

LM4985 Boomer[™] Audio Power Amplifier Series Stereo 135mW Low Noise Headphone Amplifier with Selectable Capacitively Coupled or Output Capacitor-less (OCL) Output and Digitally Controlled (I²C) Volume Control

Check for Samples: LM4985, LM4985TMEVAL

FEATURES

- **OCL or Capacitively Coupled Outputs (Patent** Pending)
- I²C Digitally Controlled Volume Control
- Available in Space-Saving 0.4mm Lead-Pitch **DSBGA** Package
- Volume Control Range: -76dB to +18dB
- **Ultra Low Current Shutdown Mode**
- 2.3V 5.5V Operation
- **Ultra Low Noise**

APPLICATIONS

- **Mobile Phones**
- **PDAs**
- **Portable Electronics Devices**
- **MP3 Players**

KEY SPECIFICATIONS ($V_{DD} = 3.6V$)

- PSRR: 217Hz and 1kHzs
 - Output Capacitor-Less (OCL)
 - $f_{RIPPLE} = 217Hz, 77dB (Typ)$
 - $f_{RIPPLE} = 1 kHz, 76 dB (Typ)$
 - Capacitor Coupled (C-CUPL)
 - $f_{RIPPLE} = 217Hz, 63dB (Typ)$
 - $f_{RIPPLE} = 1 kHz$, 62dB (Typ)
 - Output Power Per Channel
 - $(f_{IN} = 1 kHz, THD+N = 1\%),$

 $R_L = 16\Omega$, OCL

- $-V_{DD} = 2.5V, 31mW (Typ)$
- V_{DD} = 3.6V, 68mW (Typ)
- $-V_{DD} = 5.0V, 135mW (Typ)$
- THD+N (f = 1kHz)
 - $-R_{LOAD} = 16\Omega$, OCL, $P_{OUT} = 60$ mW, 0.60
 - $-R_{LOAD} = 32\Omega$, OCL, $P_{OUT} = 33$ mW, 0.031
- Shutdown Current, 0.1µA (Typ)

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DESCRIPTION

The LM4985 is a stereo audio power amplifier with internal digitally controlled volume control. This amplifier is capable of delivering 68mW_{RMS} per channel into a 16Ω load or $38mW_{RMS}$ per channel into a 32 Ω load at 1% THD when powered by a 3.6V power supply and operating in the OCL mode.

Boomer audio power amplifiers were designed specifically to provide high quality output power with a minimal amount of external components. To that end, the LM4985 features two functions that optimize system cost and minimize PCB area: an integrated, digitally controlled (I²C bus) volume control and an operational mode that eliminates output signal coupling capacitors (OCL mode). Since the LM4985 does not require bootstrap capacitors, snubber networks, or output coupling capacitors, it is optimally suited for low-power, battery powered portable systems. For added design flexibility, the LM4985 can also be configured for single-ended capacitively coupled outputs.

The LM4985 features a current shutdown mode for micropower dissipation and thermal shutdown protection.

Block Diagram

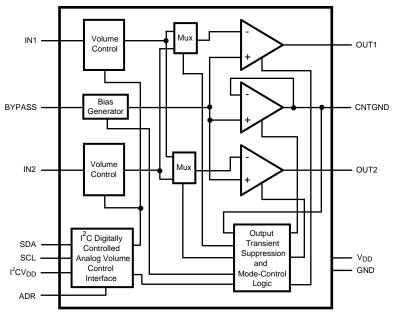


Figure 1. Block Diagram

Typical Application

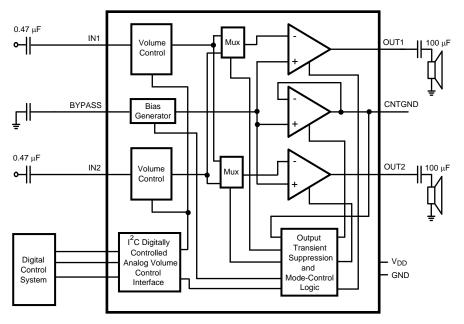


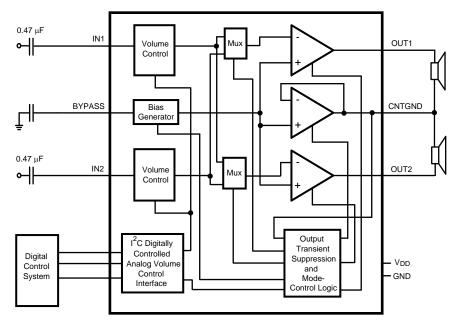
Figure 2. Typical Capacitively Coupled Output Configuration Circuit



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Connection Diagram

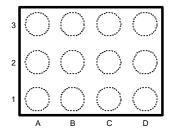


Figure 4. DSBGA Package Top View See NS Package Number YFQ0012

PIN REFERENCE, NAME, AND FUNCTION

Reference	Name	Function			
A1	ADR	I ² C serial interface address input.			
A2	IN2	Analog signal input two.			
A3	OUT2	Power amplifier two output.			
B1	SDA	I ² C serial interface data input.			
B2	BYPASS	The internal $V_{DD}/2$ ac bypass node.			
B3	CNTGND	In OCL mode, this is the ac ground return. It is biased to $V_{\mbox{\scriptsize DD}}/2.$ Leave unconnected for C-CUPL mode.			
C1	SCL	I2C serial interface clock input.			
C2	GND	The LM4985's power supply ground input.			
C3	V _{DD}	The LM4985's power supply voltage input.			
D1	I ² CV _{DD}	$\rm I^2C$ serial interface power supply input. Can be connected to the same supply that is connected to the $\rm V_{DD}$ pin.			
D2	IN1	Analog signal input one.			
D3	OUT1	Power amplifier one output.			

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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

Absolute Maximum Ratings⁽¹⁾⁽²⁾

6.0V
−65°C to +150°C
-0.3V to V _{DD} + 0.3V
-0.3V to I ² CV _{DD} + 0.3V
Internally Limited
2000V
200V
150°C
109°C/W

(1) All voltages are measured with respect to the GND pin unless otherwise specified.

(2) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional but do not ensure specific performance limits. Electrical Characteristics state DC and AC electrical specifications under particular test conditions which ensure specific performance limits. This assumes that the device is within the Operating Ratings. Specifications are not ensured for parameters where no limit is given, however, the typical value is a good indication of device performance.

(3) The maximum power dissipation must be derated at elevated temperatures and is dictated by TJ_{MAX}, θ_{JA}, and the ambient temperature, T_A. The maximum allowable power dissipation is P_{DMAX} = (T_{JMAX} - T_A)/ θ_{JA} or the number given in Absolute Maximum Ratings, whichever is lower. For the LM4985, see power derating currents for more information.

(4) Human Body Model: 100pF discharged through a $1.5k\Omega$ resistor.

(5) Machine Model: $200pF \le C_{mm} \le 220pF$ discharged through all pins.

Operating Ratings

Temperature Range	$T_{MIN} \le T_A \le T_{MAX}$	−40°C ≤ T _A ≤ 85°C
Currente Matterna	V _{DD}	$2.3 V \leq V_{CC} \leq 5.5 V$
Supply Voltage	I ² CV _{DD}	$1.7V \le I^2 CV_{DD} \le 5.5V$

Electrical Characteristics $V_{DD} = 5V^{(1)(2)}$

The following specifications apply for $R_L = 16\Omega$, f = 1kHz, and $C_B = 4.7\mu$ F unless otherwise specified. Limits apply to $T_A = 25^{\circ}$ C.

