J PACKAGE

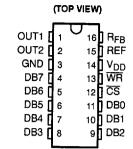
SGLS028A - SEPTEMBER 1989 - REVISED MARCH 1995

- Advanced LinCMOS™ Silicon-Gate Technology
- Easily interfaced to Microprocessors
- On-Chip Data Latches
- Monotonicity Over Entire A/D Conversion Range
- Segmented High-Order Bits Ensure Low-Glitch Output
- Designed to Be interchangeable With Analog Devices AD7524, PMI PM-7524, and Micro Power Systems MP7524
- Fast Control Signaling for Digital Signal Processor Applications Including Interface With SMJ320

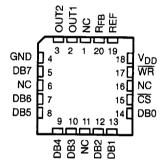
KEY PERFORMANCE SPECIFICATIONS					
Resolution 8 Bits					
Linearity error 1/2 LSB Max					
Power dissipation at V <sub>DD</sub> = 5 V 5 mW Max					
Settling time 100 ns Max					
Propagation delay 80 ns Max					

## description

The AD7524M is an Advanced LinCMOS™ 8-bit digital-to-analog converter (DAC) designed for easy interface to most popular microprocessors.



FK PACKAGE (TOP VIEW)



NC-No internal connection

The AD7524M is an 8-bit multiplying DAC with input latches and with a load cycle similar to the write cycle of a random access memory. Segmenting the high-order bits minimizes glitches during changes in the most-significant bits, which produce the highest glitch impulse. The AD7524M provides accuracy to 1/2 LSB without the need for thin-film resistors or laser trimming, while dissipating less than 5 mW typically.

Featuring operation from a 5-V to 15-V single supply, the AD7524M interfaces easily to most microprocessor buses or output ports. Excellent multiplying (2 or 4 quadrant) makes the AD7524M an ideal choice for many microprocessor-controlled gain-setting and signal-control applications.

The AD7524M is characterized for operation from -55°C to 125°C.

#### **AVAILABLE OPTIONS**

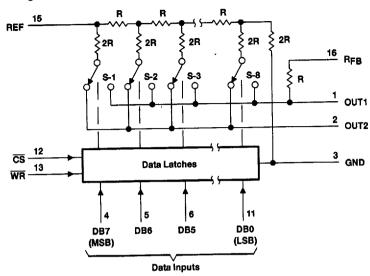
	PAC	KAGE		
TA	CERAMIC CHIP CARRIER (FK) CERAMIC DI			
-55°C to 125°C	AD7524MFK	AD7524MJ		

Advanced LinCMOS is a trademark of Texas Instruments Incorporated.

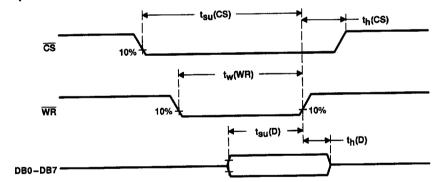


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## functional block diagram



### operating sequence



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## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)†

Supply voltage range, V <sub>DD</sub>	0.3 V to 17 V
Voltage between R <sub>FB</sub> and GND	+25 V
Digital input voltage range, V <sub>1</sub>	-0.3 V to Vnn+0.3 V
Heterence voltage range, V <sub>ref</sub>	+25 V
Peak digital input current, I <sub>1</sub>	10 пА
Operating free-air temperature range, T <sub>A</sub>	55°C to 125°C
Storage temperature range, T <sub>sto</sub>	65°C to 150°C
Case temperature for 60 seconds, T <sub>C</sub> : FK package	260°C
Lead temperature 1,6 mm (1/16 inch) from case for 60 seconds: J package	300°C

<sup>†</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

### recommended operating conditions

	V	V <sub>DD</sub> = 5 V			V <sub>DD</sub> = 15 V			
	MIN	NOM	MAX	MIN	NOM	MAX	UNIT	
Supply voltage, VDD	4.75	5	5.25	14.5	15	15.5	٧	
Reference voltage, V <sub>ref</sub>		±10			±10		٧	
High-level input voltage, VIH	2.4			13.5			V	
Low-level input volage, V <sub>IL</sub>			0.8			1.5	V	
CS setup time, t <sub>su(CS)</sub>	40			40			ns	
CS hold time, th(CS)	0	· · · · · · · · · · · · · · · · · · ·		0		-	пѕ	
Data bus input setup time, t <sub>Su(D)</sub>	25			25			ns	
Data bus input hold time, th(D)	10			10			ns	
Pulse duration, WR low, tw(WR)	40			40			ns	
Operating free-air temperature, TA	-55		125	-55		125	°C	

## AD7524M Advanced LinCMOS™ 8-BIT MULTIPLYING DIGITAL-TO-ANALOG CONVERTER

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## electrical characteristics over recommended operating free-air temperature range, $V_{ref}$ = 10 V, OUT1 and OUT2 at GND (unless otherwise noted)

