To our customers,

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Renesas Electronics Corporation

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## (5400+5400) PIXELS $\times 3$ COLOR + (5400+5400) PIXELS B\&W CCD LINEAR IMAGE SENSOR

## DESCRIPTION

The $\mu$ PD8875CY-A is a color CCD (Charge Coupled Device) linear image sensor that changes optical images to electrical signal. It has 3 rows of $(5400+5400)$ staggered color pixels and $(5400+5400)$ staggered pixels of black and white, and each row has dual-sided readout type of charge transfer register. And it has reset feed-through level clamp circuits and voltage amplifiers. Therefore, it is suitable for 1200 dpi / A4 color image scanners.

## FEATURES

- Valid photocell : $(5400+5400)$ staggered pixels of RGB $+(5400+5400)$ staggered pixels of B\&W
- Photocell's size : $5.25 \times 5.75 \mu \mathrm{~m}$ (RGB), $5.25 \times 3.2 \mu \mathrm{~m}$ (B\&W)
- Line spacing : $63 \mu \mathrm{~m}$ (12 lines) Red line - Green line, Green line - Blue line $63 \mu \mathrm{~m}$ (12 lines) Blue odd line - B\&W even line
- Color filter : High transmittance new color filter Primary colors (red, green and blue), pigment filter (with light resistance 107 lx -hour)
- Resolution : $48 \mathrm{dot} / \mathrm{mm} \mathrm{A4}(210 \times 297 \mathrm{~mm})$ size (shorter side) for color and B\&W 1200 dpi US letter ( 8.5 " $\times 11^{\prime \prime}$ ) size (shorter side) for color and B\&W
- Drive clock level : CMOS output under 5 V operation
- Data rate : $20 \mathrm{MHz} \mathrm{Max} \mathrm{(RGB)}$,40 MHz Max (B\&W at 600 dpi mode)
- Power supply : + 12 V
- On-chip circuits : Reset feed-through level clamp circuits

Voltage amplifiers

## ORDERING INFORMATION

| Part Number | Package |
| :---: | :---: |
| $\mu$ PD8875CY-A | CCD linear image sensor 22 pin plastic DIP $(10.16 \mathrm{~mm}(400))$ |

Remark The $\mu$ PD8875CY-A is a lead-free product.

[^0]
## BLOCK DIAGRAM



## PIN CONFIGURATION (Top View)

CCD linear image sensor 22-pin plastic DIP (10.16 mm (400))
$\mu$ PD8875CY-A


## Caution Connect the No connection pins (NC) to GND.

## PHOTOCELL STRUCTURE DIAGRAM



PHOTOCELL ARRAY STRUCTURE DIAGRAM


## ABSOLUTE MAXIMUM RATINGS ( $\mathrm{TA}_{\mathrm{A}}=\mathbf{+ 2 5 ^ { \circ }} \mathbf{C}$ )

| Parameter | Symbol | Ratings | Unit |
| :--- | :--- | :---: | :---: |
| Output drain voltage | $\mathrm{V}_{\text {oD1 } 1}, \mathrm{VOD} 2$ | -0.3 to +15 | V |
| Shift register clock voltage | $\mathrm{V}_{\phi 1}, \mathrm{~V}_{\phi 2}$ | -0.3 to +8 | V |
| Last gate shift register clock voltage | $\mathrm{V}_{\phi 1 \mathrm{~L},} \mathrm{~V}_{\phi 2 \mathrm{~L}}$ | -0.3 to +8 | V |
| Reset gate clock voltage | $\mathrm{V}_{\phi \mathrm{R}}$ | -0.3 to +8 | V |
| Reset feed-through level clamp clock voltage | $\mathrm{V}_{\phi \mathrm{CLB}}$ | -0.3 to +8 | V |
| Mode select signal voltage | $\mathrm{V}_{\phi \text { SEL1 }}, \mathrm{V}_{\phi \text { SEL2 }}$ | -0.3 to +8 | V |
| Transfer gate clock voltage | $\mathrm{V}_{\phi \text { TG }}$ | -0.3 to +8 | V |
| Operating ambient temperature ${ }^{\text {Note }}$ | $\mathrm{T}_{\mathrm{A}}$ | 0 to +55 | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature | $\mathrm{T}_{\text {stg }}$ | -40 to +70 | ${ }^{\circ} \mathrm{C}$ |

Note Use at the condition without dew condensation.

