



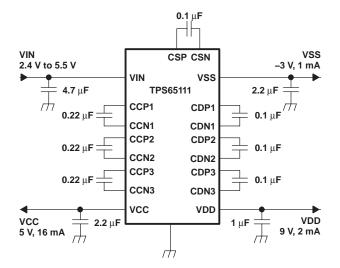
# LTPS-LCD BIAS POWER SUPPLY, TRIPLE CHARGE PUMP

# **FEATURES**

- **Complete LTPS-LCD Bias Solution**
- **Triple Output Charge Pump Providing** V<sub>CC</sub> at 16 mA, V<sub>DD</sub> at 2 mA, V<sub>SS</sub> at 1 mA
- 2.4 V to 5.5 V Input Voltage Range
- Fixed Output Voltages of 3.3 V, 7.5 V, -2.7 V or 5.0 V, 9.0 V, -3.0 V
- 50 µA Typical Quiescent Current
- Less Than 1 µA Shutdown Current
- Ultra-Low Ripple ( $V_{CC} = 5 \text{ mV}$ , Typical at 5 mA)
- Autonomous Boost for V<sub>CC</sub> Supply
- 1.5% Accuracy on Fixed VCC Output Voltage
- **Sequential Power Control**
- 24-Pin QFN Package (4 x 4)

# APPLICATIONS

- Small Form LTPS-LCD Displays
- **PDAs, Pocket PCs**
- **Smart Phones**



**APPLICATION CIRCUIT FOR TPS65111** 

# DESCRIPTION

The TPS65110/11 is a very compact power supply solution providing the three voltages required by many LTPS LCD displays.

All three regulated outputs are generated using a charge pump topology.

The VCC charge pump provides precise, high efficiency, and very low ripple dc/dc conversion for the LCD analog power. The VCC boost ratio (x1.0, x1.33, x1.5, and x2.0) is automatically set based on input and output voltage conditions. The VCC output assures 16 mA of current by using three 0.22-µF flying capacitors. If the required output current is smaller, smaller capacitors can be applied.

The VDD charge pump provides a higher positive voltage, and the VSS charge pump provides the negative output voltage. Power up/down sequences are internally set and are secured even in cases of sudden and abnormal VIN drop.

One of the most significant features of the TPS65110/11 is the ultra-low output voltage ripple, as the VCC charge pump achieves 5-mV output ripple voltage.

#### AVAILABLE OUTPUT VOLTAGE OPTIONS

PART NUMBER	VCC	V <sub>DD</sub>	V <sub>SS</sub>	V <sub>DD</sub> BOOST
TPS65110RGE	3.3 V	7.5 V	–2.7 V	X3
TPS65111RGE	5.0 V	9.0 V	–3.0 V	X2

(1) The RGE package is available in tape and reel. Add R suffix (RGER) to order quantities of 3000 parts.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.





**TPS65110** 

**TPS65111** 

These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

## ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range unless otherwise noted(1)

	UNIT
Supply voltage at VIN (2)	–0.3 V to 7.0 V
Input voltage at EN, CLK, DATA (2)	–0.3 V to VIN + 0.3 V
Power dissipation (3)	46°C/W
Virtual operation junction temperature, TJ	–40°C to 125°C
Storage temperature range	–65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

(1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) All voltage values are with respect to network ground terminal.

(3) The package thermal impedance is calculated in accordance with JESD 51-5.

# CHANNEL PERFORMANCE OVERVIEW

CHANNEL	VCC	VDD	VSS
Output Voltage Control	Regulated	Regulated	Regulated
Boost Ratio	x1; x1.333; x1.5; x2	x2 or x3	x–1
Boost setting	Autonomous Boost	Fixed	Fixed
Power Supply	VIN	VCC	VCC
Output Current	16 mA	2 mA	1 mA
Accuracy	±1.5%	±3%	±3%
Num of Ext CAP	4	4	2