Symbol	Parameter	Conditions	LN	Units		
			Typ ⁽³⁾ Limit ⁽⁴⁾⁽⁵⁾		(Limits)	
I _{DD}	Quiescent Power Supply Current	$V_{IN} = 0V, I_{OUT} = 0A$ Single-Channel no load OCL Single-Channel no load C-CUPL Dual-Channel no load OCL Dual-Channel no load C-CUPL	gle-Channel no load OCL2gle-Channel no load C-CUPL1.5al-Channel no load OCL3		mA (max)	
I _{SD}	Shutdown Current	V _{SHUTDOWN} = GND	0.1	1.0	µA (max)	
V _{SDIH}	Logic Voltage Input High			3.5	V (min)	
V _{SDIL}	Logic Voltage Input Low			1.5	V (max)	

(1) All voltages are measured with respect to the GND pin unless otherwise specified.

(2) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional but do not ensure specific performance limits. Electrical Characteristics state DC and AC electrical specifications under particular test conditions which ensure specific performance limits. This assumes that the device is within the Operating Ratings. Specifications are not ensured for parameters where no limit is given, however, the typical value is a good indication of device performance.

(3) Typicals are measured at 25°C and represent the parametric norm.

- (4) Limits are specified to Texas Instruments' AOQL (Average Outgoing Quality Level).
- (5) Datasheet min/max specification limits are specified by design, test, or statistical analysis.



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Electrical Characteristics $V_{DD} = 5V^{(1)(2)}$ (continued)

25°C.

Symbol	Parameter	Conditions	LN	Units		
			Typ ⁽³⁾ Limit ⁽⁴⁾⁽⁵⁾		(Limits)	
		THD \leq 1%; f _{IN} = 1kHz				
		$R_{LOAD} = 16\Omega \text{ OCL}$	135	115		
Po	Output Power	$R_{LOAD} = 16\Omega \text{ C-CUPL}$	135		mW (min)	
		$R_{LOAD} = 32\Omega \text{ OCL}$	79	70		
		$R_{LOAD} = 32\Omega \text{ C-CUPL}$	80			
THD+N	Total Harmonic Distortion + Noise	$ \begin{array}{l} R_{LOAD} = 16\Omega \ OCL, \ P_{O} = 100 mW \\ R_{LOAD} = 16\Omega \ C\text{-}CUPL, \ P_{O} = 100 mW \\ R_{LOAD} = 32\Omega \ OCL, \ P_{O} = 60 mW \\ R_{LOAD} = 32\Omega \ C\text{-}CUPL, \ P_{O} = 70 mW \end{array} $	0.08 0.02 0.04 0.01		%	
V _{ON}	Output Noise Voltage	15		μV		
PSRR	Power Supply Rejection Ratio	$ \begin{array}{l} V_{\text{RIPPLE}} = 200 \text{mVp-p}^{(6)} \\ f_{\text{IN}} = 217 \text{Hz sinewave} \\ \text{OCL} \\ \text{C-CUPL} \end{array} $	77 65	57	dB (min)	
		f _{IN} = 1kHz sinewave OCL C-CUPL	77 60 65			
Xtalk	Channel-to-channel Crosstalk	$\begin{array}{l} P_{out} = 40 mW. \ OCL \\ R_{LOAD} = 16 \Omega \\ R_{LOAD} = 32 \Omega \end{array}$	51 56		dB	
Λιαικ		$\begin{array}{l} P_{out} = 50 mW. C-CUPL \\ R_{LOAD} = 16 \Omega \\ R_{LOAD} = 32 \Omega \end{array}$	58 68		dB	
		$C_{BYPASS} = 4.7 \mu F^{(7)}$				
		WT1 = 0, WT0 = 0 OCL C-CUPL	75 285			
T _{WU}	Wake Up Time form Shutdown	WT1 = 0, WT0 = 1 OCL C-CUPL	110 530			
		WT1 = 1, WT0 = 0 OCL C-CUPL	180 1030		msec	
		WT1 = 1, WT0 = 1 OCL C-CUPL	320 2050			
R _{IN}	Input Resistance	Stereo mode Mono mode	20 10		kΩ	
A _{VMIN}	Minimum Gain	Code = 00000	-76		dB (min)	
A _{VMAX}	Maximum Gain	Code = 11111	18		dB (min)	
ΔA _V	Gain Accuracy per Step	$18dB \ge A_V \ge -44dB$ -44dB \ge A_V > -76dB	±0.5 ±1.0		dB	
V _{OS}	Output Offset Voltage	$\begin{array}{l} \text{OCL} \\ \text{R}_{\text{LOAD}} = 32\Omega \\ \text{V}_{\text{IN}} = \text{AC GND} \end{array}$	2.0	20	mV (max)	

(6) 10Ω terminated input. (7) The wake-up time (T_{WU}) is calculated using the following formula; T_{WU} = [C_{BYPASS} (VDD) / 2 (I_{BYPASS})] + 40ms.

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Electrical Characteristics $V_{DD} = 3.6V^{(1)(2)}$

The following specifications apply for $R_L = 16\Omega$, f = 1kHz, and $C_B = 4.7\mu F$ unless otherwise specified. Limits apply to $T_A = 25^{\circ}C$.

Symbol	Parameter	Conditions	LM	Units			
			Тур ⁽³⁾	Limit ⁽⁴⁾⁽⁵⁾	(Limits)		
		$V_{IN} = 0V, I_{OUT} = 0A$					
		Single-Channel no load OCL	1.8	3.1			
loo	Quiescent Power Supply Current	Single-Channel no load C-CUPL	1.0		mA (max)		
		Dual-Channel no load OCL	2.1	4	mA (max)		
		Dual-Channel no load C-CUPL	2.3	3			
I _{SD}	Shutdown Current	V _{SHUTDOWN} = GND	0.1	1.0	μA (max)		
V _{SDIH}	Logic Voltage Input High			2.52	V (min)		
V _{SDIL}	Logic Voltage Input Low			1.08	V (max)		
		THD+N < 1%, $f_{IN} = 1kHz$					
Po		$R_{LOAD} = 16\Omega \text{ OCL}$	68	60			
	Output Power	$R_{LOAD} = 16\Omega \text{ C-CUPL}$	70		mW (min)		
		$R_{LOAD} = 32\Omega \text{ OCL}$	38	34			
		$R_{LOAD} = 32\Omega \text{ C-CUPL}$	41				
THD+N	Total Harmonic Distortion + Noise		0.06 0.03 0.03 0.03		%		
V _{ON}	Output Noise Voltage	$V_{IN} = AC GND, A_V = 0dB, A$ -weighted	15		μV		
PSRR	Power Supply Rejection Ratio	$V_{RIPPLE} = 200 \text{mVp-p}^{(6)}$ $f_{IN} = 217 \text{Hz}$ sinewave OCL C-CUPL	77 63	55	dB (min)		
		f _{IN} = 1kHz sinewave OCL C-CUPL	76 62	57	、 ,		
Vtolle	Channel-to-Channel Crosstalk	$P_{out} = 40mW. OCL$ $R_{LOAD} = 16\Omega$ $R_{LOAD} = 32\Omega$	51 56		dB		
Xtalk	Channel-to-Channel Crossfalk	$\begin{array}{l} P_{out} = 50 mW. \ C\text{-}CUPL \\ R_{LOAD} = 16 \Omega \\ R_{LOAD} = 32 \Omega \end{array}$	58 69		dB		

(1) All voltages are measured with respect to the GND pin unless otherwise specified.

- (2) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional but do not ensure specific performance limits. Electrical Characteristics state DC and AC electrical specifications under particular test conditions which ensure specific performance limits. This assumes that the device is within the Operating Ratings. Specifications are not ensured for parameters where no limit is given, however, the typical value is a good indication of device performance.
- (3) Typicals are measured at 25°C and represent the parametric norm.
- (4) Limits are specified to Texas Instruments' AOQL (Average Outgoing Quality Level).
- (5) Datasheet min/max specification limits are specified by design, test, or statistical analysis.
- (6) 10Ω terminated input.



