

### **General Description**

The MAX2106 low-cost, direct-conversion tuner IC is designed for use in digital direct-broadcast satellite (DBS) television set-top box units and is a pin-for-pin upgrade for the MAX2104. Its direct-conversion architecture reduces system cost compared to devices with IF-based architectures. The MAX2106 directly tunes Lband signals to baseband using a broadband I/Q downconverter. The operating frequency range spans 925MHz to 2175MHz.

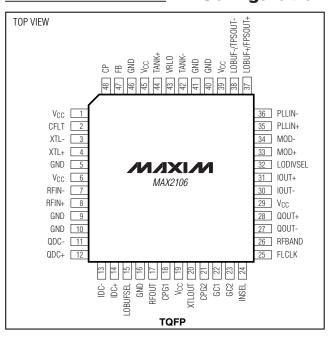
The IC includes a low-noise amplifier (LNA) with gain control, I and Q downconverting mixers, lowpass filters with gain and frequency control, a local oscillator (LO) buffer with a 90° quadrature network, and a chargepump-based phase-locked loop (PLL) for frequency control. The MAX2106 has an on-chip LO, requiring only an external varactor-tuned LC tank for operation. The LO's output drives the internal quadrature generator and has a buffer amplifier to drive off-chip circuitry. The MAX2106 comes in a 48-pin thin quad flat-pack package with exposed paddle (EP).

### **Applications**

U.S. DSS Set-Top Receivers European DVB-Compliant Systems Cellular Base Stations Wireless Local Loop

**Broadband Systems LMDS** Professional Receivers **VSAT** Microwave Links

### Pin Configuration



### Advantages Over MAX2104

- ♦ Improved Front End Achieves 10.2dB NF at 1550MHz
- ♦ Higher Input IIP3: 11.5dBm at 1550MHz
- ♦ Reduced Spurious Downconversion Products
- ♦ Capable of Using an External Synthesizer

#### **Features**

- ♦ Drop-In Replacement for MAX2104 Designs Requires Only Minor Software Upgrade and **Two External Resistor Value Changes**
- **♦** Complete Low-Cost Solution for DBS Direct Downconversion
- **♦** High Level of Integration Minimizes Component Count
- ♦ 1MBaud to 45MBaud Operation
- ♦ Selectable LO Buffer
- ♦ +5V Single-Supply Operation
- ♦ 925MHz to 2175MHz Input Frequency Range
- ♦ On-Chip Quadrature Generator, Dual-Modulus Prescaler (/32, /33)
- ♦ On-Chip Crystal Oscillator Amplifier
- ♦ PLL Phase Detector with Gain-Controlled Charge
- ♦ Input Levels: -25dBm to -68dBm per Carrier
- ♦ Over 50dB Gain Control Range
- ♦ Noise Figure = 10.2dB; IIP3 = +11.5dBm (at 1550MHz)
- ♦ Automatic Baseband Offset Correction

### **Ordering Information**

PART	TEMP RANGE	PIN-PACKAGE
MAX2106UCM	0°C to +85°C	48 TQFP-EP*
MAX2106UCM+	0°C to +85°C	48 TQFP-EP*

\*EP = Exposed paddle.

+Denotes lead-free package.

Functional Diagram appears at end of data sheet.

MIXIM

### **ABSOLUTE MAXIMUM RATINGS**

V <sub>CC</sub> to GND	0.3V to +7V
All Other Pins to GND	0.3V to (V <sub>CC</sub> + 0.3V)
RFIN+ to RFIN-, TANK+ to TANK-,	
IDC+ to IDC-, QDC+ to QDC	±2V
IOUT_, QOUT_ to GND Short-Circuit Du	uration10s
LOBUF+/PSOUT+, LOBUF-/PSOUT- Sh	ort-Circuit Duration10s
Continuous Current (any pin other than	V <sub>CC</sub> or GND)20mA