## Caution Product quality may suffer if the absolute maximum rating is exceeded even momentarily for any

 parameter. That is, the absolute maximum ratings are rated values at which the product is on the verge of suffering physical damage, and therefore the product must be used under conditions that ensure that the absolute maximum ratings are not exceeded.
## RECOMMENDED OPERATING CONDITIONS ( $\mathrm{TA}_{\mathrm{A}}=\boldsymbol{+ 2 5 ^ { \circ }}{ }^{\circ}$ )

| Parameter | Symbol | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Output drain voltage | Vodi, Vod2 | 11.4 | 12.0 | 12.6 | V |
| Shift register clock high level | $\begin{aligned} & \mathrm{V}_{\phi 1 \mathrm{H},} \mathrm{~V}_{\phi 2 \mathrm{H}}, \mathrm{~V}_{\phi 1 \mathrm{~L},}, \\ & \mathrm{~V}_{\phi 2 \mathrm{~L}} \end{aligned}$ | 4.75 | 5.0 | 5.5 | V |
| Shift register clock low level | $\begin{aligned} & V_{\phi 1 L}, V_{\phi 2 L}, V_{\phi 1 L L}, \\ & V_{\phi 2 L L} \end{aligned}$ | 0 | 0 | +0.15 | V |
| Reset gate clock high level | $V_{\phi \text { RH }}$ | 4.75 | 5.0 | 5.5 | V |
| Reset gate clock low level | $\mathrm{V}_{\phi \text { RL }}$ | 0 | 0 | +0.15 | V |
| Reset feed-through level clamp clock high level | $\mathrm{V}_{\varphi \text { CLBH }}$ | 4.75 | 5.0 | 5.5 | V |
| Reset feed-through level clamp clock low level | $\mathrm{V}_{\text {¢ CLBL }}$ | 0 | 0 | +0.15 | V |
| Mode select signal high level | $\mathrm{V}_{\phi \text { SELIH, }} \mathrm{V}_{\phi \text { SEL2H }}$ | 4.75 | 5.0 | 5.5 | V |
| Mode select signal low level | $\mathrm{V}_{\phi \text { SELIL, }} \mathrm{V}_{\phi \text { SEL2L }}$ | 0 | 0 | +0.15 | V |
| Transfer gate clock high level | $\mathrm{V}_{\text {¢TGH }}$ | 4.75 | $\mathrm{V}_{\phi \text { 1 }}{ }^{\text {Note }}$ | $\mathrm{V}_{\phi \text { 1H }}{ }^{\text {Note }}$ | V |
| Transfer gate clock low level | $\mathrm{V}_{\phi \text { TGL }}$ | 0 | 0 | +0.15 | V |
| Data rate | $\mathrm{f}_{\phi} \mathrm{R}$ | - | 2 | 20 | MHz |
| Clock pulse frequency | $\mathrm{f}_{\phi 1}, \mathrm{f}_{\phi} 2$ | - | 1 | 20 | MHz |

Note When Transfer gate clock high level ( $\mathrm{V}_{\phi \text { TGH }}$ ) is higher than shift register clock high level $\left(\mathrm{V}_{\phi 1 \mathrm{H}}\right)$, image lag increases.

## ELECTRICAL CHARACTERISTICS

$\binom{T_{A}=+25^{\circ} \mathrm{C}, \mathrm{Vod}_{00}=+12 \mathrm{~V}$, data rate $\left(\mathrm{f}_{\phi \mathrm{R}}\right)=2 \mathrm{MHz}$, storage time $=5.5 \mathrm{~ms}$, input clock $=5 \mathrm{~V}_{\mathrm{p}-\mathrm{p}}}{$ light source: 3200 K halogen lamp $+\mathrm{C}-500 \mathrm{~S}$ (infrared cut filter, $\mathrm{t}=1 \mathrm{~mm}$ ) $+\mathrm{HA}-50$ (heat absorbing filter, $\mathrm{t}=3 \mathrm{~mm}$ ) }

