  |  |
| JTAG pins | $\mathrm{C}_{J}$ | $\mathrm{~V}_{J}=0 \mathrm{~V}$ | 2.0 | pF |  |

Remark These parameters are periodically sampled and not $100 \%$ tested.
Capacitance is not tested on ZQ pin.

## Recommended AC Operating Conditions

$0^{\circ} \mathrm{C} \leq \mathrm{Tc} \leq 95^{\circ} \mathrm{C} ; 1.7 \mathrm{~V} \leq \mathrm{VDD} \leq 1.9 \mathrm{~V}$, unless otherwise noted

| Parameter | Symbol | Conditions | MIN. | MAX. | Unit | Note |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input HIGH voltage | $\mathrm{V}_{\mathrm{IH}}(\mathrm{AC})$ |  | $V_{\text {Ref }}+0.2$ |  | V | 1 |
| Input LOW voltage | VIL (AC) |  |  | Vref -0.2 | V | 1 |

Note 1. Overshoot: $\mathrm{V}_{\mathrm{IH}}(\mathrm{AC}) \leq \mathrm{V}_{\mathrm{DD}} \mathrm{Q}+0.7 \mathrm{~V}$ for $\mathrm{t} \leq \mathrm{tck} / 2$
Undershoot: VIL (AC) $\geq-0.5 \mathrm{~V}$ for $\mathrm{t} \leq \mathrm{tck} / 2$
Control input signals may not have pulse widths less than tскн (MIN.) or operate at cycle rates less than tck (MIN.).

## DC Characteristics

Idd / Isb Operating Conditions

| Parameter | Symbol | Test condition |  |  | MAX. |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | -EF25 | $\begin{aligned} & \text {-E33, } \\ & \text {-EF33 } \end{aligned}$ |  |
| Standby current | IsB1 | $\mathrm{tck}=\mathrm{Idle}$ <br> All banks idle, no inputs toggling | VDD | x36 | 48 | 48 | mA |
|  |  |  |  | x18/x9 | 48 | 48 |  |
|  |  |  | Vext |  | 26 | 26 |  |
| Active standby current | IsB2 | CS\# = HIGH, No commands, half bank / address / data change once every four clock cycles | VdD | x36 | 288 | 233 | mA |
|  |  |  |  | x18/x9 | 288 | 233 |  |
|  |  |  | Vext |  | 26 | 26 |  |
| Operating current | ldD1 | $B L=2$, sequential bank access, bank transitions once every $t_{\text {Rc }}$, half address transitions once every trc, read followed by write sequence, continuous data during WRITE commands. | VDD | x36 | 390 | 350 | mA |
|  |  |  |  | x18/x9 | 365 | 325 |  |
|  |  |  | Vext |  | 41 | 36 |  |
| Operating current | IdD2 | $B L=4$, sequential bank access, bank transitions once every $t_{\text {Rc }}$, half address transitions once every trc, read followed by write sequence, continuous data during WRITE commands. | VdD | x36 | 415 | 385 | mA |
|  |  |  |  | x18/x9 | 360 | 340 |  |
|  |  |  | Vext |  | 48 | 42 |  |
| Operating current | IdD3 | $B L=8$, sequential bank access, bank transitions once every trc, half address transitions once every trc, read followed by write sequence, continuous data during WRITE commands. | VDD | x36 | - | - | mA |
|  |  |  |  | x18/x9 | 400 | 360 |  |
|  |  |  | Vext |  | 55 | 48 |  |
| Burst refresh current | $I_{\text {REF1 }}$ | Eight bank cyclic refresh, continuous address/data, command bus remains in refresh for all banks | VDD | x36 | 650 | 540 | mA |
|  |  |  |  | x18/x9 | 650 | 540 |  |
|  |  |  | Vext |  | 133 | 111 |  |
| Disturbed refresh current | Iref2 | Single bank refresh, sequential bank access, half address transitions once every trc, continuous data | VdD | x36 | 320 | 270 | mA |
|  |  |  |  | x18/x9 | 310 | 260 |  |
|  |  |  | Vext |  | 48 | 42 |  |
| Operating burst write current | IDD2w | BL=2, cyclic bank access, half of address bits change every clock cycle, continuous data, measurement is taken during continuous WRITE | VDD | x36 | 990 | 870 | mA |
|  |  |  |  | x18/x9 | 970 | 820 |  |
|  |  |  | Vext |  | 100 | 90 |  |
| Operating burst write current | IDD4W | $B L=4$, cyclic bank access, half of address bits change every two clocks, continuous data, measurement is taken during continuous WRITE | VDD | x36 | 770 | 610 | mA |
|  |  |  |  | x18/x9 | 690 | 560 |  |
|  |  |  | Vext |  | 88 | 77 |  |
| Operating burst write current | IdD8w | $B L=8$, cyclic bank access, half of address bits change every four clocks, continuous data, measurement is taken during continuous WRITE | VdD | x36 | - | - | mA |
|  |  |  |  | x18/x9 | 600 | 450 |  |
|  |  |  | Vext |  | 60 | 51 |  |
| Operating burst read current | IDD2R | BL=2, cyclic bank access, half of address bits change every clock cycle, measurement is taken during continuous READ | VdD | x36 | 1,030 | 900 | mA |
|  |  |  |  | x18/x9 | 970 | 840 |  |
|  |  |  | Vext |  | 100 | 90 |  |
| Operating burst read current | IDD4R | $B L=4$, cyclic bank access, half of address bits change every two clocks, measurement is taken during continuous READ | Vod | x36 | 780 | 630 | mA |
|  |  |  |  | x18/x9 | 720 | 580 |  |
|  |  |  | Vext |  | 88 | 77 |  |
| Operating burst read current | IDD8R | $B L=8$, cyclic bank access, half of address bits change every four clocks, measurement is taken during continuous READ | VdD | x36 | - | - | mA |
|  |  |  |  | x18/x9 | 550 | 450 |  |
|  |  |  | Vext |  | 60 | 51 |  |

Remarks 1. IdD specifications are tested after the device is properly initialized. $0^{\circ} \mathrm{C} \leq \mathrm{Tc} \leq 95^{\circ} \mathrm{C} ; 1.7 \mathrm{~V} \leq \mathrm{VDD} \leq 1.9 \mathrm{~V}$,

2. $\mathrm{tck}=\mathrm{t} \mathrm{to}=\mathrm{MIN} ., \mathrm{trc}=\mathrm{MIN}$.
3. Input slew rate is specified in Recommended DC Operating Conditions and Recommended AC Operating Conditions.
4. IdD parameters are specified with ODT disabled.
5. Continuous data is defined as half the DQ signals changing between HIGH and LOW every half clock cycles (twice per clock).
6. Continuous address is defined as half the address signals between HIGH and LOW every clock cycles (once per clock).
7. Sequential bank access is defined as the bank address incrementing by one ever trc.
8. Cyclic bank access is defined as the bank address incrementing by one for each command access. For $B L=4$ this is every other clock.
9. CS\# is HIGH unless a READ, WRITE, AREF, or MRS command is registered. CS\# never transitions more than per clock cycle.

