

## Single Air-Core Gauge Oriver

## Description

The CS4172 is a monolithic BiCMOS integrated circuit used to translate a digital 10-bit word from a microprocessor/microcontroller to complementary DC outputs. The DC outputs drive an air-core meter commonly used in vehicle instrument panels. The 10 bits of data are used to linearly control the quadrature coils of the meter directly with a $0.35^{\circ}$ resolution and $\pm 1.0^{\circ}$ accuracy over the full $360^{\circ}$ range of the gauge. The interface from the microcontroller is by a Serial Peripheral Interface (SPI) compatible serial connection using up to a 2 MHz shift clock rate.
The digital code, which is directly proportional to the desired gauge pointer deflection, is shifted into a DAC and multiplexer. These two blocks provide a tangential conversion function to change the digital data into the appropriate DC coil voltage for the angle demanded. The tangential algorithm creates approximately $40 \%$ more torque in the meter movement than does a sin-cos algorithm at $45^{\circ}, 135^{\circ}$,
$225^{\circ}$, and $315^{\circ}$ angles. This increased torque reduces the error due to pointer droop at these critical angles.
Each output buffer is capable of supplying up to 80 mA per coil and are controlled by a common enable pin. When OE is low the output buffers are turned off but the logic portion of the chip remains powered and continues to operate normally.
The Serial Gauge Driver is self-protected against output short circuit conditions. The output drivers are disabled anytime the on-chip protection circuitry detects a short circuit condition. The outputs remain off until a falling edge is presented on CS. If the short circuit is still present the output drivers automatically disable themselves again. A thermal protection circuit limits the junction temperature to approximately $160^{\circ} \mathrm{C}$ for conditions of high supply voltage and high ambient temperature.
The status pin (ST) reflects the state of the outputs and is low whenever the outputs are disabled.


## Features

Serial Input Bus
2MHz Operating Frequency

- Tangential Drive Algorithm
- 80mA Drive Circuits
0.5 ${ }^{\circ}$ Accuracy (Typ.)
- Power-On-Reset
- Protection Features Output Short Circuit Overtemperature


## Package Options

16 Lead PDIP


16 Lead SO Wide
Internally Fused Leads


Supply Voltage
$V_{B B}$
-1.0 V to 15.0 V

Digital Inputs .......................................................................................................................................................... 1.0 V to 6.0V
Ground Potential Difference (|AGnd-DGnd |).................................................................................................................. 0.5V
Steady State Output Current ........................................................................................................................................... $\pm 100 \mathrm{~mA}$
Forced Injection Current (Inputs and Supply)............................................................................................................... $\pm 10 \mathrm{~mA}$
Operating Junction Temperature $\left(\mathrm{T}_{\mathrm{J}}\right)$............................................................................................................................... $150^{\circ} \mathrm{C}$
Storage Temperature Range ............................................................................................................................. $-65^{\circ} \mathrm{C}$ to $150^{\circ} \mathrm{C}$
Lead Temperature Soldering
Wave Solder (through hole styles only) ................................................................................... 10 sec. max, 260 C peak
Reflow (SMD styles only) .................................................................................... 60 sec. max above $183^{\circ} \mathrm{C}, 230^{\circ} \mathrm{C}$ peak
ESD Susceptibility (Human Body Model)........................................................................................................................... 2kV

## Electrical Characteristics: $-40^{\circ} \mathrm{C} \leq \mathrm{T}_{\mathrm{A}} \leq 105^{\circ} \mathrm{C} ; 7.5 \mathrm{~V} \leq \mathrm{V}_{\mathrm{BB}} \leq 14 \mathrm{~V} ; 4.5 \mathrm{~V} \leq \mathrm{V}_{\mathrm{CC}} \leq 5.5 \mathrm{~V}$; unless otherwise specified.