| Parameter |  | Symbol | Test Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Saturation voltage |  | $V_{\text {sat }}$ |  | 2.5 | 3.0 | - | V |
| Saturation exposure | Red | SE_R |  | - | 0.3 | - | Ixos |
|  | Green | SE_G |  | - | 0.33 | - | Ixos |
|  | Blue | SE_B |  | - | 0.6 | - | Ix•s |
|  | B\&W | SE_B\&W |  | - | 0.24 | - | Ix•s |
| Photo response non-uniformity |  | PRNU_RGB | Vout $=1.0 \mathrm{~V}$ | - | 6.0 | 20.0 | \% |
|  |  | PRNU_B\&W |  | - | 10.0 | 25.0 | \% |
| Average dark signal |  | ADS | Light shielding | - | 0.2 | 2.0 | mV |
| Dark signal non-uniformity |  | DSNU | Light shielding | - | 1.5 | 10.0 | mV |
| Power consumption |  | Pw | Light shielding | - | 360 | 540 | mW |
| Output impedance |  | Zo |  | - | 0.2 | 0.4 | k ת |
| Response | Red | RR |  | 7.0 | 10.0 | 13.0 | VIIx•s |
|  | Green | Rg |  | 6.3 | 9.0 | 11.7 | VIIx*s |
|  | Blue | RB |  | 3.5 | 5.0 | 6.5 | V/Ix•s |
|  | B\&W | Rbsw |  | 8.7 | 12.3 | 16.1 | V/Ix•s |
| Image lag |  | IL | Vout $=1.0 \mathrm{~V}$ | - | 3.0 | 7.0 | \% |
| Offset level |  | Vos |  | 6.5 | 7.5 | 8.5 | V |
| Output fall delay time ${ }^{\text {Note }}$ |  | td | Vout $=1.0 \mathrm{~V}$ | - | 15 | - | ns |
| Total transfer efficiency ${ }^{\text {Note }}$ |  | TTE | $\begin{aligned} & \text { Vout }=1.0 \mathrm{~V}, \\ & \text { data rate }=20 \mathrm{MHz} \end{aligned}$ | 92 | 98 | - | \% |
| Register imbalance |  | RI | Vout $=1.0 \mathrm{~V}$ | - | 1.0 | 4.0 | \% |
| Response peak | Red |  |  | - | 610 | - | nm |
|  | Green |  |  | - | 535 | - | nm |
|  | Blue |  |  | - | 460 | - | nm |
|  | B\&W |  |  | - | 540 | - | nm |
| Dynamic range |  | DR1 | $\mathrm{V}_{\text {sat }} /$ DSNU | - | 2000 | - | times |
|  |  | DR2 | $\mathrm{V}_{\text {sat }} / \sigma$ CDS | - | 1363 | - | times |
| Reset feed-through noise |  | RFTN | Light shielding | -2000 | -100 | 500 | mV |
|  |  | PRFTN | Light shielding | - | 500 | 800 | mV |
| Random noise (CDS) |  | $\sigma$ CDS | Light shielding | - | 2.2 | - | mV |

Note When the fall time of $\phi 1 \mathrm{~L}$ and $\phi 2 \mathrm{~L}(\mathrm{t} 1, \mathrm{t} 2)$ is typical value. (Refer to TIMING CHART 2-1 to 2-3)

INPUT PIN CAPACITANCE $\left(\mathrm{T}_{\mathrm{A}}=\boldsymbol{+ 2 5}{ }^{\circ} \mathrm{C}, \mathrm{Vod}=\boldsymbol{+ 1 2} \mathrm{V}\right)$