## AC Characteristics

## AC Test Conditions

Input waveform


Output waveform


Output load condition


## AC Characteristics <Read and Write Cycle>

$V_{D D Q}=1.8 \mathrm{~V}$

| Parameter | Symbol | $\begin{gathered} -E 33 \\ (300 \mathrm{MHz}) \end{gathered}$ |  | Unit | Note |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |
|  |  | MIN. | MAX. |  |  |
| Clock |  |  |  |  |  |
| Clock cycle time (CK,CK\#,DK,DK\#) | tck, tok | 3.3 | 5.7 | ns |  |
| Clock frequency (CK,CK\#,DK,DK\#) | tck, tok | 175 | 300 | MHz |  |
| Random Cycle time | trc | 20 |  | ns |  |
| Clock Jitter: period | tJIT PER | -200 | 200 | ps | 1, 2 |
| Clock Jitter: cycle-to-cycle | tıIt cc |  | 400 | ps |  |
| Clock HIGH time (CK,CK\#,DK,DK\#) | tскн, tокн | 0.45 | 0.55 | Cycle |  |
| Clock LOW time (CK,CK\#,DK,DK\#) | tckl, tokl | 0.45 | 0.55 | Cycle |  |
| Clock to input data clock | tckdk | -0.3 | 1.0 | ns |  |
| Mode register set cycle time to any command | $t_{\text {mRSC }}$ | 6 |  | Cycle |  |
| PLL Lock time | tck Lock | 15 |  | $\mu \mathrm{s}$ |  |
| Clock static to PLL reset | tck Reset | 30 |  | ns |  |
| Output Times |  |  |  |  |  |
| Output data clock HIGH time | tهкн | 0.9 | 1.1 | tck |  |
| Output data clock LOW time | takı | 0.9 | 1.1 | tckL |  |
| QK edge to clock edge skew | tскак | -0.3 | 0.3 | ns |  |
| QK edge to output data edge | takqo, takQ1 | -0.25 | 0.25 | ns | 3, 5 |
| QK edge to any output data | toko | -0.35 | 0.35 | ns | 4,5 |
| QK edge to QVLD | tekvld | -0.35 | 0.35 | ns |  |
| Setup Times |  |  |  |  |  |
| Address/command and input | tas/tcs | 0.5 |  | ns |  |
| Data-in and data mask to DK | tos | 0.3 |  | ns |  |
| Hold Times |  |  |  |  |  |
| Address/command and input | $\mathrm{tah}^{\text {/tch }}$ | 0.5 |  | ns |  |
| Data-in and data mask to DK | toh | 0.3 |  | ns |  |

Notes 1. Clock phase jitter is the variance from clock rising edge to the next expected clock rising edge.
2. Frequency drift is not allowed.
3. taкяo is referenced to DQ0-DQ17 in $\times 36$ and DQ0-DQ8 in $\times 18$. tQKQ1 is referenced to DQ18-DQ35 in $\times 36$ and DQ9-DQ17 in $\times 18$.
4. takQ takes into account the skew between any $Q K x$ and any $D Q$.
5. tqкя, tякох are guaranteed by design.

Remark All timing parameters are measured relative to the crossing point of CK/CK\#, DK/DK\# and to the crossing point with $V_{\text {ref }}$ of the command, address, and data signals.

## AC Characteristics <Read and Write Cycle>

$\mathrm{Vdo} Q=1.5 \mathrm{~V}$

| Parameter | Symbol | $\begin{gathered} -E F 25 \\ (400 \mathrm{MHz}) \\ \hline \end{gathered}$ |  | $\begin{gathered} \hline-E F 33 \\ (300 \mathrm{MHz}) \\ \hline \end{gathered}$ |  | Unit | Note |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | MIN. | MAX. | MIN. | MAX. |  |  |
| Clock |  |  |  |  |  |  |  |
| Clock cycle time (CK,CK\#,DK,DK\#) | tck, tok | 2.5 | 5.7 | 3.3 | 5.7 | ns |  |
| Clock frequency (CK,CK\#,DK,DK\#) | tck, tok | 175 | 400 | 175 | 300 | MHz |  |
| Random Cycle time | trc | 20 |  | 20 |  | ns |  |
| Clock Jitter: period | tJIt PER | -150 | 150 | -200 | 200 | ps | 1,2 |
| Clock Jitter: cycle-to-cycle | tııt cc |  | 300 |  | 400 | ps |  |
| Clock HIGH time (CK,CK\#,DK,DK\#) | tскн, tокн | 0.45 | 0.55 | 0.45 | 0.55 | Cycle |  |
| Clock LOW time (CK,CK\#,DK,DK\#) | tckL, tokl | 0.45 | 0.55 | 0.45 | 0.55 | Cycle |  |
| Clock to input data clock | tckdk | -0.3 | 0.5 | -0.3 | 1.0 | ns |  |
| Mode register set cycle time to any command | tmrsc | 6 |  | 6 |  | Cycle |  |
| PLL Lock time | tck Lock | 15 |  | 15 |  | $\mu \mathrm{s}$ |  |
| Clock static to PLL reset | tck Reset | 30 |  | 30 |  | ns |  |
| Output Times |  |  |  |  |  |  |  |
| Output data clock HIGH time | tокн | 0.9 | 1.1 | 0.9 | 1.1 | tскн |  |
| Output data clock LOW time | tokı | 0.9 | 1.1 | 0.9 | 1.1 | tckl |  |
| QK edge to clock edge skew | tскок | -0.25 | 0.25 | -0.3 | 0.3 | ns |  |
| QK edge to output data edge | tQkQo, tokQ1 | -0.2 | 0.2 | -0.25 | 0.25 | ns | 3,5 |
| QK edge to any output data | tоко | -0.3 | 0.3 | -0.35 | 0.35 | ns | 4, 5 |
| QK edge to QVLD | tqkVLD | -0.3 | 0.3 | -0.35 | 0.35 | ns |  |
| Setup Times |  |  |  |  |  |  |  |
| Address/command and input | tas/tcs | 0.4 |  | 0.5 |  | ns |  |
| Data-in and data mask to DK | tos | 0.25 |  | 0.3 |  | ns |  |
| Hold Times |  |  |  |  |  |  |  |
| Address/command and input | tah/tch | 0.4 |  | 0.5 |  | ns |  |
| Data-in and data mask to DK | tD ${ }^{\text {d }}$ | 0.25 |  | 0.3 |  | ns |  |

Notes 1. Clock phase jitter is the variance from clock rising edge to the next expected clock rising edge.
2. Frequency drift is not allowed.
3. takqo is referenced to DQ0-DQ17 in $\times 36$ and DQ0-DQ8 in $\times 18$. tQкQ1 is referenced to DQ18-DQ35 in $\times 36$ and DQ9-DQ17 in $\times 18$.
4. tокQ takes into account the skew between any QKx and any DQ.
5. tякя, tякох are guaranteed by design.

Remark All timing parameters are measured relative to the crossing point of CK/CK\#, DK/DK\# and to the crossing point with $V_{\text {ref }}$ of the command, address, and data signals.

Figure 1-1. Clock / Input Data Clock Command / Address Timings


## Temperature and Thermal Impedance

Temperature Limits

| Parameter | Symbol | MIN. | MAX. | Unit | Note |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Reliability junction temperature | $\mathrm{T}_{J}$ | 0 | +110 | ${ }^{\circ} \mathrm{C}$ | 1 |
| Operating junction temperature | $\mathrm{T}_{J}$ | 0 | +100 | ${ }^{\circ} \mathrm{C}$ | 2 |
| Operating case temperature | $\mathrm{T}_{\mathrm{c}}$ | 0 | +95 | ${ }^{\circ} \mathrm{C}$ | 3 |

Notes 1. Temperatures greater than $110^{\circ} \mathrm{C}$ may cause permanent damage to the device. This is a stress rating only and functional operation of the device at or above this is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability of the part.
2. Junction temperature depends upon cycle time, loading, ambient temperature, and airflow.
3. MAX operating case temperature; $\mathrm{Tc}_{\mathrm{c}}$ is measured in the center of the package. Device functionality is not guaranteed if the device exceeds maximum Tc during operation.