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Electrical Characteristics $V_{DD} = 3.6V^{(1)(2)}$ (continued)

The following specifications apply for $R_L = 16\Omega$, f = 1kHz, and $C_B = 4.7\mu$ F unless otherwise specified. Limits apply to $T_A = 25^{\circ}$ C.

Symbol	Parameter	Conditions	LM4	LM4985		
			Тур ⁽³⁾	Limit ⁽⁴⁾⁽⁵⁾	(Limits)	
		$C_{BYPASS} = 4.7 \mu F^{(7)}$				
		WT1 = 0, WT0 = 0 OCL C-CUPL	66 222	93		
T _{WU}	Wake Up Time from Shutdown	WT1 = 0, WT0 = 1 OCL C-CUPL	92 405			
		WT1 = 1, WT0 = 0 OCL C-CUPL	143 774		msec	
		WT1 = 1, WT0 =1 OCL C-CUPL	246 1532			
R _{IN}	Input Resistance	Stereo mode Mono mode	20 10		kΩ	
A _{VMIN}	Minimum Gain	Code = 00000	-76	-72	dB (max)	
A _{VMAX}	Maximum Gain	Code = 11111	18	17	dB (min)	
ΔA _V	Gain Accuracy per Step	$18dB \ge A_V \ge -44dB$ $-44dB \ge A_V > -76dB$	± 0.5 ± 1.0	± 1.0 ± 2.0	dB	
V _{OS}	Output Offset Voltage	$\begin{array}{l} \text{OCL} \\ \text{R}_{\text{LOAD}} = 32\Omega \\ \text{V}_{\text{IN}} = \text{AC GND} \end{array}$	2.0	20	mV (max)	

(7) The wake-up time (T_{WU}) is calculated using the following formula; $T_{WU} = [C_{BYPASS} (VDD) / 2 (I_{BYPASS})] + 40ms$.

Electrical Characteristics V_{DD} = 2.5V⁽¹⁾⁽²⁾

The following specifications apply for $R_L = 16\Omega$, f = 1kHz, and $C_B = 4.7\mu F$ unless otherwise specified. Limits apply to $T_A = 25^{\circ}C$.

Symbol	Parameter	Conditions	LN	Units	
			Тур ⁽³⁾	Limit ⁽⁴⁾⁽⁵⁾	(Limits)
I _{DD}	Quiescent Power Supply Current	V _{IN} = 0V, I _{OUT} = 0A Single-Channel no load OCL Single-Channel no load C-CUPL Dual-Channel no load OCL Dual-Channel no load C-CUPL	1.6 1 2.1 1.6		mA
I _{SD}	Shutdown Current	V _{SHUTDOWN} = GND	0.1		μA
V _{SDIH}	Logic Voltage Input High			1.75	V (min)
V _{SDIL}	Logic Voltage Input Low			0.75	V (max)
Po	Output Power	$\begin{array}{l} THD\text{+}N < 1\%, \ f_{\text{IN}} = 1\text{kHz} \\ R_{\text{LOAD}} = 16\Omega \ \text{OCL} \\ R_{\text{LOAD}} = 16\Omega \ \text{C-CUPL} \\ R_{\text{LOAD}} = 32\Omega \ \text{OCL} \\ R_{\text{LOAD}} = 32\Omega \ \text{C-CUPL} \end{array}$	31 33 19 19		mW

(1) All voltages are measured with respect to the GND pin unless otherwise specified.

(2) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is functional but do not ensure specific performance limits. Electrical Characteristics state DC and AC electrical specifications under particular test conditions which ensure specific performance limits. This assumes that the device is within the Operating Ratings. Specifications are not ensured for parameters where no limit is given, however, the typical value is a good indication of device performance.

- (3) Typicals are measured at 25°C and represent the parametric norm.
- (4) Limits are specified to Texas Instruments' AOQL (Average Outgoing Quality Level).
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Electrical Characteristics $V_{DD} = 2.5V^{(1)(2)}$ (continued)

The following specifications apply for $R_L = 16\Omega$, f = 1kHz, and $C_B = 4.7\mu F$ unless otherwise specified. Limits apply to $T_A = 25^{\circ}C$.

Symbol	Parameter	Conditions	LM4985		Units (Limits)	
			Тур ⁽³⁾	Typ ⁽³⁾ Limit ⁽⁴⁾⁽⁵⁾		
THD+N	Total Harmonic Distortion + Noise		0.07 0.05 0.06 0.04		%	
V _{ON}	Output Noise Voltage	$V_{IN} = AC GND, A_V = 0dB, A$ -weighted	10		μV	
PSRR	Power Supply Rejection Ratio	V _{RIPPLE} = 200mVp-p ⁽⁶⁾ f _{IN} = 217Hz sinewave OCL C-CUPL	75 59		dB	
		f _{IN} = 1kHz sinewave OCL C-CUPL	75 59			
Vtolk	Channel-to-Channel Crosstalk	$\begin{array}{l} P_{out} = 20mW, OCL \\ R_{LOAD} = 16\Omega \\ R_{LOAD} = 32\Omega \end{array}$	50 55		dB	
Xtalk	Channel-to-Channel Crosstaik	$\begin{array}{l} P_{out} = 20mW. \ C\text{-}CUPL \\ R_{LOAD} = 16\Omega \\ R_{LOAD} = 32\Omega \end{array}$	58 67		dB	
		$C_{BYPASS} = 4.7 \mu F^{(7)}$				
		WT1 = 0, WT0 = 0 OCL C-CUPL	66 214			
T _{WU}	Wake Up Time from Shutdown	WT1 = 0, WT0 = 1 OCL C-CUPL	92 544		- msec	
		WT1 = 1, WT0 = 0 OCL C-CUPL	145 1053			
		WT1 = 1, WT0 = 1 OCL C-CUPL	250 2098			
R _{IN}	Input Resistance	Stereo mode Mono mode	20 10		kΩ	
A _{VMIN}	Minimum Gain	Code = 00000	-76		dB	
A _{VMAX}	Maximum Gain	Code = 11111	18		dB	
ΔA _V	Gain Accuracy per Step	$18dB \ge A_V \ge -44dB$ $-44dB \ge A_V > -76dB$	± 0.5 ± 1.0		dB	
V _{OS}	Output Offset Voltage	$\begin{array}{l} \text{OCL} \\ \text{R}_{\text{LOAD}} = 32\Omega \\ \text{V}_{\text{IN}} = \text{AC GND} \end{array}$	2.0		mV	

(6) 10Ω terminated input.

(7) The wake-up time (T_{WU}) is calculated using the following formula; $T_{WU} = [C_{BYPASS} (VDD) / 2 (I_{BYPASS})] + 40$ ms.

External Components Description

(See Figure 2)

Comp	onents	Functional Description
1.	CI	Input coupling capacitor which blocks the DC voltage at the amplifier's input terminals. Also creates a high-pass filter with R_i at $f_c = 1/(2\pi R_i C_i)$. Refer to the section Proper Selection of External Components , for an explanation of how to determine the value of C_i .
2.	C _S	Supply bypass capacitor which provides power supply filtering. Refer to the POWER SUPPLY BYPASSING section for information concerning proper placement and selection of the supply bypass capacitor.