| Parameter | Symbol | Pin name | Pin No | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Shift register clock pin capacitance 1 | $\mathrm{C}_{\phi 1}$ | $\phi 1$ | 5 | - | 600 | - | pF |
|  |  |  | 9 | - | 600 | - | pF |
|  |  |  | 14 | - | 600 | - | pF |
|  | $\mathrm{C}_{\phi 1}$ total capacitance |  |  | - | 1800 | - | pF |
| Shift register clock pin capacitance 2 | $\mathrm{C}_{\phi 2}$ | $\phi 2$ | 10 | - | 600 | - | pF |
|  |  |  | 15 | - | 600 | - | pF |
|  |  |  | 17 | - | 600 | - | pF |
|  | C $\phi 2$ total capacitance |  |  | - | 1800 | - | pF |
| Last gate shift register clock pin capacitance 1 | $\mathrm{C}_{\phi 12}$ | $\phi 1 \mathrm{~L}$ | 8 | - | 10 | - | pF |
| Last gate shift register clock pin capacitance 2 | $\mathrm{C}_{\phi 2 \mathrm{~L}}$ | $\phi 2 \mathrm{~L}$ | 16 | - | 10 | - | pF |
| Reset gate clock pin capacitance | $\mathrm{C}_{\varphi} \mathrm{R}$ | $\phi \mathrm{R}$ | 3 | - | 10 | - | pF |
| Reset feed-through level clamp clock pin capacitance | $\mathrm{C}_{\phi \text { CLb }}$ | $\phi$ CLB | 4 | - | 10 | - | pF |
| Select signal pin capacitance | $\mathrm{C}_{\text {¢ SEL } 1}$ | $\phi$ SEL1 | 12 | - | 10 | - | pF |
|  | $\mathrm{C}_{\phi}$ SEL2 | $\phi$ SEL2 | 7 | - | 10 | - | pF |
| Transfer gate clock pin capacitance | $\mathrm{C}_{\phi \text { TG }}$ | $\phi$ TG | 13 | - | 300 | - | pF |

Remark 1. Pins $5,9,14(\phi 1)$ and pins $10,15,17(\phi 2)$ are each connected inside of the device.
2. $\mathrm{C}_{\phi 1}$ and $\mathrm{C}_{\phi 2}$ show the equivalent capacity of the real drive including the capacity of between $\phi 1$ and $\phi 2$.

Note Set the $\phi \mathrm{R}$ to low level and the $\phi \mathrm{CLB}$ to high level during this period.
TIMING CHART 1-2 (Color 600 dpi Mode, ( $\phi$ SEL1 = "L", $\phi$ SEL2 = "H"))

Note Set the $\phi \mathrm{R}$ to low level and the $\phi$ CLB to high level during this period.
TIMING CHART 1-3 (Color 300 dpi Mode, ( $\phi$ SEL1 = "L", $\phi$ SEL2 = "H") )

Note Set the $\phi \mathrm{R}$ to low level and the $\phi$ CLB to high level during this period.
TIMING CHART 1-4 (B\&W Mode, ( $\phi$ SEL1 = N/A, $\phi$ SEL2 = "L"))

Note Set the $\phi \mathrm{R}$ to low level and the $\phi \mathrm{CLB}$ to high level during this period.

TIMING CHART 2-1 (Color 1200 dpi Mode, ( $\phi$ SEL1 = "H", $\phi$ SEL2 = "H") $)$


| Symbol | MIN. | TYP. | MAX. | Unit |
| :--- | :---: | :---: | :---: | :---: |
| $\mathrm{t} 1, \mathrm{t} 2$ | 0 | 25 | - | ns |
| t 3 | 10 | 50 | - | ns |
| $\mathrm{t} 4, \mathrm{t} 5$ | 0 | 20 | - | ns |
| t 6 | 0 | 70 | - | ns |
| t 7 | 15 | 50 | - | ns |
| $\mathrm{t} 8, \mathrm{t} 9$ | 0 | 20 | - | ns |
| t 10 | 5 | 45 | - | ns |
| t 11 | 10 | 70 | - | ns |