## **RECOMMENDED OPERATING CONDITIONS**

	MIN	NOM	MAX	UNIT
Input voltage range, VIN	2.4		5.5	V
Main output voltage, V <sub>CC</sub>	3.0		5.2	V
Positive output voltage range, V <sub>DD</sub>	6.5		10	V
Negative output voltage range, VSS	-4.5		-2.4	V
VIN input capacitor(C <sub>i</sub> )		4.7		μF
V <sub>CC</sub> output capacitor(C <sub>CO</sub> )		2.2		μF
V <sub>DD</sub> output capacitor(C <sub>DO</sub> )		1.0		μF
V <sub>SS</sub> output capacitor(C <sub>SO</sub> )		2.2		μF
V <sub>CC</sub> flying capacitors(C <sub>C</sub> 1, C <sub>C</sub> 2, C <sub>C</sub> 3)		0.22		μF
$V_{DD}$ and $V_{SS}$ flying capacitors(CD1, CD2, CD3, CS)		0.1		μF
Operating ambient temperature, T <sub>A</sub>	-40		85	°C
Operating junction temperature, TJ	-40		125	°C

# **ELECTRICAL CHARACTERISTICS**

over recommended free-air temperature, typical at an ambient temperature of 25°C, at  $C_C1=C_C2=C_C3=0.22 \ \mu\text{F}$ ,  $C_D1=C_D2=C_D3=C_S=0.1 \ \mu\text{F}$ ,  $C_{CO}=C_{SO}=2.2 \ \mu\text{F}$ ,  $C_{DO}=1.0 \ \mu\text{F}$ ,  $V_I=3.6 \ V$ , and default output voltages (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
DEVICE		•				
VI	Input voltage range		2.4		5.5	V
IQ	Operating quiescent current	$V_I = 2.8 V$ , $EN = V_I$ , SCLK = DATA = VROM = GND, No load		50	120	μΑ
I <sub>SD</sub>	Shutdown supply current	$V_I = 2.8 V$ , EN = GND, SCLK = DATA = VROM = GND			1	μΑ
f <sub>max</sub>	Maximum operating frequency		320	400	520	kHz
		V <sub>I</sub> = 0 V to 3.6 V	2.1	2.3	2.5	
VUVLO	Under-voltage lockout threshold	V <sub>I</sub> = 3.6 V to 0 V	2.0	2.2	2.4	V
	Hysteresis		30	100		mV
LOGIC SE	CTION	1	1			
		V <sub>1</sub> = 2.4 V to 3.5 V	1.3			
VIH	EN/CLK/DATA high level input voltage	V <sub>1</sub> = 3.5 V to 5.5 V	1.5			V
VIL	EN/CLK/DATA low level input voltage				0.4	V
IIH / IIL	Logic input current	EN = GND or V		0.01	0.1	μA
	OUTPUT (V <sub>CC</sub> , V <sub>DD</sub> , V <sub>SS</sub> )	1	1		ı	
VCC	V <sub>CC</sub> Output DC voltage range	V <sub>I</sub> = 2.8 V, I <sub>VCC</sub> = 5 mA	3.25	3.3	3.35	V
IVCC	V <sub>CC</sub> Output current	$I_{VDD} = 2 \text{ mA}, I_{VSS} = 1 \text{ mA}$	16			mA
VRIPPLEC	V <sub>CC</sub> Output voltage ripple	IVCC = 5 mA		5		mV
VREGC	V <sub>CC</sub> Line regulation	V <sub>1</sub> = 2.4 V to 5.5 V		0.1	0.5	%/V
LREGC	V <sub>CC</sub> Load regulation	VI = 2.8 V, IVCC = no load to 10 mA		0.3	1	%
trC	V <sub>CC</sub> Rise time	10% to 90%, no load		100		μS
tfC	V <sub>CC</sub> Fall time	90% to 10%, no load		6		mS
		VI to VCC, VCC = $3.3$ V, IVCC = 1 mA		84%		
	V <sub>CC</sub> Efficiency	VI to VCC, VCC = 3.3 V, IVCC = 10 mA		86%		
VDD	V <sub>DD</sub> Output DC voltage range	$V_{CC} = 3.3 \text{ V}, I_{VDD} = 1.0 \text{ mA}, V_{DD} \text{ boost} = x3$	7.27	7.5	7.73	V
IVDD	V <sub>DD</sub> Output current	V <sub>CC</sub> = 3.3 V	2			mA
VRIPPLED	V <sub>DD</sub> Output voltage ripple	IVDD = 1 mA		7		mV
<sup>t</sup> rD	V <sub>DD</sub> Rise time	10% to 90%, no load		1.4		mS
<sup>t</sup> fD	V <sub>DD</sub> Fall time	90% to 10%, no load		2.4		mS
	VDD Efficiency	$V_{CC}$ to $V_{DD}$ , $V_{CC}$ = 3.3 V, $V_{DD}$ = 7.5 V, $I_{VDD}$ = 0.2 mA		70%		
		$V_{CC}$ to $V_{DD}$ , $V_{CC}$ = 3.3 V, $V_{DD}$ = 7.5 V, $I_{VDD}$ = 2 mA		70%		
VSS	VSS Output DC voltage range	V <sub>CC</sub> = 3.3 V, I <sub>VSS</sub> = 0.2 mA	-2.78	-2.7	-2.62	V
IVSS	V <sub>SS</sub> Output current	V <sub>CC</sub> = 3.3 V	1			mA
VRIPPLES	VSS Output voltage ripple	I <sub>VSS</sub> = 0.2 mA		3		mV
t <sub>rS</sub>	V <sub>SS</sub> Rise time	10% to 90%, no load		220		μS
t <sub>fS</sub>	V <sub>SS</sub> Fall time	90% to 10%, no load		2		mS
-		V <sub>CC</sub> to V <sub>SS</sub> , V <sub>CC</sub> = 3.3 V, V <sub>SS</sub> = $-2.8$ V, I <sub>VSS</sub> = $0.2$ mA		82%		
	V <sub>SS</sub> Efficiency	$V_{CC}$ to $V_{SS}$ , $V_{CC}$ = 3.3 V, $V_{SS}$ = -2.8 V, $I_{VSS}$ = 1 mA		82%		