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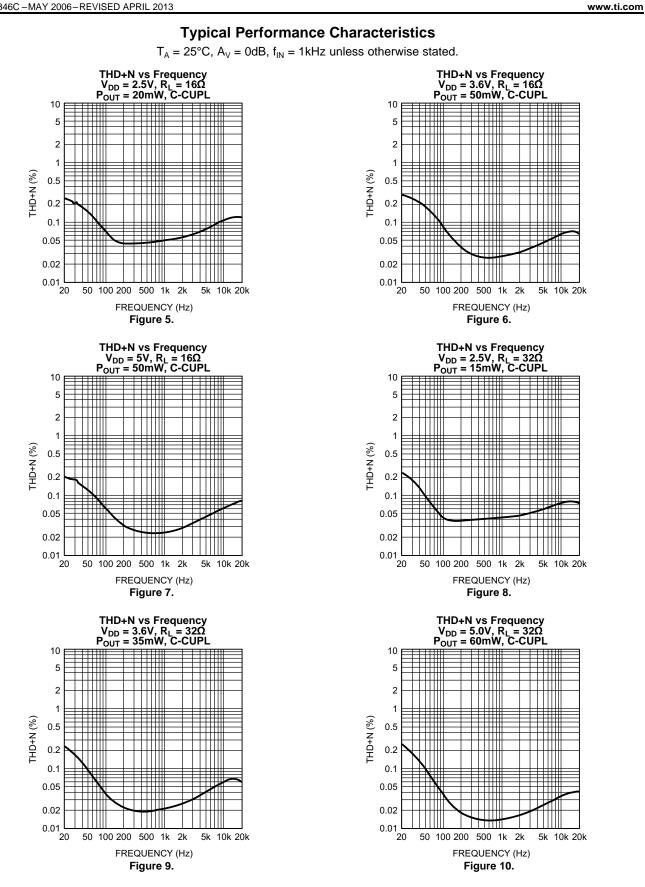
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Comp	onents	Functional Description
3.	C _B	Bypass pin capacitor which provides half-supply filtering. Refer to the section, POWER SUPPLY BYPASSING, for information concerning proper placement and selection of C_B
6.	Co	Output coupling capacitor which blocks the DC voltage at the amplifier's output. Forms a high pass filter with R_L at $f_o = 1/(2\pi R_L C_o)$





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5k 10k 20k

5k 10k 20k

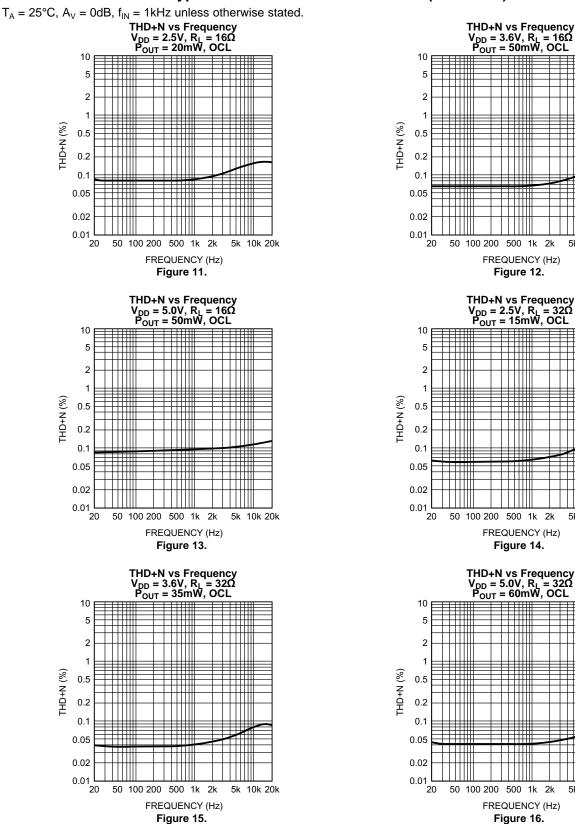
2k

5k 10k 20k

500 1k 2k

500 1k

2k

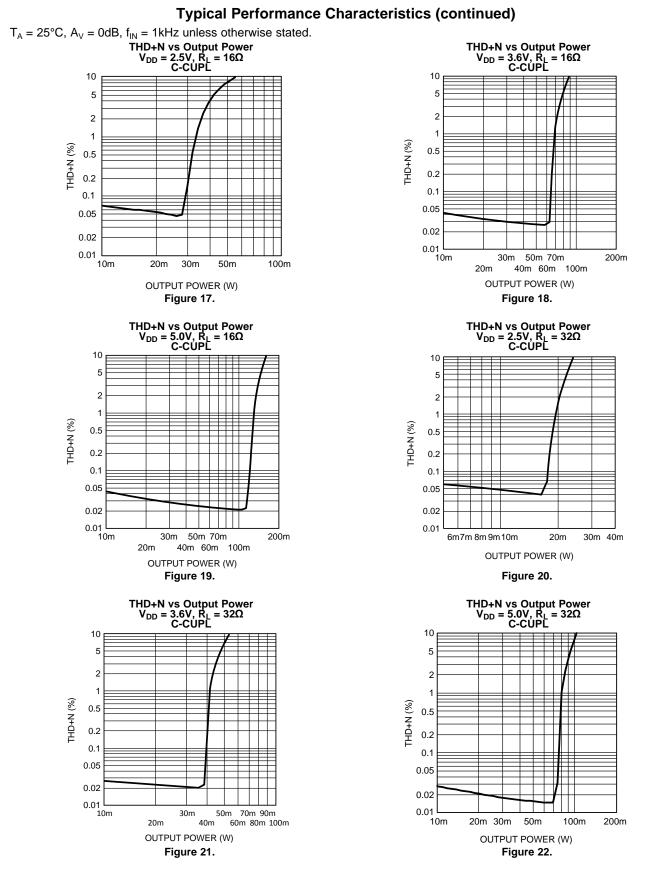


Typical Performance Characteristics (continued)

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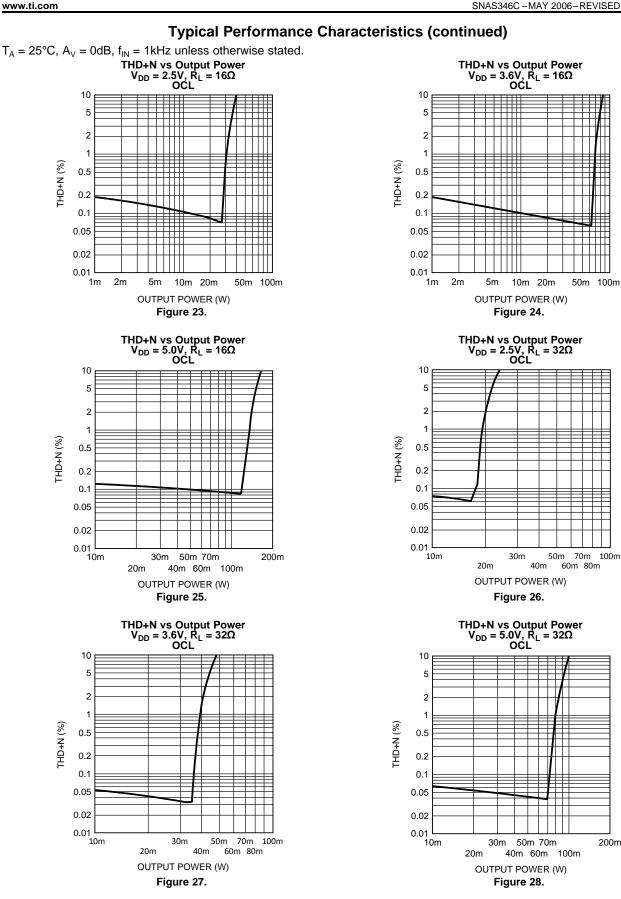
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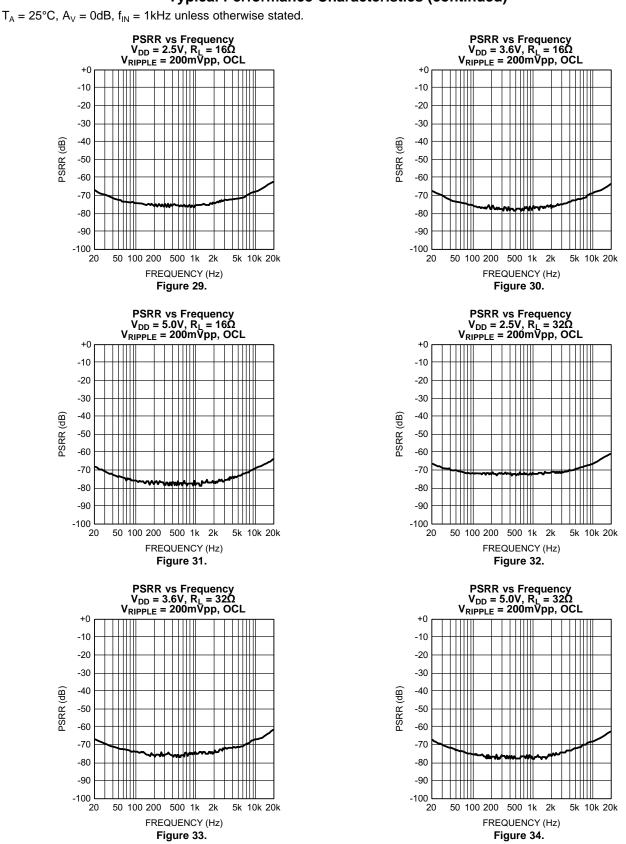
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Typical Performance Characteristics (continued)