			TTOT CONFIDENCE			٧	D = 5 V	·	۷۵	D = 15	V	UNIT
	PARAMETER		TEST CONDITIONS		MIN TYP MAX			MIN	TYP	MAX	01411	
				Full-range			10			10	μА	
lН	High-level input cu	rrent	VI = VDD	25°C			1			1	٠٠٠	
				Full-range			-10			-10	μА	
IIL	Low-level input cur	rent	V <sub>I</sub> = 0	25°C			-1			-1	, ,	
		OUT1	DB0-DB7 at 0, WR and CS at 0 V	Full-range			±400			±200		
	Output leakage		V <sub>ref</sub> = ±10 V	25°C			±50			±50	nA	
l <sub>pkg</sub>	current	OUT2	DB0-DB7 at V <sub>DD</sub> , WR and CS at 0	Full-range			±400			±200	15.	
			V <sub>ref</sub> = ±10 V	25°C			±50			±50		
		Quiescent	DB0-DB7 at VIHmin or VII	DB0-DB7 at VIHmin or VILmax			2			2	mA	
dal	Supply current		DD0 DD7 -+ 0 \/ -= \/	Full-range			500		<u> </u>	500	μΑ	
		Standby	DB0-DB7 at 0 V or VDD	25°C			100			100	, , , , , , , , , , , , , , , , , , ,	
	Supply voltage ser	nsitivity,		Full-range			0.16			0.04	%/%	
ksvs	∆gain/∆V <sub>DD</sub>		ΔV <sub>DD</sub> = 10%	25°C		0.002	0.02		0.001	0.02	ρF	
Ci	Input capacitance, DB0-DB7, WR, CS		V <sub>I</sub> = 0	V <sub>I</sub> = 0			5			5	рF	
	,	OUT1					30			30		
	Output	OUT2	DB0-DB7 at 0, WR and C	Sat 0 V			120			120	pF	
Co	capacitance	OUT1		. ==			120			120	"	
		OUT2	DB0-DB7 at V <sub>DD</sub> , WR an	d CS at 0 V			30			30		
	Reference input in (REF to GND)	npedance			5		20	5		20	kΩ	

# operating characteristics over recommended operating free-air temperature range, $V_{ref}$ = 10 V, OUT1 and OUT2 at GND (unless otherwise noted)

		V <sub>CC</sub> = 5 V		V <sub>DD</sub> = 15 V		UNIT	
PARAMETER	TEST CONDITIONS	MIN	MAX	MIN	MAX	0111	
Linearity error				±0.2		±0.2	%FSR
		Full range		±1.4		±0.6	%FSR
Gain error	See Note 1	25°C		±1		±0.5	70F3R
Settling time (to 1/2 LSB)	See Note 2			100		100	ns
Propagation delay from digital input to 90% of final analog output current	See Note 2			80		80	ns
	$V_{ref} = \pm 10 \text{ V } (100 \text{ kHz sinewave}),$	Full range		0.5		0.5	%FSR
Feedthrough at OUT1 or OUT2	WR and CS at 0, DB0-DB7 at 0 25			0.25		0.25	701 31
Temperature coefficient of gain	T <sub>A</sub> = 25°C to t <sub>min</sub> or t <sub>max</sub>			±0.004		±0.001	%FSR/°C

NOTES: 1. Gain error is measured using the internal feedback resistor. Nominal Full Scale Range (FSR) = V<sub>ref</sub> - 1 LSB.

2. OUT1 load =  $100 \Omega$ ,  $C_{ext} = \overline{13} pF$ ,  $\overline{WR}$  at 0 V,  $\overline{CS}$  at 0 V, DB0-DB7 at 0 V to VDD or VDD to 0 V.

### PRINCIPLES OF OPERATION

The AD7524M is an 8-bit multiplying D/A converter consisting of an inverted R-2R ladder, analog switches, and data input latches. Binary weighted currents are switched between the OUT1 and OUT2 bus lines, thus maintaining a constant current in each ladder leg independent of the switch state. The high-order bits are decoded and these decoded bits, through a modification in the R-2R ladder, control three equally weighted current sources. Most applications only require the addition of an external operational amplifier and a voltage reference.

The equivalent circuit for all digital inputs low is seen in Figure 1. With all digital inputs low, the entire reference current, I<sub>ref</sub>, is switched to OUT2. The current source 1/256 represents the constant current flowing through the termination resistor of the R-2R ladder, while the current source I<sub>lkg</sub> represents leakage currents to the substrate. The capacitances appearing at OUT1 and OUT2 are dependent upon the digital input code. With all digital inputs high, the off-state switch capacitance (30 pF maximum) appears at OUT2 and the on-state switch capacitance (120 pF maximum) appears at OUT1. With all digital inputs low, the situation is reversed as shown in Figure 1. Analysis of the circuit for all digital inputs high is similar to Figure 1; however, in this case, I<sub>ref</sub> would be switched to OUT1.