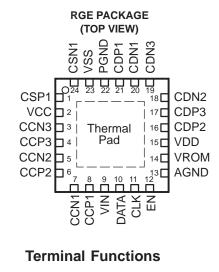
## **ELECTRICAL CHARACTERISTICS Continued**

over recommended free-air temperature, typical at an ambient temperature of 25°C, at C<sub>C</sub>1=C<sub>C</sub>2=C<sub>C</sub>3=0.22  $\mu$ F, C<sub>D</sub>1=C<sub>D</sub>2=C<sub>D</sub>3=C<sub>S</sub>=0.1  $\mu$ F, C<sub>CO</sub>=C<sub>SO</sub>=2.2  $\mu$ F, C<sub>DO</sub>=1.0  $\mu$ F, V<sub>I</sub>=3.6 V, and default output voltages (unless otherwise noted)

TPS65111	OUTPUT (V <sub>CC</sub> , V <sub>DD</sub> , V <sub>SS</sub> )					
VCC	V <sub>CC</sub> Output dc voltage range	VI = 3.6 V, IVCC = 5 mA	4.925	5.0	5.075	V
IVCC	Maximum V <sub>CC</sub> output current	$I_{VDD} = 2 \text{ mA}, I_{VSS} = 1 \text{ mA}$	16			mA
VRIPPLEC	V <sub>CC</sub> Output voltage ripple	IVCC = 5 mA		5		mV
VREGC	V <sub>CC</sub> Line regulation	V <sub>I</sub> = 2.7 V to 5.5 V		0.1	0.5	%/V
LREGC	V <sub>CC</sub> Load regulation	$V_I = 3.6 V$ , $I_{VCC} =$ no load to 10 mA		0.3	1	%
<sup>t</sup> rC	V <sub>CC</sub> Rise time	10% to 90%, no load		200		μS
<sup>t</sup> fC	V <sub>CC</sub> Fall time	90% to 10%, no load		9		mS
		VI to VCC, VCC = 5.0 V, IVCC = 1 mA		88%		
	V <sub>CC</sub> Efficiency	$V_{I}$ to $V_{CC}$ , $V_{CC}$ = 5.0 V, $I_{VCC}$ = 10 mA		90%		
V <sub>DD</sub>	V <sub>DD</sub> Output dc voltage range	$V_{CC} = 5.0 \text{ V}, \text{ I}_{VDD} = 1.0 \text{ mA}, \text{ V}_{DD} \text{ boost} = x2$	8.73	9.0	9.27	V
IVDD	Maximum V <sub>DD</sub> output current	V <sub>CC</sub> = 5.0 V	2			mA
VRIPPLED	V <sub>DD</sub> Output voltage ripple	IVDD = 1 mA		8		mV
<sup>t</sup> rD	V <sub>DD</sub> Rise time	10% to 90%, no load		1.8		mS
<sup>t</sup> fD	V <sub>DD</sub> Fall time	90% to 10%, no load		3		mS
		$V_{CC}$ to $V_{DD}$ , $V_{CC}$ = 5.0 V, $V_{DD}$ = 9.0 V, $I_{VDD}$ = 0.2mA		87%		