## Thermal Impedance

| Substrate | Ball | $\theta \mathrm{ja}\left({ }^{\circ} \mathrm{C} / \mathrm{W}\right)$ |  |  | $\begin{gathered} \theta \mathrm{jb} \\ \left({ }^{\circ} \mathrm{C} / \mathrm{W}\right) \end{gathered}$ | $\theta j c$ ( ${ }^{\circ} \mathrm{C} / \mathrm{W}$ ) |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Air Flow $=0 \mathrm{~m} / \mathrm{s}$ | Air Flow $=1 \mathrm{~m} / \mathrm{s}$ | Air Flow $=2 \mathrm{~m} / \mathrm{s}$ |  |  |
| 4 - Layer | Lead | 32.4 | 26.8 | 24.6 | 23.0 | 1.8 |
| 8 - Layer | Lead | 26.5 | 22.3 | 20.8 | 16.8 | 1.8 |
| 4 - Layer | Lead free | 32.1 | 26.6 | 24.4 | 22.7 | 1.8 |
| 8 - Layer | Lead free | 26.3 | 22.1 | 20.6 | 16.6 | 1.8 |

## 2. Operation

### 2.1 Command Operation

According to the functional signal description, the following command sequences are possible. All input states or sequences not shown are illegal or reserved. All command and address inputs must meet setup and hold times around the rising edge of CK.

Table 2-1. Address Widths at Different Burst Lengths

| Burst Length | Configuration |  |  |
| :--- | :---: | :---: | :---: |
|  | $\mathbf{x} 36$ | $\mathbf{x 1 8}$ | $\mathbf{x 9}$ |
| $\mathrm{BL}=2$ | $\mathrm{~A} 0-\mathrm{A} 18$ | $\mathrm{~A} 0-\mathrm{A} 19$ | $\mathrm{~A} 0-\mathrm{A} 20$ |
| $\mathrm{BL}=4$ | $\mathrm{~A} 0-\mathrm{A} 17$ | $\mathrm{~A} 0-\mathrm{A} 18$ | A |
| $\mathrm{BL}=8$ | NA | A | A 19 |

Table 2-2. Command Table

| Operation | Code | CS\# | WE\# | REF\# | A0-An $^{\text {Note1 }}$ | BA0-BA2 | Note |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Device DESELECT / No Operation | DESEL / NOP | H | X | X | X | X |  |
| MRS: Mode Register Set | MRS | L | L | L | OPCODE | X | 2 |
| READ | READ | L | H | H | A | BA | 3 |
| WRITE | WRITE | L | L | H | A | BA | 3 |
| AUTO REFRESH | AREF | L | H | L | X | BA |  |

Notes 1. $\mathrm{n}=20$
2. Only A0-A17 are used for the MRS command.
3. See Table 2-1

Remark $\mathrm{X}=$ "Don't Care", $\mathrm{H}=$ logic HIGH, L = logic LOW, $\mathrm{A}=$ valid address, BA = valid bank address

### 2.2 Description of Commands

DESEL / NOP ${ }^{\text {Note1 }}$
The NOP command is used to perform a no operation to the $\mu$ PD48288209/18/36-A, which essentially deselects the chip. Use the NOP command to prevent unwanted commands from being registered during idle or wait states. Operations already in progress are not affected. Output values depend on command history.

## MRS

The mode register is set via the address inputs A0-A17. See Figure 2-5. Mode Register Bit Map for further information. The MRS command can only be issued when all banks are idle and no bursts are in progress.

## READ

The READ command is used to initiate a burst read access to a bank. The value on the BA0-BA2 inputs selects the bank, and the address provided on inputs A0-A20 selects the data location within the bank.

## WRITE

The WRITE command is used to initiate a burst write access to a bank. The value on the BA0-BA2 inputs selects the bank, and the address provided on inputs A0-A20 selects the data location within the bank. Input data appearing on the DQ is written to the memory array subject to the DM input logic level appearing coincident with the data. If the DM signal
is registered LOW, the corresponding data will be written to memory. If the DM signal is registered HIGH, the corresponding data inputs will be ignored (i.e., this part of the data word will not be written).

## AREF

The AREF is used during normal operation of the $\mu$ PD48288209/18/36-A to refresh the memory content of a bank. The command is non-persistent, so it must be issued each time a refresh is required. The value on the BA0-BA2 inputs selects the bank. The refresh address is generated by an internal refresh controller, effectively making each address bit a "Don't Care" during the AREF command. The $\mu$ PD48288209/18/36-A requires 64 K cycles at an average periodic interval of $0.49 \mu \mathrm{~s}$ Note2 (MAX.). To improve efficiency, eight AREF commands (one for each bank) can be posted to $\mu \mathrm{PD} 48288209 / 18 / 36-\mathrm{A}$ at periodic intervals of $3.9 \mu \mathrm{~s}{ }^{\text {Note3 }}$.

Within a period of 32 ms , the entire memory must be refreshed. The delay between the AREF command and a subsequent command to same bank must be at least trc as continuous refresh. Other refresh strategies, such as burst refresh, are also possible.

Notes 1. When the chip is deselected, internal NOP commands are generated and no commands are accepted.
2. Actual refresh is $32 \mathrm{~ms} / 8 \mathrm{k} / 8=0.488 \mu \mathrm{~s}$.
3. Actual refresh is $32 \mathrm{~ms} / 8 \mathrm{k}=3.90 \mu \mathrm{~s}$.

### 2.3 Initialization

The $\mu$ PD48288209/18/36-A must be powered up and initialized in a predefined manner. Operational procedures other than those specified may result in undefined operations or permanent damage to the device. The following sequence is used for Power-Up:

1. Apply power (VExt, $\left.V_{d D}, V_{D D Q}, V_{R E F}, V_{T T}\right)$ and start clock as soon as the supply voltages are stable. Apply $V_{d D}$ and $V_{\text {ext }}$ before or at the same time as $V_{d d Q}$. Apply VddQ before or at the same time as $V_{\text {ref }}$ and $V_{t t}$. Although there is no timing relation between $V_{E X T}$ and $V_{D D}$, the chip starts the power-up sequence only after both voltages are at their nominal levels. VdDQ supply must not be applied before Vdd supply. CK/CK\# must meet VID(DC) prior to being applied. Maintain all remaining balls in NOP conditions.

Note No rule of apply power sequence is the design target.
2. Maintain stable conditions for $200 \mu \mathrm{~s}$ (MIN.).
3. Issue three or more back-to-back and clock consecutive MRS commands: two dummies plus one valid MRS. It is recommended that the dummy MRS commands are the same value as the desired MRS.
4. tmRsc after valid MRS, an AUTO REFRESH command to all 8 banks must be issued and wait for $15 \mu \mathrm{~s}$ with CK/CK\# toggling in order to lock the PLL prior to normal operation.
5. After trc, the chip is ready for normal operation.

### 2.4 Power-On Sequence

Figure 2-1. Power-Up Sequence


Notes 1. Recommended all address pins held LOW during dummy MRS commands.
2. A10-A17 must be LOW.

Remark MRS:MRS command
RFp : REFRESH bank p
AC : Any command

### 2.5 Programmable Impedance Output Buffer

The $\mu$ PD48288209/18/36-A is equipped with programmable impedance output buffers. This allows a user to match the driver impedance to the system. To adjust the impedance, an external precision resistor ( RQ ) is connected between the ZQ ball and Vss. The value of the resistor must be five times the desired impedance. For example, a $300 \Omega$ resistor is required for an output impedance of $60 \Omega$. To ensure that output impedance is one fifth the value of $R Q$ (within 15 percent), the range of $R Q$ is $125 \Omega$ to $300 \Omega$. Output impedance updates may be required because, over time, variations may occur in supply voltage and temperature. The device samples the value of RQ. An impedance update is transparent to the system and does not affect device operation. All data sheet timing and current specifications are met during an update.

### 2.6 PLL Reset

The $\mu$ PD48288209/18/36-A utilizes internal Phase-locked loops for maximum output, data valid windows. It can be placed into a stopped-clock state to minimize power with a modest restart time of $15 \mu \mathrm{~s}$. The clock (CK/CK\#) must be toggled for $15 \mu \mathrm{~s}$ in order to stabilize PLL circuits for next READ operation.