Note TYP. is the case of $\phi \mathrm{R}=2 \mathrm{MHz}$

TIMING CHART 2-2 (Color 600 dpi Mode, ( $\phi$ SEL1 = "L", $\phi$ SEL2 = "H") / B\&W Mode, ( $\phi$ SEL1 = N/A, $\phi S E L 2=$ "L"))


| Symbol | MIN. | TYP. | MAX. | Unit |
| :--- | :---: | :---: | :---: | :---: |
| $\mathrm{t} 1, \mathrm{t} 2$ | 0 | 25 | - | ns |
| t 3 | 10 | 50 | - | ns |
| $\mathrm{t} 4, \mathrm{t} 5$ | 0 | 20 | - | ns |
| t 6 | 0 | 70 | - | ns |
| t 7 | 15 | 50 | - | ns |
| $\mathrm{t} 8, \mathrm{t} 9$ | 0 | 20 | - | ns |
| t 10 | 5 | 45 | - | ns |
| t 11 | 10 | 70 | - | ns |

Note TYP. is the case of $\phi \mathrm{R}=2 \mathrm{MHz}$

TIMING CHART 2-3 (Color 300 dpi Mode, ( $\phi$ SEL1 = "L", $\phi$ SEL2 = "H")


| Symbol | MIN. | TYP. | MAX. | Unit |
| :--- | :---: | :---: | :---: | :---: |
| $\mathrm{t} 1, \mathrm{t} 2$ | 0 | 25 | - | ns |
| t 3 | 10 | 50 | - | ns |
| $\mathrm{t} 4, \mathrm{t} 5$ | 0 | 20 | - | ns |
| t 6 | 0 | 70 | - | ns |
| t 7 | 15 | 50 | - | ns |
| $\mathrm{t} 8, \mathrm{t} 9$ | 0 | 20 | - | ns |
| t 10 | 5 | 45 | - | ns |
| t 11 | 10 | 70 | - | ns |

Note TYP. is the case of $\phi \mathrm{R}=1 \mathrm{MHz}$
$\phi$ TG, $\phi 1$ ( $\phi 1 \mathrm{~L}$ ), $\phi 2$ ( $\phi 2 \mathrm{~L}$ ) TIMING CHART


| Symbol | MIN. | TYP. | MAX. | Unit |
| :--- | :---: | :---: | :---: | :---: |
| t 6 | 0 | 70 | - | ns |
| t 7 t | 15 | 50 | - | ns |
| t 11 | 10 | 50 | - | ns |
| t 13 | 5000 | 10000 | 50000 | ns |
| $\mathrm{t} 14, \mathrm{t} 15$ | 0 | 50 | - | ns |
| $\mathrm{t} 16, \mathrm{t} 17$ | 900 | 1000 | - | ns |
| $\mathrm{t} 18, \mathrm{t} 19$ | 200 | 400 | - | ns |
| t 20 | 10 | 350 | - | ns |

Note Set the $\phi \mathrm{R}$ to low level and the $\phi \mathrm{CLB}$ to high level during this period.
$\phi 1, \phi 2$ CROSS POINT


| Symbol | MIN. | TYP. | MAX. | Unit |
| :--- | :---: | :---: | :---: | :---: |
| t 25 | 50 | - | - | ns |
| $\mathrm{t} 26, \mathrm{t} 27$ | 20 | - | - | ns |

$\phi 1, \phi 2 L$ CROSS POINT

$\phi 2, \phi 1 L$ CROSS POINT


## $\phi$ TG, $\phi$ SEL TIMING CHART



| Symbol | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: |
| t 28 | 0 | 0 | - | ns |
| t 29 | 4500 | 9500 | - | ns |

## SELECTION OF RESOLUTION MODE

The uPD8875CY-A has function of two readout modes, High Resolution Mode and Low Resolution Mode. These two modes can be selected by $\phi$ SEL1 switch.

| Mode | Description | $\phi$ SEL1 |
| :--- | :--- | :--- |
| High Resolution Mode | 1200 dpi (Max.) | High level |
| Low Resolution Mode | 600 dpi (Max.) (even line readout mode) | Low level |

## (1) High Resolution Mode

In this mode, both signals in odd lines and even lines can be read out. This mode enables 1200 dpi (max.) resolution with A4 size ( $210 \times 297 \mathrm{~mm}$, shorter side).
Please refer to TIMING CHART 1-1 and TIMING CHART 2-1.

## (2) Low Resolution Mode

In this mode, only signal output in even lines can be read out.