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2k

Figure 36.

500 1k

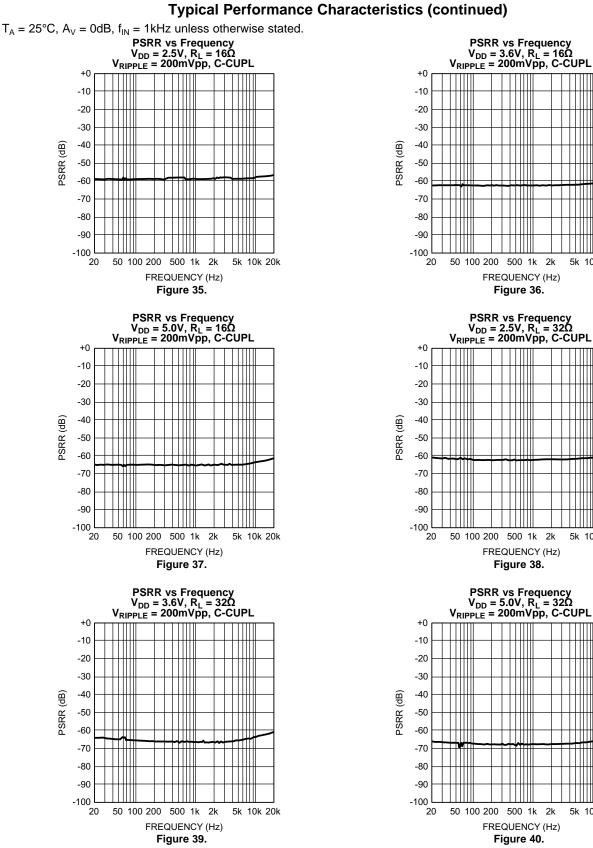
Figure 38.

Figure 40.

2k

5k 10k 20k

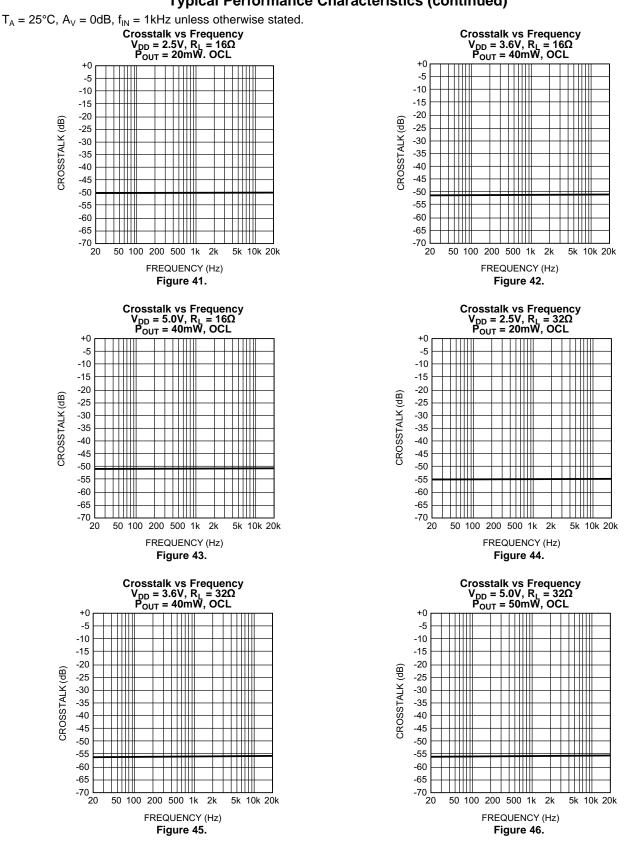
5k 10k 20k



5k 10k 20k

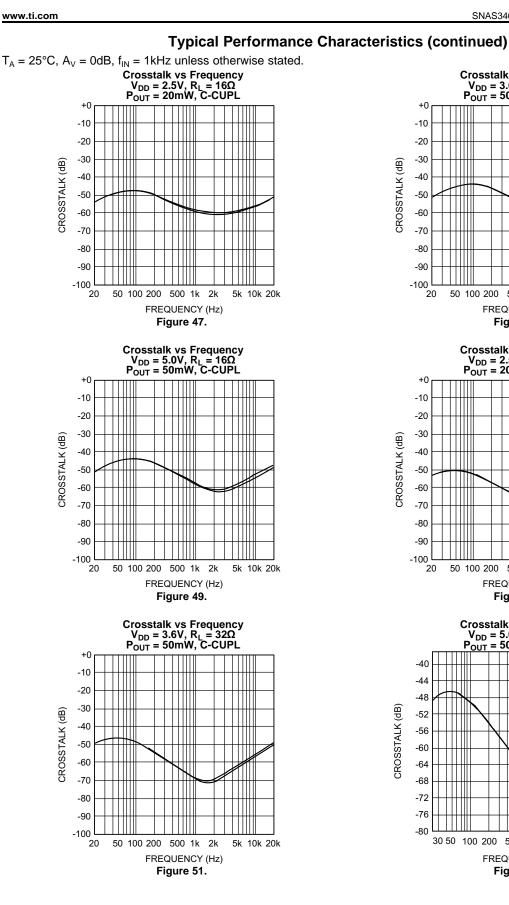


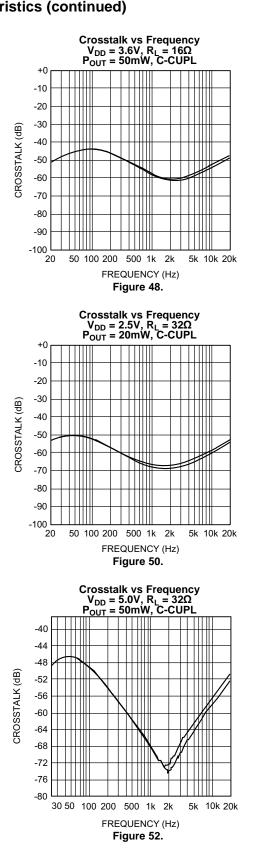
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Typical Performance Characteristics (continued)