## 

Supply Voltages and Currents


- Digital Inputs and Outputs

| Output High Voltage | $\mathrm{SO}, \mathrm{I}_{\mathrm{OH}}=0.8 \mathrm{~mA}$ | $\mathrm{V}_{\mathrm{CC}}-0.8$ |  |  | V |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Output Low Voltage | $\begin{aligned} & \mathrm{SO}, \mathrm{I}_{\mathrm{OL}}=0.8 \mathrm{~mA} \\ & \mathrm{ST}, \mathrm{I}_{\mathrm{OL}}=2.5 \mathrm{~mA} \end{aligned}$ |  |  | $\begin{aligned} & 0.4 \\ & 0.8 \end{aligned}$ | $\begin{aligned} & \text { V } \\ & \text { V } \end{aligned}$ |
| Output High Current | ST, $\mathrm{V}_{\mathrm{CC}}=5.0 \mathrm{~V}$ |  |  | 25 | $\mu \mathrm{A}$ |
| Input High Voltage | CS, SCLK, SI, OE | $0.7 \times \mathrm{V}_{\mathrm{CC}}$ |  |  | V |
| Input Low Voltage | CS, SCLK, SI, OE |  |  | $0.3 \times \mathrm{V}_{\mathrm{CC}}$ | V |
| Input High Current | CS, SCLK, SI, OE; $\mathrm{V}_{\mathrm{IN}}=0.7 \times \mathrm{V}_{\text {CC }}$ |  |  | 1 | $\mu \mathrm{A}$ |
| Input Low Current | CS, SCLK, SI, OE; $\mathrm{V}_{\mathrm{IN}}=0.3 \times \mathrm{V}_{\mathrm{CC}}$ |  |  | 1 | $\mu \mathrm{A}$ |
| Analog Outputs |  |  |  |  |  |
| Output Function Accuracy |  | -1.2 |  | +1.2 | deg |
| Output Shutdown Current, Source | $\mathrm{V}_{\mathrm{BB}}=14.0 \mathrm{~V}$ | 70 |  | 250 | mA |
| Output Shutdown Current, Sink | $\mathrm{V}_{\mathrm{BB}}=14.0 \mathrm{~V}$ | 70 |  | 250 | mA |
| Output Shutdown Current, Source | $\mathrm{V}_{\mathrm{BB}}=7.5 \mathrm{~V}$ | 43 |  | 250 | mA |
| Output Shutdown Current, Sink | $\mathrm{V}_{\mathrm{BB}}=7.5 \mathrm{~V}$ | 43 |  | 250 | mA |
| Coil Drive Output Voltage |  |  | $0.748 \times \mathrm{V}_{\text {BB }}$ |  | V |
| Minimum Load Resistance | $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$ |  | 229 |  | $\Omega$ |
|  | $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ |  | 171 |  | $\Omega$ |
|  | $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ |  | 150 |  | $\Omega$ |


| PARAMETER | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Shift Clock Frequency |  |  |  | 2.0 | MHz |
| SCLK High Time |  | 175 |  |  | ns |
| SCLK Low Time |  | 175 |  |  | ns |
| SO Rise Time | 0.75 V to $\mathrm{V}_{\mathrm{CC}}-1.2 \mathrm{~V} ; \mathrm{C}_{\mathrm{L}}=90 \mathrm{pF}$ |  |  | 150 | ns |
| SO Fall Time | 0.75 V to $\mathrm{V}_{\mathrm{CC}}-1.2 \mathrm{~V} ; \mathrm{C}_{\mathrm{L}}=90 \mathrm{pF}$ |  |  | 150 | ns |
| SO Delay Time | $\mathrm{C}_{\mathrm{L}}=90 \mathrm{pF}$ |  |  | 150 | ns |
| SI Setup Time |  | 75 |  |  | ns |
| SI Hold Time |  | 75 |  |  | ns |
| CS Setup Time |  | 0 |  |  | ns |
| CS Hold Time |  | 75 |  |  | ns |


| Package Pin Description |  |  |  |
| :---: | :---: | :---: | :---: |
| PACKAGE PIN\# |  | PIN SYMBOL | FUNCTION |
| 16 Lead SO Wide | 16 Lead PDIP |  |  |
| 1 | 1 | $\mathrm{SIN}^{-}$ | Negative output for SINE coil. |
| 2 | 2 | SIN ${ }^{+}$ | Positive output for SINE coil. |
| 3 | 4 | $\mathrm{V}_{\text {BB }}$ | Analog supply. Nominally 13.5V. |
| 4,5,12,13 | 3,13,14 | Gnd | Ground. |
| 6 | 6 | SI | Serial data input. Data present at the rising edge of the clock signal is shifted into the internal shift register. |
| 7 | 7 | $\mathrm{V}_{\mathrm{CC}}$ | 5 V logic supply. The internal registers and latches are reset by a POR generated by the rising edge of the voltage on this pin. |
| 8 | 8 | OE | Controls the state of the output buffers. A logic low on this pin turns them off. |
| 9 | 9 | SCLK | Serial clock for shifting in/out of data. Rising edge shifts data on SI into the shift register and the falling edge changes the data on SO. |
| 10 | 10 | CS | When high allows data at SI to be shifted into part with the rising edges of SCLK. The falling edge transfers the shift register contents into the DAC and multiplexer to update the output buffers. The falling edge also re-enables the output drivers if they have been disabled by a fault. |
| 11 | 11 | ST | STATUS reflects the state of the outputs and is low anytime the outputs are disabled, either by OE or the internal protection circuitry. Requires external pull-up resistor. |
| 14 | 5 | SO | Serial data output. Existing 10-bit data is shifted out when new data is shifted in. Allows cascading of multiple devices on common serial port. |
| 15 | 15 | $\mathrm{COS}^{-}$ | Negative output for COSINE coil. |
| 16 | 16 | $\mathrm{COS}^{+}$ | Positive output for COSINE coil. |
|  | 12 | NC | No connection. |