Interfacing the AD7524M D/A converter to a microprocessor is accomplished via the data bus and the  $\overline{\text{CS}}$  and  $\overline{\text{WR}}$  control signals. When  $\overline{\text{CS}}$  and  $\overline{\text{WR}}$  are both low, the AD7524M analog output responds to the data activity on the DB0–DB7 data bus inputs. In this mode, the input latches are transparent and input data directly affects the analog output. When either the CS signal or WR signal goes high, the data on the DB0–DB7 inputs are latched until the  $\overline{\text{CS}}$  and  $\overline{\text{WR}}$  signals go low again. When  $\overline{\text{CS}}$  is high, the data inputs are disabled regardless of the state of the  $\overline{\text{WR}}$  signal.

The AD7524M is capable of performing 2-quadrant or full 4-quadrant multiplication. Circuit configurations for 2-quadrant or 4-quadrant multiplication are shown in Figures 2 and 3. Input coding for unipolar and bipolar operation are summarized in Tables 1 and 2, respectively.

### PRINCIPLES OF OPERATION

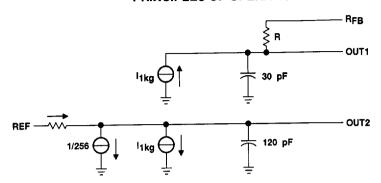


Figure 1. AD7524M Equivalent Circuit With All Digital Inputs Low

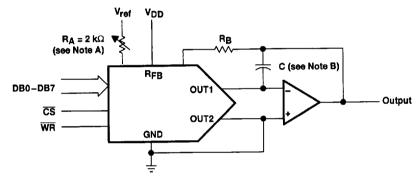


Figure 2. Unipolar Operation (2-Quadrant Multiplication)

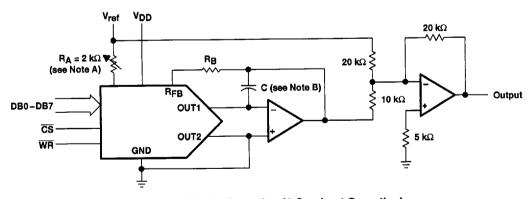


Figure 3. Bipolar Operation (4-Quadrant Operation)

NOTES: A.  $R_{\mbox{\scriptsize A}}$  and  $R_{\mbox{\scriptsize B}}$  used only if gain adjustment is required.

B. C phase compensation (10 - 15 pF) is required when using high-speed amplifiers to prevent ringing or oscillation.

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### PRINCIPLES OF OPERATION

Table 1. Unipolar Binary Code

DIGITAL (SEE N	INPUT IOTE 3)	ANALOG OUTPUT
MSB	LSB	
11111	1111	-V <sub>ref</sub> (255/256)
10000	0001	-V <sub>ref</sub> (129/256)
10000	0000	$-V_{ref}$ (128/256) = $-V_{ref}$ /2
0111	1111	-V <sub>ref</sub> (127/256)
00000	0001	-V <sub>ref</sub> (1/256)
00000	0000	0

NOTES: 3. LSB =  $1/256 \text{ (V}_{ref}$ ). 4. LSB =  $1/128 \text{ (V}_{ref}$ ).

Table 2. Bipolar (Offset Binary) Code

DIGITAL (SEE N	INPUT OTE 4)	ANALOG OUTPUT
MSB	LSB	
1000	0000 1111 0001	V <sub>ref</sub> (127/128) V <sub>ref</sub> (128) 0 -V <sub>ref</sub> (128) -V <sub>ref</sub> (127/128) -V <sub>ref</sub>

### microprocessor interfaces

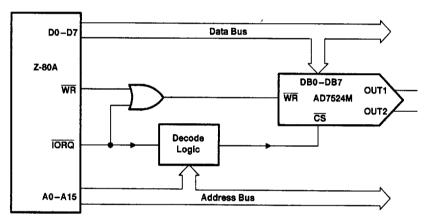


Figure 4. AD7524M-Z-80A Interface

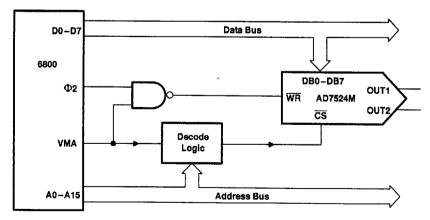


Figure 5. AD7524M-6800 Interface

### microprocessor interfaces (continued)

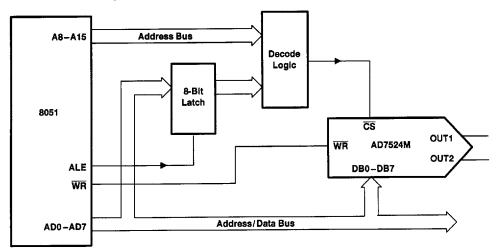


Figure 6. AD7524M-8051 Interface