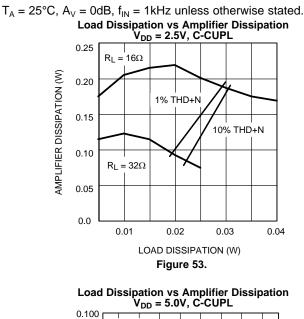


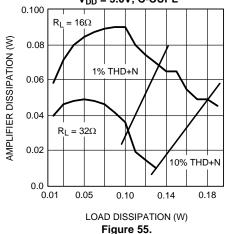




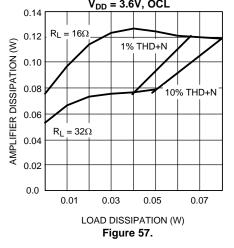


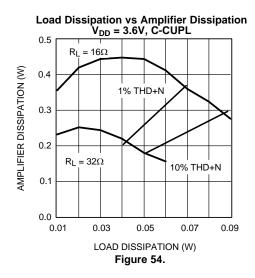
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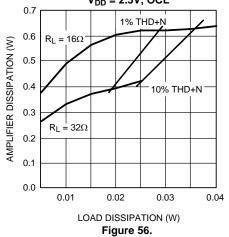


Load Dissipation vs Amplifier Dissipation $V_{DD} = 3.6V, OCL$

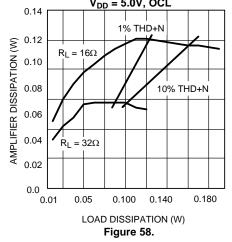




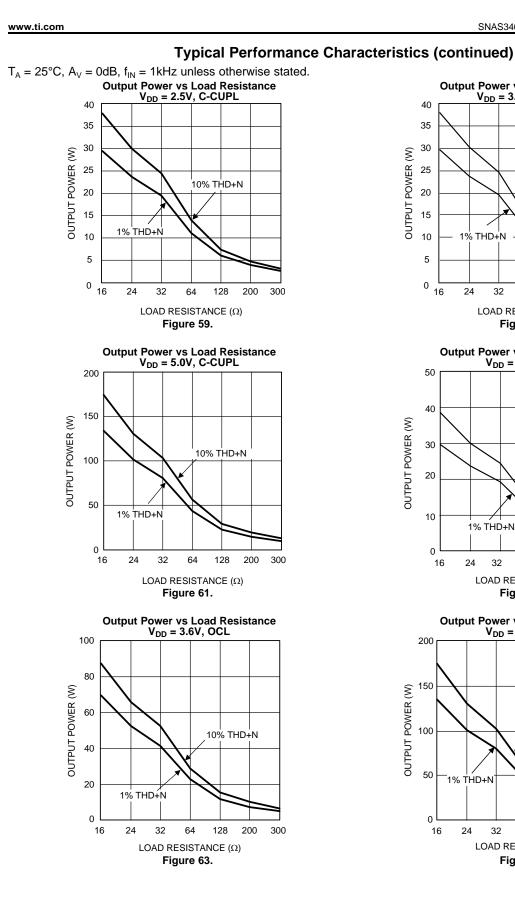
Load Dissipation vs Amplifier Dissipation $V_{DD} = 2.5V$, OCL



Load Dissipation vs Amplifier Dissipation $V_{DD} = 5.0V, OCL$

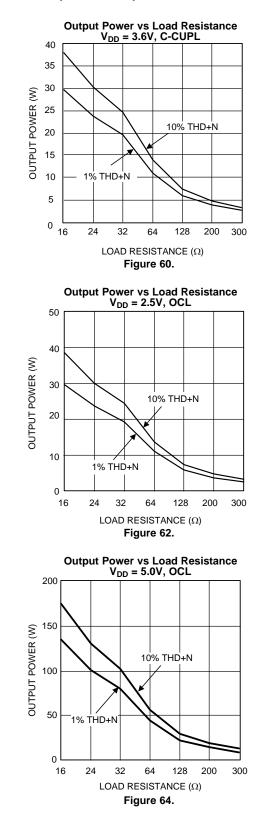


Typical Performance Characteristics (continued)



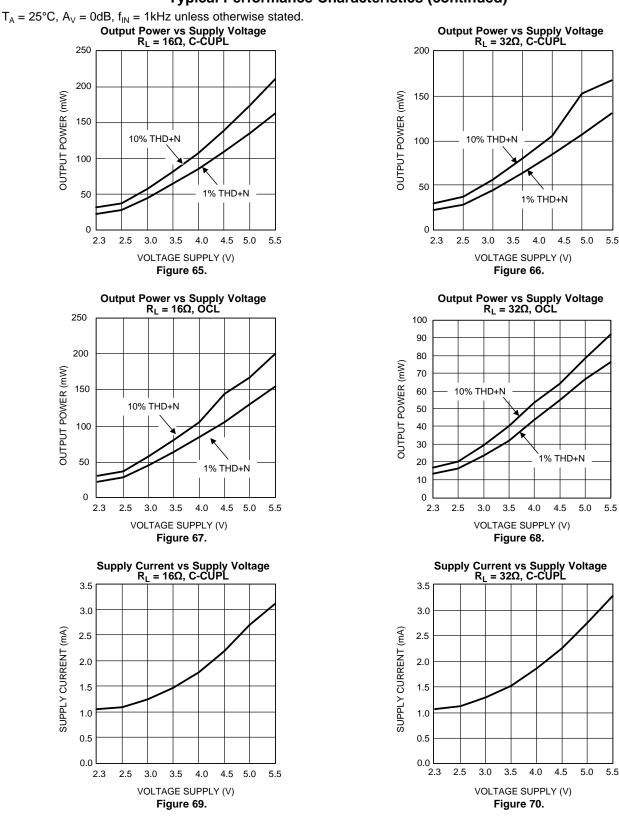
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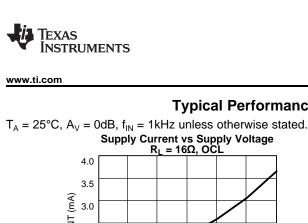
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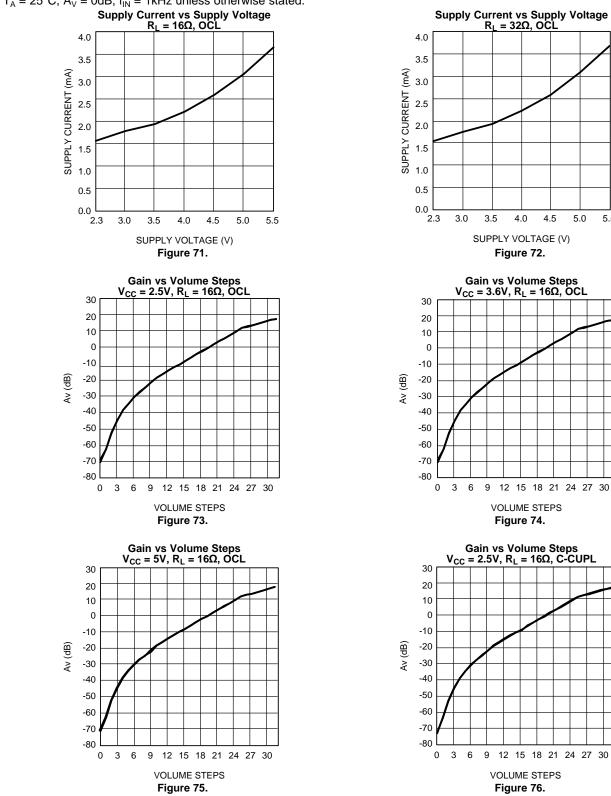
Typical Performance Characteristics (continued)