Signal output in even lines: Can be read out
Signal output in odd lines: Can not be read out

This mode enables 600 dpi (max) resolution with A4 size.
To use intermittent reset drive enable signal charges of adjacent pixels in even line to add at the charge to voltage conversion area. Then it can achieve low resolution with A4 size such as 300, 200, 150 dpi.
Please refer to TIMING CHART 1-2, 1-3 and TIMING CHART 2-2, 2-3.

## DEFINITIONS OF CHARACTERISTIC

1. Saturation voltage: Vsat

Output signal voltage at which the response linearity is lost.
2. Saturation exposure : SE

Product of intensity of illumination (lx) and storage time (s) when saturation of output voltage occurs.

## 3. Photo response non-uniformity : PRNU

The output signal non-uniformity of all the valid pixels when the photosensitive surface is applied with the light of uniform illumination. This is calculated by the following formula.


## 4. Average dark signal : ADS

Average output signal voltage of all the valid pixels at light shielding. This is calculated by the following formula.

$$
\operatorname{ADS}(\mathrm{mV})=\frac{\sum_{\mathrm{j}=1}^{10800} d_{j}}{10800}
$$

## 5. Dark signal non-uniformity : DSNU

Absolute maximum of the difference between ADS and voltage of the highest or lowest output pixel of all the valid pixels at light shielding. This is calculated by the following formula.

DSNU (mV) : maximum of $|\mathrm{dj}-\operatorname{ADS}| \mathrm{j}=1$ to 10800
dj : Dark signal of valid pixel number j


## 6. Output impedance: Zo

Impedance of the output pins viewed from outside.
7. Response : R

Output voltage divided by exposure ( $\mathrm{l} \times \cdot \mathrm{s}$ ).
Note that the response varies with a light source (spectral characteristic).

## 8. Image lag: IL

The rate between the last output voltage and the next one after read out the data of a line.

$\mathrm{IL}(\%)=\frac{\mathrm{V}_{1}}{\text { Vout }} \times 100$

## 9. Register imbalance : RI

The rate of the difference between the averages of the output voltage of Odd and Even pixels, against the average output voltage of all the valid pixels.

$$
R I(\%)=\frac{\frac{2}{n}\left|\sum_{j=1}^{\frac{n}{2}}\left(V_{2 j-1}-V_{2 j}\right)\right|}{\frac{1}{n} \sum_{j=1}^{n} V_{j}} \times 100 \quad \begin{aligned}
n & : \text { Number of valid pixels } \\
V_{j} & : \text { Output voltage of each pixel }
\end{aligned}
$$

10. Offset level : VOS

DC level of output signal is defined as follows.
11. Reset feed-through noise : RFTN, PRFTN

Reset feed-through noise (RFTN) and peak of RFTN (PRFTN) are defined as follows.


## 12. Random noise (CDS) : $\sigma C D S$

Random noise $\sigma$ CDS is defined as the standard deviation of a valid pixel output signal with 100 times (= 100 lines) data sampling at dark (light shielding). $\sigma$ CDS is calculated by the following procedure.

1. One valid photocell in one reading is fixed as measurement point.
2. The output level is measured during the reset feed-through period which is averaged over 100 ns to get "VDi".
3. The output level is measured during the video output time averaged over 100 ns to get "VOi".
4. The correlated double sampling output is defined by the following formula.

$$
\mathrm{VCDS}_{\mathrm{i}}=\mathrm{VD}_{\mathrm{i}}-\mathrm{VO}_{\mathrm{i}}
$$

5. Repeat the above procedure ( 1 to 4 ) for 100 times (= 100 lines).
6. Calculate the standard deviation $\sigma$ CDS using the following formula equation.

$$
\sigma \operatorname{CDS}(\mathrm{mV})=\sqrt{\frac{\sum_{i=1}^{100}\left(\mathrm{VCDS}_{\mathrm{i}}-\overline{\mathrm{V}}\right)^{2}}{100}}, \overline{\mathrm{~V}}=\frac{1}{100} \sum_{\mathrm{i}=1}^{100} \mathrm{VCDS}_{\mathrm{i}}
$$

The following figure shows output waveform (valid photocell under dark condition).