### 2.7 Clock Input

Table 2-3. Clock Input Operation Conditions

| Parameter | Symbol | Conditions | MIN. | MAX. | Unit | Note |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Clock Input Voltage Level | $\mathrm{VIN}(\mathrm{DC})$ | CK and CK\# | -0.3 | $V_{\text {do }}+0.3$ | V |  |
| Clock Input Differential Voltage Level | VID (DC) | CK and CK\# | 0.2 | VDDQ +0.6 | V | 8 |
| Clock Input Differential Voltage Level | $\mathrm{VID} \mathrm{(AC)}$ | CK and CK\# | 0.4 | VDDQ +0.6 | V | 8 |
| Clock Input Crossing Point Voltage Level | VIX (AC) | CK and CK\# | VdoQ/2-0.15 | $V_{\text {doQ }} / 2+0.15$ | V | 9 |

Figure 2-2. Clock Input


Notes 1. DKx and DKx\# have the same requirements as CK and CK\#.
2. All voltages referenced to Vss.
3. Tests for AC timing, IDD and electrical AC and DC characteristics may be conducted at normal reference/supply voltage levels; but the related specifications and device operations are tested for the full voltage range specified.
4. $A C$ timing and IDD tests may use a $\mathrm{V}_{\mathrm{IL}}$ to $\mathrm{V}_{\mathrm{IH}}$ swing of up to 1.5 V in the test environment, but input timing is still referenced to VREF (or the crossing point for CK/CK\#), and parameters specifications are tested for the specified AC input levels under normal use conditions. The minimum slew rate for the input signals used to test the device is $2 \mathrm{~V} / n$ n in the range between $\mathrm{V}_{\mathrm{IL}}(\mathrm{AC})$ and $\mathrm{VIH}_{\mathrm{I}}(\mathrm{AC})$.
5. The AC and DC input level specifications are as defined in the HSTL Standard (i.e. the receiver will effectively switch as a result of the signal crossing the AC input level, and will remain in that state as long as the signal does not ring back above[below] the DC input LOW[HIGH] level).
6. The CK/CK\# input reference level (for timing referenced to CK/CK\#) is the point at which CK and CK\# cross. The input reference level for signal other than CK/CK\# is VrEF.
7. CK and $\mathrm{CK} \mathrm{\#}$ input slew rate must be $>=2 \mathrm{~V} / \mathrm{ns}$ (>= $4 \mathrm{~V} / \mathrm{ns}$ if measured differentially).
8. $\mathrm{V}_{\mathrm{I}}$ is the magnitude of the difference between the input level on CK and input level on CK\#.
9. The value of $V_{1 x}$ is expected to equal $V_{d D Q} / 2$ of the transmitting device and must track variations in the DC level of the same.
10. CK and CK\# must cross within the region.
11. $C K$ and $C K \#$ must meet at least $\operatorname{VID(DC)}$ (MIN.) when static and centered around $V_{D D Q} / 2$.
12. Minimum peak-to-peak swing.

### 2.8 Mode Register Set Command (MRS)

The mode register stores the data for controlling the operating modes of the memory. It programs the $\mu$ PD48288209/18/36-A configuration, burst length, and I/O options. During a MRS command, the address inputs A0-A17 are sampled and stored in the mode register. tmRsc must be met before any command can be issued to the $\mu$ PD48288209/18/36-A. The mode register may be set at any time during device operation. However, any pending operations are not guaranteed to successfully complete.

Since MRS is used for internal test mode entry, the designated bit at Figure 2-5. Mode Register Bit Map and Figure 2-27. Mode Register Set Command in Multiplexed Address Mode should be set.

Figure 2-3. Mode Register Set Timing


Figure 2-4. Mode Register Set


Remark COD: code to be loaded into the register.

Figure 2-5. Mode Register Bit Map


Notes 1. Bits A10-A17 must be set to all ' 0 '. A18-An are "Don't Care".
2. $B L=8$ is not available for configuration 1 .
3. $\pm 30 \%$ temperature variation.
4. Within $15 \%$.

### 2.9 Read \& Write configuration (Non Multiplexed Address Mode)

Table 2-4 shows, for different operating frequencies, the different $\mu$ PD48288209/18/36-A configurations that can be programmed into the mode register. The READ and WRITE latency (trL and twL) values along with the row cycle times $(\operatorname{trc})$ are shown in clock cycles as well as in nanoseconds. The shaded areas correspond to configurations that are not allowed.

Table 2-4. Configuration Table

| Frequency | Symbol | Configuration |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $1^{\text {Note }}$ | 2 | 3 |  |
|  | trc | 4 | 6 | 8 | Cycles |
|  | trL | 4 | 6 | 8 | Cycles |
|  | twL | 5 | 7 | 9 | Cycles |
| 400 MHz | trc |  |  | 20.0 | ns |
|  | tRL |  |  | 20.0 | ns |
|  | twL |  |  | 22.5 | ns |
| 300 MHz | trc |  | 20.0 | 26.7 | ns |
|  | trL |  | 20.0 | 26.7 | ns |
|  | twL |  | 23.3 | 30.0 | ns |
| 200 MHz | trc | 20.0 | 30.0 | 40.0 | ns |
|  | tRL | 20.0 | 30.0 | 40.0 | ns |
|  | twL | 25.0 | 35.0 | 45.0 | ns |

Note $B L=8$ is not available for configuration 1.

### 2.10 Write Operation (WRITE)

Write accesses are initiated with a WRITE command, as shown in Figure 2-6. Row and bank addresses are provided together with the WRITE command. During WRITE commands, data will be registered at both edges of DK according to the programmed burst length (BL). A WRITE latency (WL) one cycle longer than the programmed READ latency (RL +1 ) is present, with the first valid data registered at the first rising DK edge WL cycles after the WRITE command.

Any WRITE burst may be followed by a subsequent READ command. Figure 2-10. WRITE Followed By READ: BL=2, RL=4, WL=5, Configuration 1 and Figure 2-11. WRITE Followed By READ: BL=4, RL=4, WL=5, Configuration 1 illustrate the timing requirements for a WRITE followed by a READ for bursts of two and four, respectively.
Setup and hold times for incoming input data relative to the DK edges are specified as tos and tor. The input data is masked if the corresponding DM signal is HIGH. The setup and hold times for data mask are also tos and tor.

Figure 2-6. WRITE Command


Remark A : Address
BA : Bank address

Figure 2-7. Basic WRITE Burst / DM Timing


V/A Don't care

Figure 2-8. WRITE Burst Basic Sequence: $B L=2, R L=4, W L=5$, Configuration 1


Figure 2-9. WRITE Burst Basic Sequence: BL=4, RL=4, WL=5, Configuration 1


Remarks 1. WR : WRITE command
A/BAp: Address A of bank p
WL : WRITE latency
Dpq : Data q to bank p
2. Any free bank may be used in any given command. The sequence shown is only one example of a bank sequence.

Figure 2-10. WRITE Followed By READ: BL=2, RL=4, WL=5, Configuration 1


V/A Don't care Undefined

Figure 2-11. WRITE Followed By READ: BL=4, RL=4, WL=5, Configuration 1


Remark WR : WRITE command
RD : READ command
A/BAp : Address A of bank p
WL : WRITE latency
RL : READ latency
Dpq : Data q to bank p
Qpq : Data q from bank p

### 2.11 Read Operation (READ)

Read accesses are initiated with a READ command, as shown in Figure 2-12. Row and bank addresses are provided with the READ command.
During READ bursts, the memory device drives the read data edge-aligned with the QK signal. After a programmable READ latency, data is available at the outputs. The data valid signal indicates that valid data will be present in the next half clock cycle.
The skew between QK and the crossing point of CK is specified as tскок. taкøo is the skew between QKO and the last valid data edge considered the data generated at the DQ0-DQ17 in $x 36$ and DQ0-DQ8 in $x 18$ data signals. tokQ1 is the skew between QK1 and the last valid data edge considered the data generated at the DQ18-DQ35 in x36 and DQ9DQ17 in x18 data signals. tokQx is derived at each QKx clock edge and is not cumulative over time.
After completion of a burst, assuming no other commands have been initiated, DQ will go High-Z. Back-to-back READ commands are possible, producing a continuous flow of output data.
Minimum READ data valid window can be expressed as MIN.(taкн, toкц) $-2 \times$ MAX.(taкQx)
Any READ burst may be followed by a subsequent WRITE command. Figure 2-16. READ followed by WRITE, BL=2, RL=4, WL=5, Configuration 1 and Figure 2-17. READ followed by WRITE, BL=4, RL=4, WL=5, Configuration 1 illustrate the timing requirements for a READ followed by a WRITE.