5.5

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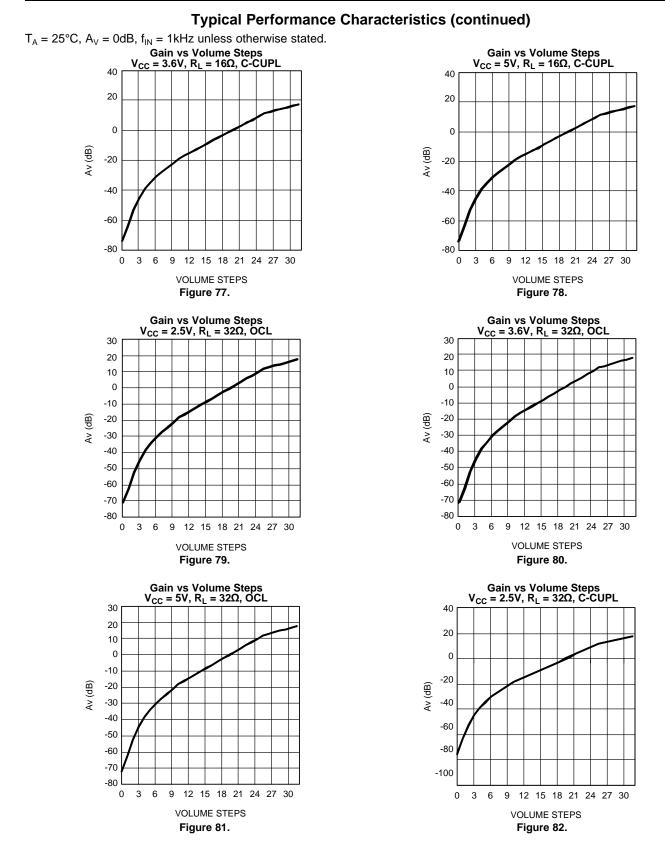


Typical Performance Characteristics (continued)





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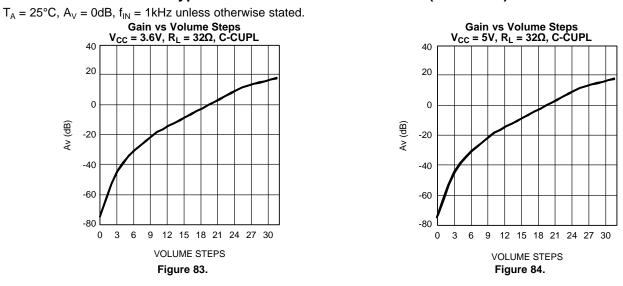


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Typical Performance Characteristics (continued)



APPLICATION INFORMATION

AMPLIFIER CONFIGURATION EXPLANATION

As shown in Figure 1, the LM4985 has three internal power amplifiers. Two of the amplifiers which amplify signals applied to their inputs, have internally configurable gain. The remaining third amplifier provides both half-supply output bias and AC ground return.

Loads, such as a headphone speaker, are connected between OUT1 and CNTGND or OUT2 and CNTGND. This configuration does not require an output coupling capacitor. The classical single-ended amplifier configuration, where one side of the load is connected to ground, requires large, expensive output coupling capacitors.

A configuration such as the one used in the LM4985 has a major advantage over single supply, single-ended amplifiers. Since the outputs OUT1, OUT2, and CNTGND are all biased at $1/2 V_{DD}$, no net DC voltage exists across each load. This eliminates the need for output coupling capacitors which are required in a single-supply, single-ended amplifier configuration. Without output coupling capacitors in a typical single-supply, single-ended amplifier, the bias voltage is placed across the load resulting in both increased internal IC power dissipation and possible loudspeaker damage.

The LM4985 eliminates these output coupling capacitors when operating in Output Capacitor-less (OCL) mode. Unless shorted to ground, VoC is internally configured to apply a $1/2 V_{DD}$ bias voltage to a stereo headphone jack's sleeve. This voltage matches the bias voltage present on VoA and VoB outputs that drive the headphones. The headphones operate in a manner similar to a bridge-tied load (BTL). Because the same DC voltage is applied to both headphone speaker terminals this results in no net DC current flow through the speaker. AC current flows through a headphone speaker as an audio signal's output amplitude increases on the speaker's terminal.

The headphone jack's sleeve is not connected to circuit ground when used in OCL mode. Using the headphone output jack as a line-level output will place the LM4985's $1/2 V_{DD}$ bias voltage on a plug's sleeve connection. This presents no difficulty when the external equipment uses capacitively coupled inputs. For the very small minority of equipment that is DC coupled, the LM4985 monitors the current supplied by the amplifier that drives the headphone jack's sleeve. If this current exceeds $500mA_{PEAK}$, the amplifier is shutdown, protecting the LM4985 and the external equipment.

POWER DISSIPATION

Power dissipation is a major concern when using any power amplifier. When operating in capacitor-coupled mode (C-CUPL), Equation 1 states the maximum power dissipation point for a single-ended amplifier operating at a given supply voltage and driving a specified output load.

$$P_{DMAX} = 2(V_{DD})^2 / (2\pi^2 R_L)$$

When operating in the OCL mode, the LM4985's three operational amplifiers produce a maximum power dissipation given in Equation 2:

$$P_{DMAX} = [2(V_{DD})^{2} / (2\pi^{2}R_{L})] + [V_{DD}^{2} / (4\pi R_{L})]$$

The maximum power dissipation point obtained from Equation 1 or Equation 2 must not be greater than the power dissipation that results from Equation 3:

$$\mathsf{P}_{\mathsf{DMAX}} = (\mathsf{T}_{\mathsf{JMAX}} - \mathsf{T}_{\mathsf{A}}) / \theta_{\mathsf{JA}}$$

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For package YFQ0012, $\theta_{JA} = 190^{\circ}$ C/W. $T_{JMAX} = 150^{\circ}$ C for the LM4985. Depending on the ambient temperature, T_A , of the system surroundings, Equation 3 can be used to find the maximum internal power dissipation supported by the IC packaging. If the result of Equation 2 is greater than that of Equation 3, then either the supply voltage must be decreased, the load impedance increased or T_A reduced.

For a typical application using a 3.6V power supply, with a 32Ω load, the maximum ambient temperature possible without violating the maximum junction temperature is approximately 144°C provided that device operation is around the maximum power dissipation point. Thus, for typical applications, power dissipation is not an issue. Power dissipation is a function of output power and thus, if typical operation is not around the maximum power dissipation point, the ambient temperature may be increased accordingly. Refer to the Typical Performance Characteristics curves for power dissipation information for lower output powers.

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(1)

(2)

(3)

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POWER SUPPLY BYPASSING

As with any amplifier, proper supply bypassing is important for low noise performance and high power supply rejection. The capacitor location on the power supply pins should be as close to the device as possible.

Typical applications employ a regulator with 10μ F tantalum or electrolytic capacitor and a ceramic bypass capacitor which aid in supply stability. This does not eliminate the need for bypassing the supply nodes of the LM4985. A bypass capacitor value in the range of 0.1μ F to 1μ F is recommended for C_S.

MICRO POWER SHUTDOWN

The LM4985's micropower shutdown is activated or deactivated through its I^2C digital interface . Please refer to Table 1 for the I^2C Address, Register Select, and Mode Control registers. Each amplifier within the LM4985 can be shutdown individually.

Please observe the following protocol when placing an individual amplifier channel in shutdown while the other channel remains active. The protocol requires activating both channels' shutdown simultaneously, then deactivating the shutdown of the channel whose output is desired (or leaving the desire channel in shutdown mode). Also, when operating in the C-CUPL mode, a short delay time is required between activating one channel after placing both channels in shutdown. If the user finds that both channels activate when only one was chosen, increase the delay.

SELECTION OF INPUT CAPACITOR SIZE

Amplifying the lowest audio frequencies requires a high value input coupling capacitor, C_i. A high value capacitor can be expensive and may compromise space efficiency in portable designs. In many cases, however, the headphones used in portable systems have little ability to reproduce signals below 60Hz. Applications using headphones with this limited frequency response reap little improvement by using a high value input capacitor.