## STANDARD CHARACTERISTIC CURVES (1) (Reference Value)



TOTAL SPECTRAL RESPONSE CHARACTERISTICS
(Without infrared cut filter and heat absorbing filter) ( $\mathrm{T}_{\mathrm{A}}=\mathbf{+ 2 5}{ }^{\circ} \mathrm{C}$ )


## APPLICATION CIRCUIT EXAMPLE



## Caution Connect the no connection pins (NC) to GND.

Remark The inverters are the 74AC04, and pins 5, 9, 10, 13, 14 and 17 connect two or three inverters in parallel.


## PACKAGE DRAWING

## $\mu$ PD8875CY-A

CCD LINEAR IMAGE SENSOR 22-PIN PLASTIC DIP (10.16 mm (400))


| Name | Dimensions | Refractive index |
| :---: | :---: | :---: |
| Plastic cap | $42.7 \times 8.35 \times 0.8\left(0.7^{* 5}\right)$ | 1.5 |

$※ 1$ Distance between the 1st valid pixel and the center of the pin1
$※ 2$ Distance between the top of the cap and the surface of the CCD chip
$※ 3$ Distance between the bottom of the package and the surface of the CCD chip ※4 Transparent window
$※ 5$ Thickness of the transparent window

## RECOMMENDED SOLDERING CONDITIONS

When soldering this product, it is highly recommended to observe the conditions as shown below.
If other soldering processes are used, or if the soldering is performed under different conditions, please make sure to consult with our sales offices.

Type of Through-hole Device
$\mu$ PD8875CY-A: CCD linear image sensor 22-pin plastic DIP (10.16 mm (400))

| Process |  |
| :---: | :--- |
| Partial heating method | Pin temperature: $380^{\circ} \mathrm{C}$ or below, Heat time: 3 seconds or less (per pin). |

Cautions 1. During assembly care should be taken to prevent solder or flux from contacting the glass cap. The optical characteristics could be degraded by such contact.
2. Soldering by the solder flow method may have deleterious effects on prevention of glass cap soiling and heat resistance. So the method cannot be guaranteed.

## NOTES ON HANDLING THE PACKAGES

## (1) DUST AND DIRT PROTECTING

The optical characteristics of the CCD will be degraded if the cap is scratched during cleaning. Don't either touch plastic cap surface by hand or have any object come in contact with plastic cap surface. Should dirt stick to a plastic cap surface, blow it off with an air blower. For dirt stuck through electricity ionized air is recommended. And if the plastic cap surface is grease stained, clean with our recommended solvents.

O CLEANING THE PLASTIC CAP
Care should be taken when cleaning the surface to prevent scratches.
We recommend cleaning the cap with a soft cloth moistened with one of the recommended solvents below. Excessive pressure should not be applied to the cap during cleaning. If the cap requires multiple cleanings it is recommended that a clean surface or cloth be used.

## O RECOMMENDED SOLVENTS

The following are the recommended solvents for cleaning the CCD plastic cap.
Use of solvents other than these could result in optical or physical degradation in the plastic cap. Please consult your sales office when considering an alternative solvent.

| Solvents | Symbol |
| :--- | :--- |
| Ethyl Alcohol | EtOH |
| Methyl Alcohol | MeOH |
| Isopropyl Alcohol | IPA |
| N-methyl Pyrrolidone | NMP |

## MOUNTING OF THE PACKAGE

The application of an excessive load to the package may cause the package to warp or break, or cause chips to come off internally. Particular care should be taken when mounting the package on the circuit board. Don't have any object come in contact with plastic cap. You should not reform the lead frame. We recommended to use a IC-inserter when you assemble to PCB.

Also, be care that the any of the following can cause the package to crack or dust to be generated.

1. Applying heat to the external leads for an extended period of time with soldering iron.
2. Applying repetitive bending stress to the external leads.
3. Rapid cooling or heating

## OPERATE AND STORAGE ENVIRONMENTS

Operate in clean environments. CCD image sensors are precise optical equipment that should not be subject to mechanical shocks. Exposure to high temperatures or humidity will affect the characteristics. So avoid storage or usage in such conditions.