	V <sub>DD</sub> Efficiency	$V_{CC}$ to $V_{DD}$ , $V_{CC}$ = 5.0 V, $V_{DD}$ = 9.0 V, $I_{VDD}$ = 2mA		88%		
VSS	VSS Output dc voltage range	V <sub>CC</sub> = 5.0 V, I <sub>VSS</sub> = 0.2 mA	-3.09	-3.0	-2.91	V
IVSS	Maximum VSS output current	V <sub>CC</sub> = 5.0 V	1			mA
VRIPPLES	VSS Output voltage ripple	$I_{VSS} = 0.2 \text{ mA}$		3		mV
<sup>t</sup> rS	V <sub>SS</sub> Rise time	10% to 90%, no load		250		μS
<sup>t</sup> fS	V <sub>SS</sub> Fall time	90% to 10%, no load		2.4		mS
		V <sub>CC</sub> to V <sub>SS</sub> , V <sub>CC</sub> = 5.0 V, V <sub>SS</sub> = $-3.0$ V, I <sub>VSS</sub> = $0.2$ mA		58%		
	V <sub>SS</sub> Efficiency	V <sub>CC</sub> to V <sub>SS</sub> , V <sub>CC</sub> = 5.0 V, V <sub>SS</sub> = $-3.0$ V, I <sub>VSS</sub> = 1 mA		58%		



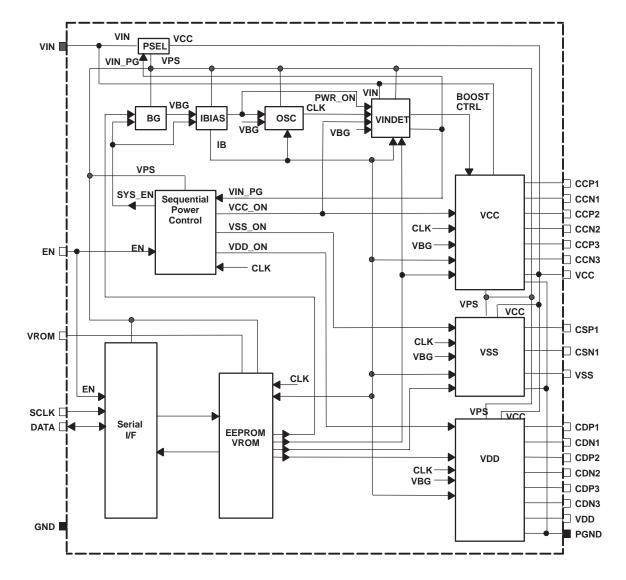
# **PIN ASSIGNMENTS**



TEF	RMINAL	DECODIDEION
NO.	NAME	DESCRIPTION
1	CSP1	VSS Positive terminal for CS
2	VCC	VCC Charge pump output
3	CCN3	VCC Negative terminal for CC3
4	CCP3	VCC Positive terminal for CC3
5	CCN2	VCC Negative terminal for CC2
6	CCP2	VCC Positive terminal for CC2
7	CCN1	VCC Negative terminal for CC1
8	CCP1	VCC Positive terminal for CC1
9	VIN	Input supply voltage
10	DATA	I <sup>2</sup> C serial data input
11	CLK	I <sup>2</sup> C serial clock input
12	EN	Power on/off enable logic input (H : active / L : shutdown)
13	AGND	Analog GND
14	VROM	EEPROM power supply
15	VDD	VDD Charge pump output
16	CDP2	VDD Positive terminal for CD2
17	CDP3	VDD Positive terminal for CD3
18	CDN2	VDD Negative terminal for CD2
19	CDN3	VDD Negative terminal for CD3
20	CDN1	VDD Negative terminal for CD1
21	CDP1	VDD Positive terminal for CD1
22	PGND	Power GND
23	VSS	VSS Charge pump output
24	CSN1	VSS Negative terminal for CS