In addition to system cost and size, turn on time is affected by the size of the input coupling capacitor C_i . A larger input coupling capacitor requires more charge to reach its quiescent DC voltage. This charge comes from the output via the feedback Thus, by minimizing the capacitor size based on necessary low frequency response, turn-on time can be minimized. A small value of C_i (in the range of 0.22µF to 0.68µF), is recommended.

MAXIMIZING OCL MODE CHANNEL-to-CHANNEL SEPARATION

The OCL mode AC ground return (CNT_GND pin) is shared by both amplifiers. As such, any resistance between the CNT_GND pin and the load will create a voltage divider with respect to the load resistance. In a typical circuit, the amount of CNT_GND resistance can be very small, but still significant. It is significant because of the relatively low load impedances for which the LM4985 was designed to drive: 16Ω to 32Ω . The ratio of this voltage divider will determine the magnitude of any residual signal present at the CNT_GND pin. It is this residual signal that leads to channel-to-channel separation (crosstalk) degradation.

For example, for a 60dB channel-to-channel separation while driving a 16 Ω load, the resistance between the LM4985's CNT_GND pin and the load must be less than 16m Ω . This is achieved by ensuring that the trace that connects the CNT_GND pin to the headphone jack sleeve should be as short and massive as possible, given the physical constraints of any specific printed circuit board layout and design.

DEMONSTRATION BOARD AND PCB LAYOUT

Information concerning PCB layout considerations and demonstration board use and performance is found in Application Note AN-1452 (Literature Number SNAA029).

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I²C Control Register

Table 1 shows the actions that are implemented by manipulating the bits within the two internal I^2C control registers.

		L	_M4985 I2	C Contorl I	Register Ac	dressing an	d Data Chart		
		A6	A5	A4	A3	A2	A1	A0	Function
I2C Address		1	1	0	0	1	1	A0	
	D7	D6	D5	D4	D3	D2	RS1	RS0	
Register Select	0	0	0	0	0	0	0	0	Read and write the mode control register
Ocicot	0	0	0	0	0	0	0	1	Read and write the volume control register
	D7	D6	D5	D4	D3	D2	D1	D0	
		WT1	WT0	PHG	SDCH1	SDCH2	CHSEL1	CHSEL2	
	0	Х	Х	Х	Х	Х	Х	Х	D7 must always be set to 0
	-	0	0	Х	Х	Х	Х	Х	Wake-up time: 80ms (OCL), 250ms (C-CUPL)
	-	0	1	Х	Х	Х	Х	Х	Wake-up time: 110ms (OCL), 450ms (C-CUPL)
	-	1	0	Х	Х	Х	Х	Х	Wake-up time: 170ms (OCL), 850ms (C-CUPL)
	-	1	1	Х	Х	Х	х	Х	Wake-up time: 290ms (OCL), 1650ms (C-CUPL)
	-	Х	Х	1	Х	Х	х	х	Output capacitor-less mode active
Mode Control	-	х	Х	0	Х	Х	Х	Х	Output capacitor-less mode inactive
Register	-	х	Х	Х	0	0	Х	Х	Amplifier's SHUTDOWN mode active
-	-	Х	Х	Х	0	1	Х	Х	Illegal mode
	-	Х	Х	Х	1	0	Х	Х	Illegal mode
	-	х	Х	Х	1	1	Х	х	Amplifier's SHUTDOWN mode inactive
	-	х	Х	Х	Х	Х	0	02	Amplifier's Chan. 1 is Input 1 , Chan 2. is Input 2
	-	х	Х	х	Х	Х	0	1	Amplifier's Chan. 1 is Input 1 , Chan 2. is Input 1
	-	х	Х	Х	Х	Х	1	0	Amplifier's Chan. 1 is Input 2 , Chan 2. is Input 2
	-	Х	Х	Х	Х	Х	1	1	Amplifier's Chan. 1 is Input 2, Chan 2. is Input 1

Table 1. LM4985 I²C Control Register Addressing and Data Format Chart



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Volume Control Settings Binary Values

The minimum volume setting is set to -76dB when 00000 is loaded into the volume control register. Incrementing the volume control register in binary fashion increases the volume control setting, reaching full scale at 11111. Table 2 shows the value of the gain for each of the 32 binary volume control settings.

Table 2. Binary Values for the Different Volume Control Gain Settings

Gain	B4	B3	B2	B1	B0
18	1	1	1	1	1
17	1	1	1	1	0
16	1	1	1	0	1
15	1	1	1	0	0
14	1	1	0	1	1
13	1	1	0	1	0
12	1	1	0	0	1
10	1	1	0	0	0
8	1	0	1	1	1
6	1	0	1	1	0
4	1	0	1	0	1
2	1	0	1	0	0
0	1	0	0	1	1
-2	1	0	0	1	0
-4	1	0	0	0	1
6	1	0	0	0	0
-8	0	1	1	1	1
-10	0	1	1	1	0
-12	0	1	1	0	1
-14	0	1	1	0	0
-16	0	1	0	1	1
-18	0	1	0	1	0
-21	0	1	0	0	1
-24	0	1	0	0	0
-27	0	0	1	1	1
-30	0	0	1	1	0
-34	0	0	1	0	1
-38	0	0	1	0	0
-44	0	0	0	1	1
-52	0	0	0	1	0
-62	0	0	0	0	1
-76	0	0	0	0	0

Revision History

Rev	Date	Description
1.0	05/17/06	Initial WEB release.

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REVISION HISTORY

Cł	nanges from Revision B (April 2013) to Revision C	Page
•	Changed layout of National Data Sheet to TI format	27



9-Aug-2013

PACKAGING INFORMATION

Orderable Device	Status	Package Type	e Package	Pins	Package	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking	Samples
	(1)		Drawing		Qty	(2)		(3)		(4/5)	
LM4985TM/NOPB	ACTIVE	DSBGA	YFQ	12		TBD	Call TI	Call TI	-40 to 85	G H2	Samples

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes. **Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

⁽³⁾ MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

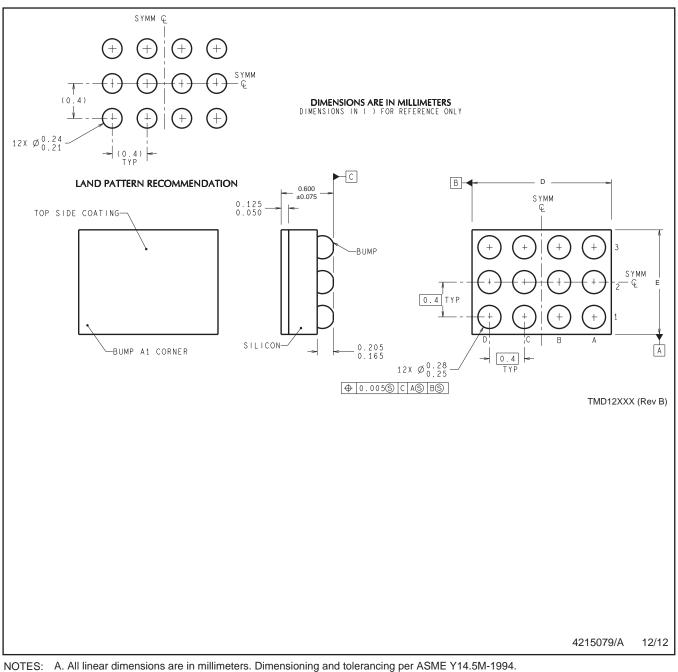
⁽⁴⁾ There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

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YFQ0012



B. This drawing is subject to change without notice.



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