Figure 2-12. READ Command


Figure 2-13. Basic READ Burst Timing


Undefined

Note 1. Minimum READ data valid window can be expressed as MIN.(tякн, taкц) $-2 \times$ MAX.(tøках) tскн and tckl are recommended to have $50 \% / 50 \%$ duty.

Remarks 1. taк⿰o is referenced to DQ0-DQ17 in $x 36$ and DQ0-DQ8 in $x 18$.
toкQ1 is referenced to DQ18-DQ35 in $\times 36$ and DQ9-DQ17 in $\times 18$.
2. toка takes into account the skew between any $Q K x$ and any $D Q$.
3. tскяк is specified as $C K$ rising edge to QK rising edge.

Figure 2-14. READ Burst Basic Sequence: BL=2, RL=4, Configuration 1


Figure 2-15. READ Burst Basic Sequence: $B L=4, R L=4$, Configuration 1


ADDRESS



Remark RD : READ command
A/BAp: Address $A$ of bank $p$
RL : READ latency
Qpq : Data q from bank p

Figure 2-16. READ followed by WRITE, BL=2, RL=4, WL=5, Configuration 1


Figure 2-17. READ followed by WRITE, BL=4, RL=4, WL=5, Configuration 1


Remark WR : WRITE command
RD : READ command
A/BAp : Address A of bank p
WL : WRITE latency
RL : READ latency
Dpq : Data q to bank $p$
Qpq : Data q from bank p

### 2.12 Refresh Operation: AUTO REFRESH Command (AREF)

AREF is used to perform a REFRESH cycle on one row in a specific bank. The row addresses are generated by an internal refresh counter; external address balls are "Don't Care." The delay between the AREF command and a subsequent command to the same bank must be at least trc.
Within a period of 32 ms (tref), the entire memory must be refreshed. Figure 2-19 illustrates an example of a continuous refresh sequence. Other refresh strategies, such as burst refresh, are also possible.

Figure 2-18. AUTO REFRESH Command


Don't care

Remark BA: Bank address

Figure 2-19. AUTO REFRESH Cycle


Remarks 1. ACx : Any command on bank $x$
ARFx: Auto refresh bank x
ACy : Any command on different bank.
2. trC $^{\text {is configuration-dependent. Refer to Table 2-4. Configuration Table. }}$

### 2.13 On-Die Termination

On-die termination (ODT) is enabled by setting A9 to " 1 " during an MRS command. With ODT on, all the DQs and DM are terminated to $V_{\text {тт }}$ with a resistance $\mathrm{R}_{\text {тт. }}$. The command, address, and clock signals are not terminated. Figure 2-20 below shows the equivalent circuit of a DQ receiver with ODT. ODTs are dynamically switched off during READ commands and are designed to be off prior to the $\mu$ PD48288209/18/36-A driving the bus. Similarly, ODTs are designed to switch on after the $\mu \mathrm{PD} 48288209 / 18 / 36-\mathrm{A}$ has issued the last piece of data.

Table 2-5. On-Die Termination DC Parameters

| Description | Symbol | MIN. | MAX. | Units | Note |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Termination voltage | $\mathrm{V}_{\mathrm{TT}}$ | $0.95 \times \mathrm{V}_{\mathrm{REF}}$ | $1.05 \times \mathrm{V}_{\mathrm{REF}}$ | V | 1,2 |
| On-Die termination | $\mathrm{R}_{\mathrm{TT}}$ | 125 | 185 | $\Omega$ | 3 |

Notes 1. All voltages referenced to Vss (GND).
2. $V_{T t}$ is expected to be set equal to $V_{\text {ref }}$ and must track variations in the $D C$ level of $V_{\text {ReF }}$.
3. The $\mathrm{R}_{\mathrm{Tt}}$ value is measured at $95^{\circ} \mathrm{C} \mathrm{Tc}$.

Figure 2-20. On- Die Termination-Equivalent Circuit


Figure 2-21. READ Burst with ODT: BL=2, Configuration 1


Remark RD : READ command
A/BAp : Address A of bank p
RL : READ latency
Qpq : Data q from bank $p$

Figure 2-22. READ NOP READ with ODT: BL=2, Configuration 1


Figure 2-23. READ NOP NOP READ with ODT: BL=2, Configuration 1


Remark RD : READ command A/BAp: Address $A$ of bank $p$
RL : READ latency
Qpq : Data q from bank p

Figure 2-24. READ followed by WRITE with ODT: BL=2, Configuration 1


Figure 2-25. WRITE followed by READ with ODT: BL=2, Configuration 1


Remark RD : READ command
WR : WRITE command
A/BAp: Address A of bank p
RL : READ latency
WL : WRITE latency
Qpq : Data q from bank p
Dpq : Data q to bank p

### 2.14 Operation with Multiplexed Address

In multiplexed address mode, the address can be provided to the $\mu \mathrm{PD} 48288209 / 18 / 36-\mathrm{A}$ in two parts that are latched into the memory with two consecutive rising clock edges. This provides the advantage that a maximum of 11 address balls are required to control the $\mu$ PD48288209/18/36-A, reducing the number of balls on the controller side. The data bus efficiency in continuous burst mode is not affected for $B L=4$ and $B L=8$ since at least two clocks are required to read the data out of the memory. The bank addresses are delivered to the $\mu$ PD48288209/18/36-A at the same time as the WRITE command and the first address part, Ax.

This option is available by setting bit A5 to " 1 " in the mode register. Once this bit is set, the READ, WRITE, and MRS commands follow the format described in Figure 2-26. See Figure 2-28. Power-Up Sequence in Multiplexed Address Mode for the power-up sequence.

Figure 2-26. Command Description in Multiplexed


Remarks 1. $\mathrm{Ax}, \mathrm{Ay}$ : Address
BA : Bank Address
2. The minimum setup and hold times of the two address parts are defined tas and tar.

Figure 2-27. Mode Register Set Command in Multiplexed Address Mode


Notes 1. Bits A10-A17 must be set to all ' 0 '.
2. $\mathrm{BL}=8$ is not available for configuration 1 .
3. $\pm 30 \%$ temperature variation.
4. Within $15 \%$.

Remark The address A0, A3, A4, A5, A8, and A9 must be set as follows in order to activate the mode register in the multiplexed address mode.

Figure 2-28. Power-Up Sequence in Multiplexed Address Mode


Notes 1. Recommended all address pins held LOW during dummy MRS command.
2. A10-A17 must be LOW.
3. Address A5 must be set HIGH (muxed address mode setting when $\mu$ PD48288209/18/36-A is in normal mode of operation).
4. Address A5 must be set HIGH (muxed address mode setting when $\mu$ PD48288209/18/36-A is already in muxed address mode).

Remark MRS: MRS command
RFp: REFRESH Bank $p$
AC : any command

### 2.15 Address Mapping in Multiplexed Mode

The address mapping is described in Table 2-6 as a function of data width and burst length.