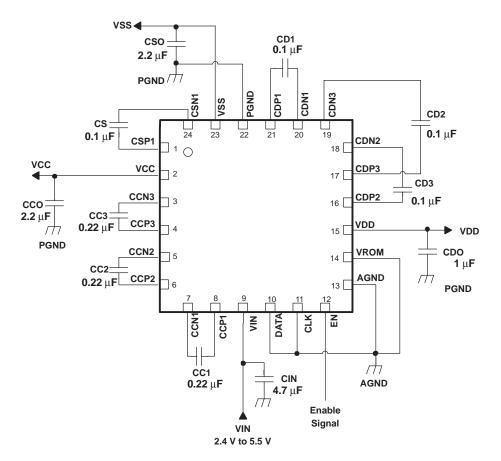


## FUNCTIONAL BLOCK DIAGRAM





# **TYPICAL APPLICATION CIRCUIT**

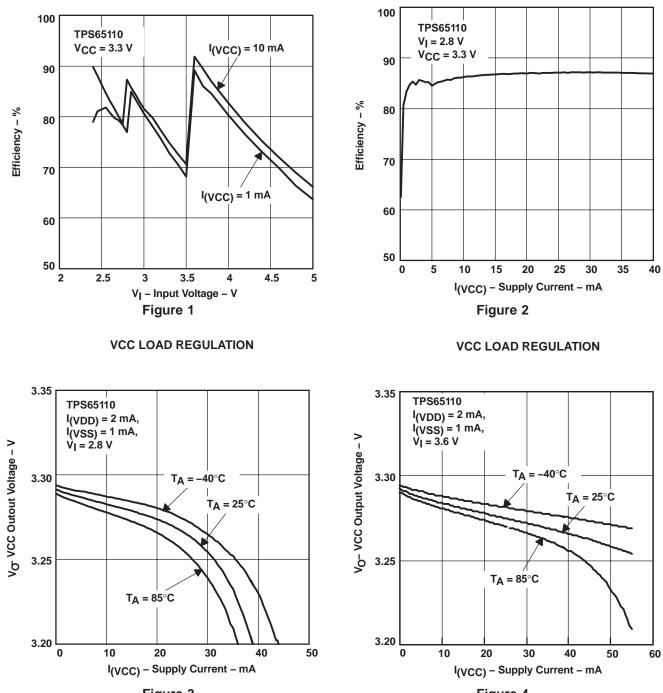




**TYPICAL CHARACTERISTICS** 

#### VCC EFFICIENCY

VCC EFFICIENCY



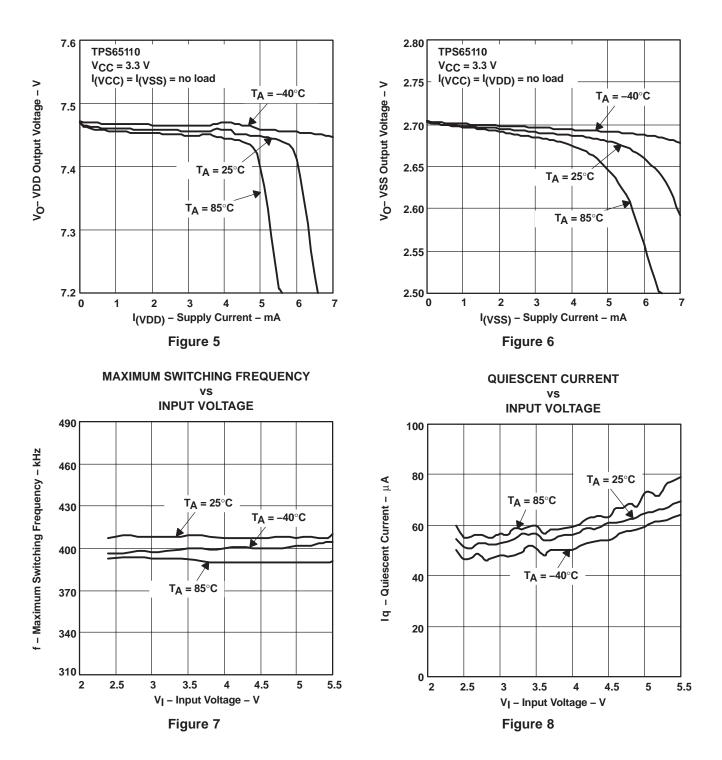




**TPS65110 TPS65111** SLVS495 – SEPTEMBER 2003

#### VDD LOAD REGULATION

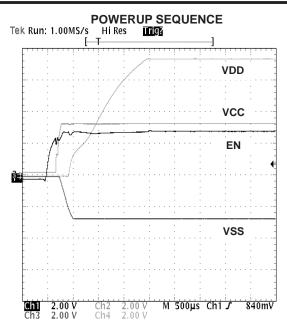
**VSS LOAD REGULATION** 





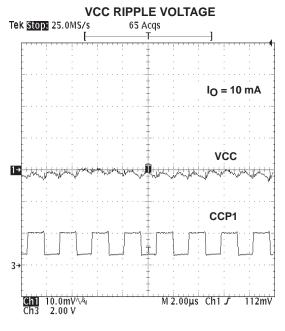
**TPS65110** 

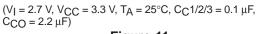
**TPS65111** 



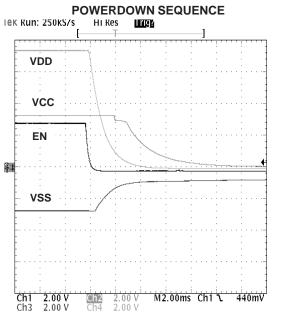
(VI = 3.0 V, V\_{CC} = 3.3 V, V\_DD = 7.5 V, V\_SS = –2.7 V, V\_DDBOOST = x3, No load, C\_C1/2/3 = 0.22  $\mu F,$  C\_CO = 2.2  $\mu F)$ 

Figure 9









(VI = 3.0 V, V<sub>CC</sub> = 3.3 V, V<sub>DD</sub> = 7.5 V, V<sub>SS</sub> = -2.7 V, V<sub>DDBOOST</sub> = x3, No load, C<sub>C</sub>1/2/3 = 0.22  $\mu$ F, C<sub>CO</sub> = 2.2  $\mu$ F) Figure 10

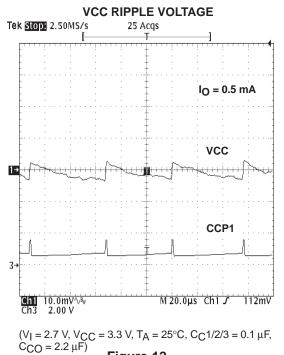


Figure 12



# DETAILED DESCRIPTION

### VCC Charge Pump

The VCC output provides a very high efficiency, regulated, dc/dc conversion through a wide input range by supporting x1.0, x1.33, x1.5, and x2.0 boost charge pump operation. TPS65110 automatically sets the boost ratio based on input and output voltage conditions. For example, when the input voltage from a battery becomes lower, the device automatically increases the boost ratio from x1.33 to x1.5. In a fixed input voltage mode, the device provides for higher conversion efficiency; for example, in the case of 2.8 V to 3.3 V conversion or 2.8 V to 5.0 V conversion. In this case, the VCC charge pump can enter into a *SKIP* mode operation in order to maintain the efficiency of a low load condition. The highest frequency of the charge pump is 400 kHz (typ). The charge pump operates by using higher frequencies in the heavier load current conditions, and decreases the frequency in the lighter load conditions. Maximum output current and operating frequency characteristics are dependent on external conditions such as the flying capacitor, output capacitor, and ambient temperature range.