Keep in a case to protect from dust and dirt. Dew condensation may occur on CCD image sensors when the devices are transported from a low-temperature environment to a high-temperature environment. Avoid such rapid temperature changes.

For more details, refer to our document "Review of Quality and Reliability Handbook" (C12769E)

## (4) ELECTROSTATIC BREAKDOWN

CCD image sensor is protected against static electricity, but destruction due to static electricity is sometimes detected. Before handling be sure to take the following protective measures.

1. Ground the tools such as soldering iron, radio cutting pliers of or pincer.
2. Install a conductive mat or on the floor or working table to prevent the generation of static electricity.
3. Either handle bare handed or use non-chargeable gloves, clothes or material.
4. Ionized air is recommended for discharge when handling CCD image sensor.
5. For the shipment of mounted substrates, use box treated for prevention of static charges.
6. Anyone who is handling CCD image sensors, mounting them on PCBs or testing or inspecting PCBs on which CCD image sensors have been mounted must wear anti-static bands such as wrist straps and ankle straps which are grounded via a series resistance connection of about $1 \mathrm{M} \Omega$.

## NOTES FOR CMOS DEVICES

## (1) VOLTAGE APPLICATION WAVEFORM AT INPUT PIN

Waveform distortion due to input noise or a reflected wave may cause malfunction. If the input of the CMOS device stays in the area between $\mathrm{V}_{\mathrm{IL}}$ (MAX) and $\mathrm{V}_{\mathrm{IH}}(\mathrm{MIN})$ due to noise, etc., the device may malfunction. Take care to prevent chattering noise from entering the device when the input level is fixed, and also in the transition period when the input level passes through the area between VIL (MAX) and $\mathrm{V}_{\mathrm{H}}$ (MIN).

## HANDLING OF UNUSED INPUT PINS

Unconnected CMOS device inputs can be cause of malfunction. If an input pin is unconnected, it is possible that an internal input level may be generated due to noise, etc., causing malfunction. CMOS devices behave differently than Bipolar or NMOS devices. Input levels of CMOS devices must be fixed high or low by using pull-up or pull-down circuitry. Each unused pin should be connected to Vod or GND via a resistor if there is a possibility that it will be an output pin. All handling related to unused pins must be judged separately for each device and according to related specifications governing the device.

## (3) PRECAUTION AGAINST ESD

A strong electric field, when exposed to a MOS device, can cause destruction of the gate oxide and ultimately degrade the device operation. Steps must be taken to stop generation of static electricity as much as possible, and quickly dissipate it when it has occurred. Environmental control must be adequate. When it is dry, a humidifier should be used. It is recommended to avoid using insulators that easily build up static electricity. Semiconductor devices must be stored and transported in an anti-static container, static shielding bag or conductive material. All test and measurement tools including work benches and floors should be grounded. The operator should be grounded using a wrist strap. Semiconductor devices must not be touched with bare hands. Similar precautions need to be taken for PW boards with mounted semiconductor devices.

## (4) STATUS BEFORE INITIALIZATION

Power-on does not necessarily define the initial status of a MOS device. Immediately after the power source is turned ON, devices with reset functions have not yet been initialized. Hence, power-on does not guarantee output pin levels, I/O settings or contents of registers. A device is not initialized until the reset signal is received. A reset operation must be executed immediately after power-on for devices with reset functions.

## POWER ON/OFF SEQUENCE

In the case of a device that uses different power supplies for the internal operation and external interface, as a rule, switch on the external power supply after switching on the internal power supply. When switching the power supply off, as a rule, switch off the external power supply and then the internal power supply. Use of the reverse power on/off sequences may result in the application of an overvoltage to the internal elements of the device, causing malfunction and degradation of internal elements due to the passage of an abnormal current.
The correct power on/off sequence must be judged separately for each device and according to related specifications governing the device.

## (6) INPUT OF SIGNAL DURING POWER OFF STATE

Do not input signals or an I/O pull-up power supply while the device is not powered. The current injection that results from input of such a signal or I/O pull-up power supply may cause malfunction and the abnormal current that passes in the device at this time may cause degradation of internal elements. Input of signals during the power off state must be judged separately for each device and according to related specifications governing the device.

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