Table 2-6. Address Mapping in Multiplexed Address Mode

| Data <br> Width | Burst <br> Length | Ball | Address |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | A0 | A3 | A4 | A5 | A8 | A9 | A10 | A13 | A14 | A17 | A18 |
| x36 | BL=2 | Ax | A0 | A3 | A4 | A5 | A8 | A9 | A10 | A13 | A14 | A17 | A18 |
|  |  | Ay | X | A1 | A2 | X | A6 | A7 | X | A11 | A12 | A16 | A15 |
|  | $B L=4$ | Ax | A0 | A3 | A4 | A5 | A8 | A9 | A10 | A13 | A14 | A17 | X |
|  |  | Ay | X | A1 | A2 | X | A6 | A7 | X | A11 | A12 | A16 | A15 |
| x18 | BL=2 | Ax | A0 | A3 | A4 | A5 | A8 | A9 | A10 | A13 | A14 | A17 | A18 |
|  |  | Ay | X | A1 | A2 | X | A6 | A7 | A19 | A11 | A12 | A16 | A15 |
|  | $B L=4$ | Ax | A0 | A3 | A4 | A5 | A8 | A9 | A10 | A13 | A14 | A17 | A18 |
|  |  | Ay | X | A1 | A2 | X | A6 | A7 | X | A11 | A12 | A16 | A15 |
|  | BL=8 | Ax | A0 | A3 | A4 | A5 | A8 | A9 | A10 | A13 | A14 | A17 | X |
|  |  | Ay | X | A1 | A2 | X | A6 | A7 | X | A11 | A12 | A16 | A15 |
| x9 | BL=2 | Ax | A0 | A3 | A4 | A5 | A8 | A9 | A10 | A13 | A14 | A17 | A18 |
|  |  | Ay | A20 | A1 | A2 | X | A6 | A7 | A19 | A11 | A12 | A16 | A15 |
|  | $B L=4$ | Ax | A0 | A3 | A4 | A5 | A8 | A9 | A10 | A13 | A14 | A17 | A18 |
|  |  | Ay | X | A1 | A2 | X | A6 | A7 | A19 | A11 | A12 | A16 | A15 |
|  | BL=8 | Ax | A0 | A3 | A4 | A5 | A8 | A9 | A10 | A13 | A14 | A17 | A18 |
|  |  | Ay | X | A1 | A2 | X | A6 | A7 | X | A11 | A12 | A16 | A15 |

Remark X means "Don't care".

### 2.16 Read \& Write configuration in Multiplexed Address Mode

In multiplexed address mode, the READ and WRITE latencies are increased by one clock cycle. The $\mu$ PD48288209/18/36-A cycle time remains the same, as described in Table 2-7.

Table 2-7. Configuration in Multiplexed Address Mode

| Frequency | Symbol | Configuration |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $1^{\text {Note }}$ | 2 | 3 |  |
|  | trc | 4 | 6 | 8 | Cycles |
|  | trL | 5 | 7 | 9 | Cycles |
|  | twL | 6 | 8 | 10 | Cycles |
| 400 MHz | trc |  |  | 20.0 | ns |
|  | trL |  |  | 22.5 | ns |
|  | twL |  |  | 25.0 | ns |
| 300 MHz | trc |  | 20.0 | 26.7 | ns |
|  | trL |  | 23.3 | 30.0 | ns |
|  | twL |  | 26.7 | 33.3 | ns |
| 200 MHz | trc | 20.0 | 30.0 | 40.0 | ns |
|  | trL | 25.0 | 35.0 | 45.0 | ns |
|  | twL | 30.0 | 40.0 | 50.0 | ns |

Note BL=8 is not available for configuration 1.

### 2.17 Refresh Command in Multiplexed Address Mode

Similar to other commands, the refresh command is executed on the next rising clock edge when in the multiplexed address mode. However, since only bank address is required for AREF, the next command can be applied on the following clock. The operation of the AREF command and any other command is represented in Figure 2-29.

Figure 2-29. Burst REFRESH Operation


## Remark AREF:AUTO REFRESH

AC : Any command
Ax : First part Ax of address
Ay : Second part Ay of address
BAp : Bank $p$ is chosen so that trc is met.

Figure 2-30. WRITE Burst Basic Sequence: BL=4, with Multiplexed Addresses, Configuration 1


Figure 2-31. READ Burst Basic Sequence: BL=4, with Multiplexed Addresses, Configuration 1, RL=5


Remark WR : WRITE command
RD : READ command
Ax/BAp: Address Ax of bank p
Ay : Address Ay of bank p
Dpq : Data q to bank p
Qpq : Data q from bank p
WL : WRITE latency
RL : READ latency

## 3. JTAG Specification

These products support a limited set of JTAG functions as in IEEE standard 1149.1.

Table 3-1. Test Access Port (TAP) Pins

| Pin name | Pin assignments | Description |
| :--- | :--- | :--- |
| TCK | 12 A | Test Clock Input. All input are captured on the rising edge of TCK and all outputs <br> propagate from the falling edge of TCK. |
| TMS | 11 A | Test Mode Select. This is the command input for the TAP controller state machine. |
| TDI | 12 V | Test Data Input. This is the input side of the serial registers placed between TDI and <br> TDO. The register placed between TDI and TDO is determined by the state of the TAP <br> controller state machine and the instruction that is currently loaded in the TAP instruction. |
| TDO | 11 V | Test Data Output. This is the output side of the serial registers placed between TDI and <br> TDO. Output changes in response to the falling edge of TCK. |

Remark The device does not have TRST (TAP reset). The Test-Logic Reset state is entered while TMS is held HIGH for five rising edges of TCK. The TAP controller state is also reset on the POWER-UP.

Table 3-2. JTAG DC Characteristics ( $0^{\circ} \mathrm{C} \leq \mathrm{Tc} \leq 95^{\circ} \mathrm{C}, 1.7 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD}} \leq 1.9 \mathrm{~V}$, unless otherwise noted)

| Parameter | Symbol | Conditions | MIN. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| JTAG Input leakage current | $\mathrm{l}, \mathrm{I}$ | $0 \mathrm{~V} \leq \mathrm{V}_{\text {IN }} \leq \mathrm{V}_{\text {dD }}$ | -5.0 | +5.0 | $\mu \mathrm{A}$ |
| JTAG I/O leakage current | ILo | $0 \mathrm{~V} \leq \mathrm{V}_{\mathrm{IN}} \leq \mathrm{V}_{\mathrm{DD}} \mathrm{Q},$ <br> Outputs disabled | -5.0 | +5.0 | $\mu \mathrm{A}$ |
| JTAG input HIGH voltage | $\mathrm{V}_{\text {IH }}$ |  | $V_{\text {REF }}+0.15$ | $V_{\text {do }}+0.3$ | V |
| JTAG input LOW voltage | VIL |  | VssQ - 0.3 | $V_{\text {ref }}-0.15$ | V |
| JTAG output HIGH voltage | Voh1 | \| Іонс | $=100 \mu \mathrm{~A}^{\text {a }}$ | VdDQ-0.2 |  | V |
|  | Voh2 | $\mid$ Іонт \| = 2 mA | VDDQ - 0.4 |  | V |
| JTAG output LOW voltage | Vol1 | lolc $=100 \mu \mathrm{~A}$ |  | 0.2 | V |
|  | Vol2 | lolt $=2 \mathrm{~mA}$ |  | 0.4 | V |

Note 1. All voltages referenced to Vss (GND).
2. Overshoot: $\mathrm{VIH}_{(\mathrm{AC}} \leq \mathrm{VDD}^{2}+0.7 \mathrm{~V}$ for $\mathrm{t} \leq \mathrm{tck} / 2$.

Undershoot: VIL (AC) $\geq-0.5 \mathrm{~V}$ for $\mathrm{t} \leq \mathrm{tck} / 2$.
During normal operation, VDDQ must not exceed VDD.