		VIN [V]																								
VCC[V]	2.4	2.5	2.6	2.7	2.8	2.9	3.0	3.1	3.2	3.3	3.4	3.5	3.6	3.7	3.8	3.9	4.0	4.1	4.2	4.4	4.6	4.8	5.0	5.2	5.4	5.5
3.3		x1.5	5		x1.33					x1																
5.0		NA x2								x1.	5					x1.3	33				x1					

NOTE: Gray portion is HYSTERESIS.

Of importance, the VCC charge pump is also used as the power source for the VDD and VSS charge pumps. Therefore, consider a case where the VDD charge pump's output current is required to be 2mA, and the boost ratio is x3. With this condition, the required (additional) current for the VCC output is slightly more than 6 mA. If the VSS charge pump output current requirement is 1 mA, then the (additional) required current from VCC is another 1 mA. (Note: the VCC charge pump maintains a minimum of 16-mA output capability in addition to the loads required to support the VDD and VSS charge pumps under the recommended conditions.)

#### VDD Charge Pump

The power source for the VDD charge pump is the VCC charge pump. The output voltage and boost ratio of the VDD charge pump are fixed at either a 7.5 V and x3 boost (TPS65110), or a 9.0 V and x2 boost (TPS65111). The topology of this charge pump is *SKIP mode*, and the maximum frequency is 400 kHz. Maximum output current is dependent on the flying capacitors and ambient temperature range (refer to the typical characteristics).

#### VSS Charge Pump

The VSS charge pump is powered from the VCC charge pump and has a fixed output voltage of either -2.7 V (TPS65110) or -3.0 V (TPS65111). The boost ratio for the VSS charge pump is fixed at x–1. The operation topology is SKIP mode and has a maximum frequency of 400 kHz. Maximum output current is dependent on the flying capacitor and ambient temperature range (refer to the typical characteristics).

#### UVLO – Under Voltage Lockout

The UVLO provides for the save operation of the device. It prevents the converter from turning on when the voltage on the VIN pin is less than the threshold voltage of UVLO. Note that although the input voltage range of the product is shown to be down to 2.4 V, the maximum threshold of the UVLO for a rising VIN is 2.5 V. Therefore, to operate down to 2.4 V, the device must first be powered by a source of more than 2.5 V.

#### Enable

Low logic on the EN pin forces the TPS6511x into shutdown mode. In shutdown, the power switch, drivers, voltage reference, oscillator, and all other functions are turned off. The supply current is reduced to less than 1  $\mu$ A in shutdown mode.

#### Power-Up and Power-Down Sequencing

The TPS65110/11 controls power-up and power-down sequence through an enable pin. This signal should be terminated and not be left floating to prevent miss-operation.

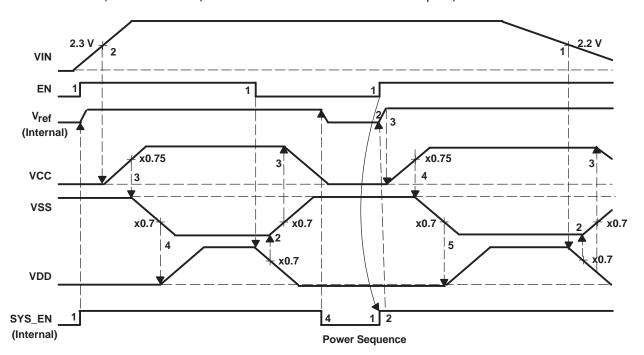
#### **Power-Up Sequence**

When the enable pin EN is pulled high, the device starts its power on sequencing. The VCC output starts up first. When the output voltage VCC has reached 75% of its nominal value, the VSS output comes up next. When VSS has reached 75% of the nominal value, the positive output VDD finally comes up.



## **Power-Down Sequencing**

When the enable pin EN is pulled low, the device starts its power-down sequencing. The VDD output goes down first. When the output voltage VDD has reached 70% of its nominal value, the VSS output goes down next. When VSS has reached 70% of the nominal value, the positive output VCC finally goes down. The TPS6511x ensures this power-down sequence even in the case of a sudden  $V_1$  drop.