## Theory of Operation

The SACD is for interfacing between a microcontroller or microprocessor and air-core meter movements commonly used in automotive vehicles for speedometers and tachometers. These movements are built using 2 coils placed at a $90^{\circ}$ orientation to each other. A magnetized disc floats in the middle of the coils and responds to the magnetic field generated by each coil. The disc has a shaft attached to it that protrudes out of the assembly. A pointer indicator is attached to this shaft and in conjunction with a separate printed scale displays the vehicle's speed or the engine's speed.
The disc (and pointer) respond to the vector sum of the voltages applied to the coils. Ideally, this relationship follows a $\frac{\text { sine }}{\text { cosine }}$ equation. Since this is a transcendental and non-linear function, devices of this type use an approximation for this relationship. The SACD uses a tangential algorithm as shown in Figure 1. Only 1 output varies in any 45 degree range.


Figure 1. Major gauge outputs.

## Quadrant I

$$
\theta=\operatorname{Tan}^{-1}\left[\frac{\mathrm{~V}_{\mathrm{SIN}+}-\mathrm{V}_{\mathrm{SIN}-}}{\mathrm{V}_{\mathrm{COS}+}-\mathrm{V}_{\mathrm{COS}-}}\right]
$$

For $\theta=0.176^{\circ}$ to $44.824^{\circ}$ :

$$
\begin{aligned}
& \mathrm{V}_{\mathrm{SIN}}=\operatorname{Tan} \theta \times 0.748 \times \mathrm{V}_{\mathrm{BB}} \\
& \mathrm{~V}_{\mathrm{COS}}=0.748 \times \mathrm{V}_{\mathrm{BB}}
\end{aligned}
$$

For $\theta=45.176^{\circ}$ to $89.824^{\circ}$ :

$$
\begin{aligned}
& \mathrm{V}_{\mathrm{SIN}}=0.748 \times \mathrm{V}_{\mathrm{BB}} \\
& \mathrm{~V}_{\mathrm{COS}}=\operatorname{Tan}\left(90^{\circ}-\theta\right) \times 0.748 \times \mathrm{V}_{\mathrm{BB}}
\end{aligned}
$$

## Quadrant II

$$
\theta=180^{\circ}-\operatorname{Tan}^{-1}\left[\frac{\mathrm{~V}_{\mathrm{SIN}+}-\mathrm{V}_{\mathrm{SIN}-}}{\mathrm{V}_{\mathrm{COS}+}-\mathrm{V}_{\mathrm{COS}-}}\right]
$$

For $\theta=90.176^{\circ}$ to $134.824^{\circ}$ :

$$
\begin{aligned}
& \mathrm{V}_{\mathrm{SIN}}=0.748 \times \mathrm{V}_{\mathrm{BB}} \\
& \mathrm{~V}_{\mathrm{COS}}=-\operatorname{Tan}\left(\theta-90^{\circ}\right) \times 0.748 \times \mathrm{V}_{\mathrm{BB}}
\end{aligned}
$$

For $\theta=135.176^{\circ}$ to $179.824^{\circ}$ :

$$
\begin{aligned}
& \mathrm{V}_{\mathrm{SIN}}=\operatorname{Tan}\left(180^{\circ}-\theta\right) \times 0.748 \times \mathrm{V}_{\mathrm{BB}} \\
& \mathrm{~V}_{\mathrm{COS}}=-0.748 \times \mathrm{V}_{\mathrm{BB}}
\end{aligned}
$$

## Quadrant III

$\theta=180^{\circ}+\operatorname{Tan}^{-1}\left[\frac{\mathrm{~V}_{\mathrm{SIN}+}-\mathrm{V}_{\mathrm{SIN}-}}{\mathrm{V}_{\mathrm{COS}+}-\mathrm{V}_{\mathrm{COS}-}}\right]$
For $\theta=180.176^{\circ}$ to $224.824^{\circ}$ :

$$
\begin{aligned}
& \mathrm{V}_{\mathrm{SIN}}=-\operatorname{Tan}\left(\theta-180^{\circ}\right) \times 0.748 \times \mathrm{V}_{\mathrm{BB}} \\
& \mathrm{~V}_{\mathrm{COS}}=-0.748 \times \mathrm{V}_{\mathrm{BB}}
\end{aligned}
$$