## JTAG AC Test Conditions

Input waveform (Rise / Fall time $\leq 0.3 \mathrm{~ns}$ )


## Output waveform



Output load condition


Table 3-3. JTAG AC Characteristics $\left(0^{\circ} \mathrm{C} \leq \mathrm{Tc} \leq 95^{\circ} \mathrm{C}\right)$

| Parameter | Symbol | Conditions | MIN. | MAX. | Unit | Note |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Clock |  |  |  |  |  |  |
| Clock cycle time | tтhth |  | 20 |  | ns |  |
| Clock frequency | $\mathrm{f}_{\mathrm{TF}}$ |  |  | 50 | MHz |  |
| Clock HIGH time | tthti |  | 10 |  | ns |  |
| Clock LOW time | tтLth |  | 10 |  | ns |  |
| Output time |  |  |  |  |  |  |
| TCK LOW to TDO unknown | ttoox |  | 0 |  | ns |  |
| TCK LOW to TDO valid | ttoov |  |  | 10 | ns |  |
| Setup time |  |  |  |  |  |  |
| TMS setup time | $\mathrm{t}_{\text {mVTH }}$ |  | 5 |  | ns |  |
| TDI valid to TCK HIGH | tovth |  | 5 |  | ns |  |
| Capture setup time | tcs |  | 5 |  | ns | 1 |
| Hold time |  |  |  |  |  |  |
| TMS hold time | tтhmx |  | 5 |  | ns |  |
| TCK HIGH to TDI invalid | tthdx |  | 5 |  | ns |  |
| Capture hold time | tchu |  | 5 |  | ns | 1 |

Note 1. tcss and tchu refer to the setup and hold time requirements of latching data from the boundary scan register.

## JTAG Timing Diagram



Table 3-4. Scan Register Definition (1)

| Register name | Description |
| :--- | :--- |
| Instruction register | The 8 bit instruction registers hold the instructions that are executed by the TAP controller. The <br> register can be loaded when it is placed between the TDI and TDO pins. The instruction register is <br> automatically preloaded with the IDCODE instruction at power-up whenever the controller is placed <br> in test-logic-reset state. |
| Bypass register | The bypass register is a single bit register that can be placed between TDI and TDO. It allows serial <br> test data to be passed through the RAMs TAP to another device in the scan chain with as little delay <br> as possible. The bypass register is set LOW (Vss) when the bypass instruction is executed. |
| ID register | The ID Register is a 32 bit register that is loaded with a device and vendor specific 32 bit code when <br> the controller is put in capture-DR state with the IDCODE command loaded in the instruction register. <br> The register is then placed between the TDI and TDO pins when the controller is moved into shift-DR <br> state. |
| Boundary register | The boundary register, under the control of the TAP controller, is loaded with the contents of the <br> RAMs I/O ring when the controller is in capture-DR state and then is placed between the TDI and <br> TDO pins when the controller is moved to shift-DR state. Several TAP instructions can be used to <br> activate the boundary register. <br> The Scan Exit Order tables describe which device bump connects to each boundary register <br> location. The first column defines the bit's position in the boundary register. The second column is <br> the name of the input or I/O at the bump and the third column is the bump number. |

Table 3-5. Scan Register Definition (2)

| Register name | Bit size | Unit |
| :--- | :---: | :---: |
| Instruction register | 8 | bit |
| Bypass register | 1 | bit |
| ID register | 32 | bit |
| Boundary register | 113 | bit |

Table 3-6. ID Register Definition

| Part number | Organization | ID [31:28] vendor revision no. | ID [27:12] part no. | ID [11:1] vendor ID no. | ID [0] fix bit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mu$ PD48288209-A | $32 \mathrm{M} \times 9$ | 0000 | 0001000010100111 | 00000010000 | 1 |
| $\mu$ PD48288218-A | $16 \mathrm{M} \times 18$ | 0001 | 0001000010100111 | 000000010000 | 1 |
| $\mu \mathrm{PD} 48288236-\mathrm{A}$ | $8 \mathrm{M} \times 36$ | 0010 | 0001000010100111 | 000000010000 | 1 |

Table 3-7. SCAN Exit Order

| Bit no. | Signal name |  |  | Bump ID | Bit no. | Signal name |  |  |  | Bit no. | Signal name |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | x9 | x18 | x36 |  |  | x9 | x18 | x36 |  |  | x9 | x18 | x36 |  |
| 1 | DK | DK | DK1 | K1 | 39 | DNU | DNU | DQ30 | R11 | 77 | DNU | DNU | DQ2 | C11 |
| 2 | DK\# | DK\# | DK1\# | K2 | 40 | DNU | DNU | DQ30 | R11 | 78 | DNU | DNU | DQ2 | C11 |
| 3 | CS\# | CS\# | CS\# | L2 | 41 | DNU | DNU | DQ32 | P11 | 79 | DQ1 | DQ1 | DQ3 | C10 |
| 4 | REF\# | REF\# | REF\# | L1 | 42 | DNU | DNU | DQ32 | P11 | 80 | DQ1 | DQ1 | DQ3 | C10 |
| 5 | WE\# | WE\# | WE\# | M1 | 43 | DQ5 | DQ10 | DQ33 | P10 | 81 | DNU | DNU | DQ0 | B11 |
| 6 | A17 | A17 | A17 | M3 | 44 | DQ5 | DQ10 | DQ33 | P10 | 82 | DNU | DNU | DQ0 | B11 |
| 7 | A16 | A16 | A16 | M2 | 45 | DNU | DNU | DQ34 | N11 | 83 | DQ0 | DQ0 | DQ1 | B10 |
| 8 | A18 | A18 | A18 | N1 | 46 | DNU | DNU | DQ34 | N11 | 84 | DQ0 | DQ0 | DQ1 | B10 |
| 9 | A15 | A15 | A15 | P1 | 47 | DQ4 | DQ9 | DQ35 | N10 | 85 | DNU | DQ4 | DQ9 | B3 |
| 10 | DNU | DQ14 | DQ25 | N3 | 48 | DQ4 | DQ9 | DQ35 | N10 | 86 | DNU | DQ4 | DQ9 | B3 |
| 11 | DNU | DQ14 | DQ25 | N3 | 49 | DM | DM | DM | P12 | 87 | DNU | DNU | DQ8 | B2 |
| 12 | DNU | DNU | DQ24 | N2 | 50 | A19 | A19 | (A19) | N12 | 88 | DNU | DNU | DQ8 | B2 |
| 13 | DNU | DNU | DQ24 | N2 | 51 | A11 | A11 | A11 | M11 | 89 | DNU | DQ5 | DQ11 | C3 |
| 14 | DNU | DQ15 | DQ23 | P3 | 52 | A12 | A12 | A12 | M10 | 90 | DNU | DQ5 | DQ11 | C3 |
| 15 | DNU | DQ15 | DQ23 | P3 | 53 | A10 | A10 | A10 | M12 | 91 | DNU | DNU | DQ10 | C2 |
| 16 | DNU | DNU | DQ22 | P2 | 54 | A13 | A13 | A13 | L12 | 92 | DNU | DNU | DQ10 | C2 |
| 17 | DNU | DNU | DQ22 | P2 | 55 | A14 | A14 | A14 | L11 | 93 | DNU | DQ6 | DQ13 | D3 |
| 18 | DNU | QK1 | QK1 | R2 | 56 | BA1 | BA1 | BA1 | K11 | 94 | DNU | DQ6 | DQ13 | D3 |
| 19 | DNU | QK1\# | QK1\# | R3 | 57 | CK\# | CK\# | CK\# | K12 | 95 | DNU | DNU | DQ12 | D2 |
| 20 | DNU | DNU | DQ20 | T2 | 58 | CK | CK | CK | J12 | 96 | DNU | DNU | DQ12 | D2 |
| 21 | DNU | DNU | DQ20 | T2 | 59 | BA0 | BA0 | BAO | J11 | 97 | DNU | DNU | DQ14 | E2 |
| 22 | DNU | DQ16 | DQ21 | T3 | 60 | A4 | A4 | A4 | H11 | 98 | DNU | DNU | DQ14 | E2 |
| 23 | DNU | DQ16 | DQ21 | T3 | 61 | A3 | A3 | A3 | H12 | 99 | DNU | DQ7 | DQ15 | E3 |
| 24 | DNU | DNU | DQ18 | U2 | 62 | A0 | A0 | A0 | G12 | 100 | DNU | DQ7 | DQ15 | E3 |
| 25 | DNU | DNU | DQ18 | U2 | 63 | A2 | A2 | A2 | G10 | 101 | DNU | DNU | DQ16 | F2 |
| 26 | DNU | DQ17 | DQ19 | U3 | 64 | A1 | A1 | A1 | G11 | 102 | DNU | DNU | DQ16 | F2 |
| 27 | DNU | DQ17 | DQ19 | U3 | 65 | A20 | (A20) | (A20) | E12 | 103 | DNU | DQ8 | DQ17 | F3 |
| 28 | ZQ | ZQ | ZQ | V2 | 66 | QVLD | QVLD | QVLD | F12 | 104 | DNU | DQ8 | DQ17 | F3 |
| 29 | DQ8 | DQ13 | DQ27 | U10 | 67 | DQ3 | DQ3 | DQ7 | F10 | 105 | (A21) | (A21) | (A21) | E1 |
| 30 | DQ8 | DQ13 | DQ27 | U10 | 68 | DQ3 | DQ3 | DQ7 | F10 | 106 | A5 | A5 | A5 | F1 |
| 31 | DNU | DNU | DQ26 | U11 | 69 | DNU | DNU | DQ6 | F11 | 107 | A6 | A6 | A6 | G2 |
| 32 | DNU | DNU | DQ26 | U11 | 70 | DNU | DNU | DQ6 | F11 | 108 | A7 | A7 | A7 | G3 |
| 33 | DQ7 | DQ12 | DQ29 | T10 | 71 | DQ2 | DQ2 | DQ5 | E10 | 109 | A8 | A8 | A8 | G1 |
| 34 | DQ7 | DQ12 | DQ29 | T10 | 72 | DQ2 | DQ2 | DQ5 | E10 | 110 | BA2 | BA2 | BA2 | H1 |
| 35 | DNU | DNU | DQ28 | T11 | 73 | DNU | DNU | DQ4 | E11 | 111 | A9 | A9 | A9 | H2 |
| 36 | DNU | DNU | DQ28 | T11 | 74 | DNU | DNU | DQ4 | E11 | 112 | NF | NF | DKO\# | J2 |
| 37 | DQ6 | DQ11 | DQ31 | R10 | 75 | QK0 | QK0 | QK0 | D11 | 113 | NF | NF | DK0 | J1 |
| 38 | DQ6 | DQ11 | DQ31 | R10 | 76 | QKO\# | QKO\# | QKO\# | D10 |  |  |  |  |  |