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## PACKAGING INFORMATION

Orderable Device	Status <sup>(1)</sup>	Package Type	Package Drawing	Pins	Package Qty	Eco Plan <sup>(2)</sup>	Lead/ Ball Finish	MSL Peak Temp <sup>(3)</sup>	Samples (Requires Login)
TPS65110RGER	ACTIVE	VQFN	RGE	24		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	
TPS65110RGERG4	ACTIVE	VQFN	RGE	24		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	

<sup>(1)</sup> The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

<sup>(2)</sup> Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

<sup>(3)</sup> MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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# PACKAGE MATERIALS INFORMATION

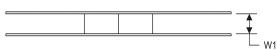
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## TAPE AND REEL INFORMATION

#### REEL DIMENSIONS

Texas Instruments





#### TAPE DIMENSIONS



A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

# TAPE AND REEL INFORMATION

\*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS65110RGER	VQFN	RGE	24	0	330.0	12.4	4.3	4.3	1.5	8.0	12.0	Q2

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# PACKAGE MATERIALS INFORMATION

16-Sep-2011



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS65110RGER	VQFN	RGE	24	0	340.5	333.0	20.6

# **MECHANICAL DATA**



NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.

- B. This drawing is subject to change without notice.
- C. Quad Flatpack, No-Leads (QFN) package configuration.
- D. The package thermal pad must be soldered to the board for thermal and mechanical performance.
- E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions. F. Falls within JEDEC MO-220.
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# RGE (S-PVQFN-N24)

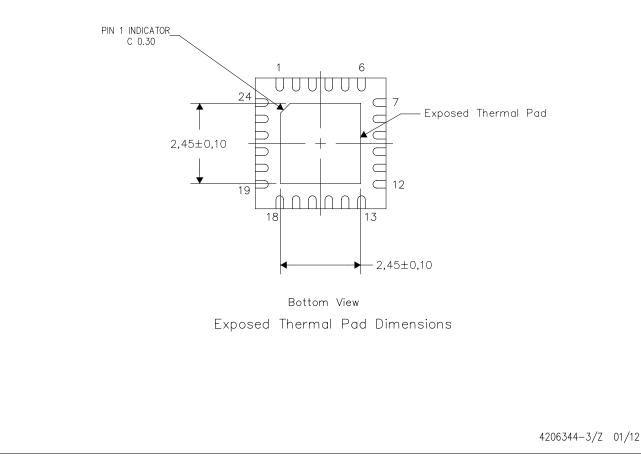
# PLASTIC QUAD FLATPACK NO-LEAD

### THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.

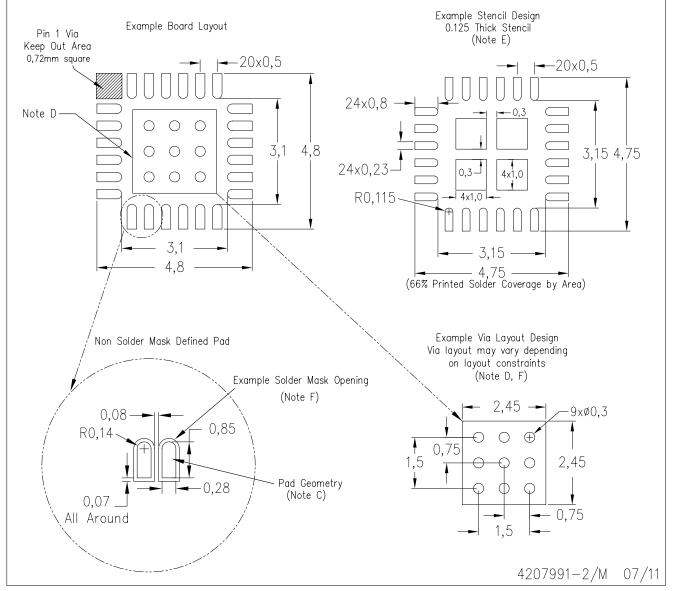


#### NOTES: A. All linear dimensions are in millimeters



# RGE (S-PVQFN-N24)

# PLASTIC QUAD FLATPACK NO-LEAD



NOTES:

- : A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Publication IPC-7351 is recommended for alternate designs.
  - D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat-Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <a href="http://www.ti.com">http://www.ti.com</a>.
  - E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
  - F. Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting recommendations for vias placed in the thermal pad.



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