For $\theta=225.176^{\circ}-269.824^{\circ}$ :

$$
\begin{aligned}
& \mathrm{V}_{\mathrm{SIN}}=-0.748 \times \mathrm{V}_{\mathrm{BB}} \\
& \mathrm{~V}_{\mathrm{COS}}=-\operatorname{Tan}\left(270^{\circ}-\theta\right) \times 0.748 \times \mathrm{V}_{\mathrm{BB}}
\end{aligned}
$$

## Quadrant IV

$\theta=360^{\circ}-\operatorname{Tan}^{-1}\left[\frac{\mathrm{~V}_{\mathrm{SIN}+}-\mathrm{V}_{\mathrm{SIN}-}}{\mathrm{V}_{\mathrm{COS}+}-\mathrm{V}_{\mathrm{COS}-}}\right]$
For $\theta=270.176^{\circ}$ to $314.824^{\circ}$ :
$\mathrm{V}_{\mathrm{SIN}}=-0.748 \times \mathrm{V}_{\text {BB }}$
$\mathrm{V}_{\mathrm{COS}}=\operatorname{Tan}\left(\theta-270^{\circ}\right) \times 0.748 \times \mathrm{V}_{\mathrm{BB}}$

For $\theta=315.176^{\circ}-359.824^{\circ}$ :

$$
\begin{aligned}
& \mathrm{V}_{\mathrm{SIN}}=-\operatorname{Tan}\left(360^{\circ}-\theta\right) \times 0.748 \times \mathrm{V}_{\mathrm{BB}} \\
& \mathrm{~V}_{\mathrm{COS}}=0.748 \times \mathrm{V}_{\mathrm{BB}}
\end{aligned}
$$



Graph 1. Major gauge response.

To drive the gauge's pointer to a particular angle, the microcontroller sends a 10-bit digital word into the serial port. These 10 bits are divided as shown in Figure 2. However, from a software programmers viewpoint, a $360^{\circ}$ circle is divided into 1024 equal parts of $.35^{\circ}$ each. Table 1 shows the data associated with the $45^{\circ}$ divisions of the $360^{\circ}$ driver.

| Major Gauge (360ㅇ) | MSB |  |  |  |  |  |  |  |  | LSB |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | D9 | D8 | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
|  | D9 - D7 select which octant |  |  | Divides a $45^{\circ}$ octant into 128 equal parts to achieve a $.35^{\circ}$ resolution Code $0-127_{10}$ |  |  |  |  |  |  |

Figure 2. Definition of serial word.

| Input Code <br> (Decimal) | Ideal <br> Degrees | Nominal <br> Degrees | $\mathbf{V}_{\text {SIN }}$ <br> $\mathbf{( V )}$ | $\mathbf{V}_{\text {COS }}$ <br> $(\mathbf{V})$ |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0.176 | 0.032 | 10.476 |
| 128 | 45 | 45.176 | 10.476 | 10.412 |
| 256 | 90 | 90.176 | 10.476 | -0.032 |
| 384 | 135 | 135.176 | 10.412 | -10.476 |
| 512 | 180 | 180.176 | -0.032 | -10.476 |
| 640 | 225 | 225.176 | -10.476 | -10.412 |
| 768 | 270 | 270.176 | -10.476 | 0.032 |
| 896 | 315 | 315.176 | -10.476 | 10.412 |
| 1023 | 359.65 | 359.826 | -0.032 | 10.476 |

Table 1. Nominal output for major gauge $\left(V_{B B}=14 V\right)$.

The 10 bits are shifted into the device's shift register MSB first using an SPI compatible scheme. This method is shown in Figure 3. The CS must be high and remain high for SCLK to be enabled. Data on SI is shifted in on the rising edge of the synchronous clock signal. Data in the shift
register changes at SO on the falling edge of SCLK. This arrangement allows the cascading of devices. SO is always enabled. Data shifts through without affecting the outputs until CS is brought low. At this time the internal DAC is updated and the outputs change accordingly.


Figure 3. Serial data timing diagram.


Figure 4. Power-up sequence.

Application Diagram



## Ordering Information

| $\frac{\text { Part Number }}{\text { CS4172XN16 }}$ |  | Description |
| :--- | :--- | :--- |
| $\frac{\text { CS4172XDWF16 }}{\text { CS4172XDWFR16 }}$ |  | $\frac{16 \text { Lead PDIP }}{16 \text { Lead SO Wide* }}$ |

ON Semiconductor and the ON Logo are trademarks of Semiconductor Components Industries, LLC (SCILLC). ON Semiconductor reserves the right to make changes without further notice to any products herein. For additional information and the latest available information, please contact your local ON Semiconductor representative.
*Internally Fused Leads