Note Any unused balls that are in the order will read as a logic " 0 ".

## JTAG Instructions

Many different instructions $\left(2^{8}\right)$ are possible with the 8 -bit instruction register. All used combinations are listed in Table 3-8, Instruction Codes. These six instructions are described in detail below. The remaining instructions are reserved and should not be used.

The TAP controller used in this RAM is fully compliant to the 1149.1 convention. Instructions are loaded into the TAP controller during the Shift-IR state when the instruction register is placed between TDI and TDO. During this state, instructions are shifted through the instruction register through the TDI and TDO balls. To execute the instruction once it is shifted in, the TAP controller needs to be moved into the Update-IR state.

Table 3-8

| Instructions | Instruction Code [7:0] | Description |
| :---: | :---: | :---: |
| EXTEST | 00000000 | The EXTEST instruction allows circuitry external to the component package to be tested. Boundary-scan register cells at output pins are used to apply test vectors, while those at input pins capture test results. Typically, the first test vector to be applied using the EXTEST instruction will be shifted into the boundary scan register using the PRELOAD instruction. Thus, during the update-IR state of EXTEST, the output drive is turned on and the PRELOAD data is driven onto the output pins. |
| IDCODE | 00100001 | The IDCODE instruction causes the ID ROM to be loaded into the ID register when the controller is in capture-DR mode and places the ID register between the TDI and TDO pins in shift-DR mode. The IDCODE instruction is the default instruction loaded in at power up and any time the controller is placed in the test-logic-reset state. |
| SAMPLE / PRELOAD | 00000101 | SAMPLE / PRELOAD is a Standard 1149.1 mandatory public instruction. When the SAMPLE / PRELOAD instruction is loaded in the instruction register, moving the TAP controller into the capture-DR state loads the data in the RAMs input and DQ pins into the boundary scan register. Because the RAM clock(s) are independent from the TAP clock (TCK) it is possible for the TAP to attempt to capture the I/O ring contents while the input buffers are in transition (i.e., in a metastable state). Although allowing the TAP to sample metastable input will not harm the device, repeatable results cannot be expected. RAM input signals must be stabilized for long enough to meet the TAPs input data capture setup plus hold time (tcs plus tch). The RAMs clock inputs need not be paused for any other TAP operation except capturing the I/O ring contents into the boundary scan register. Moving the controller to shift-DR state then places the boundary scan register between the TDI and TDO pins. |
| CLAMP | 00000111 | When the CLAMP instruction is loaded into the instruction register, the data driven by the output balls are determined from the values held in the boundary scan register. Selects the bypass register to be connected between TDI and TDO. Data driven by output balls are determined from values held in the boundary scan register. |
| High-Z | 00000011 | The High-z instruction causes the boundary scan register to be connected between the TDI and TDO. This places all RAMs outputs into a High-Z state. <br> Selects the bypass register to be connected between TDI and TDO. All outputs are forced into high impedance state. |
| BYPASS | 11111111 | When the BYPASS instruction is loaded in the instruction register, the bypass register is placed between TDI and TDO. This occurs when the TAP controller is moved to the shift-DR state. This allows the board level scan path to be shortened to facilitate testing of other devices in the scan path. |
| Reserved for Future Use | - | The remaining instructions are not implemented but are reserved for future use. Do not use these instructions. |

## TAP Controller State Diagram



## 4. Package Drawing

144-PIN TAPE FBGA ( $\mu \mathrm{BGA}$ ) $(18.5 \times 11)$


Detail of Apart


|  | (UNIT:mm) |
| :---: | :--- |
| ITEM | DIMENSIONS |
| D | $18.50 \pm 0.10$ |
| D1 | 17.90 |
| D2 | 14.52 |
| $E$ | $11.00 \pm 0.10$ |
| E1 | 10.70 |
| E2 | 2.184 |
| $w$ | 0.20 |
| $A$ | $1.07 \pm 0.10$ |
| A1 | $0.39 \pm 0.05$ |
| A2 | 0.68 |
| A3 | 0.08 MAX. |
| eD | 1.00 |
| eE | 0.80 |
| SD | 0.50 |
| SE | 2.00 |
| $b$ | $0.51 \pm 0.05$ |
| $x$ | 0.15 |
| $y$ | 0.10 |
| $y 1$ | 0.20 |
| ZD | 0.75 |
| ZE | 1.10 |
|  | P144FF-80-DW1 |

## 5. Recommended Soldering Condition

Please consult with our sales offices for soldering conditions of these products.

## Types of Surface Mount Devices

$\mu$ PD48288209FF-DW1-A : 144-pin TAPE FBGA ( $18.5 \times 11$ )
$\mu$ PD48288218FF-DW1-A : 144-pin TAPE FBGA ( $18.5 \times 11$ )
$\mu$ PD48288236FF-DW1-A : 144-pin TAPE FBGA ( $18.5 \times 11$ )

## Quality Grade

- A quality grade of the products is "Standard".
- Anti-radioactive design is not implemented in the products.
- Semiconductor devices have the possibility of unexpected defects by affection of cosmic ray that reach to the ground and so forth.

| Revision History   $\mu$ PD48288209-A, $\mu$ PD48288218-A, $\mu$ PD48288236-A <br> Rev. Date   <br>  Page Description  <br> Rev.1.00 '13.02.01 - New Data Sheet |
| :--- |

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