Continuous Power Dissipation (T <sub>A</sub> = +70°C	C)
48-Pin TQFP-EP (derate 27mW/°C abov	re +70°C)1.5W
Operating Temperature	
Junction Temperature	+150°C
Storage Temperature Range	
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### DC ELECTRICAL CHARACTERISTICS

 $(V_{CC} = +4.75 \text{V to } +5.25 \text{V}, V_{FB} = +2.4 \text{V}, C_{IOUT} = C_{QOUT} = 10 \text{pF}, f_{FLCLK} = 2 \text{MHz}, RFIN_ = \text{unconnected}, R_{IOUT} = R_{QOUT} = 10 \text{k}\Omega, V_{LOBUFSEL} = 0.5 \text{V}, V_{RFBAND} = V_{INSEL} = V_{CPG1} = V_{CPG2} = +2.4 \text{V}, V_{PLLIN} = V_{MOD} = +1.3 \text{V}, V_{PLLIN} = V_{MOD} = +1.1 \text{V}, T_{A} = +25 ^{\circ}\text{C}, unless otherwise noted}.$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
Operating Supply Voltage	Vcc		4.75		5.25	V	
Operating Supply Current	Icc			195	275	mA	
STANDARD DIGITAL INPUTS (	INSEL, CPG	i1, CPG2, LOBUFSEL, LODIVSEL)					
Input Voltage High	VIH		2.4			V	
Input Voltage Low	VIL				0.5	V	
Input Current	I <sub>IN</sub>		-15		10	μΑ	
RFBAND Input Current			-200		200	μΑ	
SLEW-RATE-LIMITED DIGITAL	INPUT (f <sub>LC</sub>	ELK)				1	
FLCLK Input Voltage High			1.85			V	
FLCLK Input Voltage Low					1.45	V	
FLCLK Input Current (Note 1)		R <sub>SOURCE</sub> = 50kΩ, V <sub>FLCLK</sub> = 1.65V	-1		1	μΑ	
DIFFERENTIAL DIGITAL INPU	TS (MOD+, I	MOD-, PLLIN+, PLLIN-)					
Common-Mode Input Voltage	VCMI		1.08	1.2	1.32	V	
Input Voltage Low		Referenced to V <sub>CMI</sub>			-100	mV	
Input Voltage High		Referenced to V <sub>CMI</sub>	100			mV	
Input Current (Note 1)			-5		5	μΑ	
DIFFERENTIAL DIGITAL OUTP	PUTS (LOBU	IF+/PSOUT+, LOBUF-/PSOUT-)					
Common-Mode Output Voltage	VCMO		2.16	2.4	2.64	V	
Output Voltage Low (Note 2)		Referenced to V <sub>CMO</sub> , LOBUFSEL ≤ 0.5V			-150	mV	
Output Voltage High (Note 2)		Referenced to V <sub>CMO</sub> , LOBUFSEL ≤ 0.5V	150			mV	
FREQUENCY SYNTHESIZER/L	O BUFFER						
		(V <sub>MOD+</sub> - V <sub>MOD-</sub> ) ≥ 200mV, LOBUFSEL ≤ 0.5V	32		32		
Prescaler Ratio		(VMOD+ - VMOD-) ≤ -200mV, LOBUFSEL ≤ 0.5V	33		33		
Prescaler Ralio		LOBUFSEL ≥ 2.4V, LODIVSEL ≤ 0.5V	2		2		
		LOBUFSEL ≥ 2.4V, LODIVSEL ≥ 2.4V	1		1		
Reference Divider Ratio			8		8		
XTLOUT Output DC Voltage				1.9		V	
		VCPG1 ≤ 0.5V, VCPG2 ≤ 0.5V	0.08	0.1	0.12		
Charge-Pump Output High		VCPG1 ≤ 0.5V, VCPG2 ≥ 2.4V	0.24	0.3	0.36	^	
Measured at FB		V <sub>CPG1</sub> ≥ 2.4V, V <sub>CPG2</sub> ≤ 0.5V	0.48	0.6	0.72	mA	
		V <sub>CPG1</sub> ≥ 2.4V, V <sub>CPG2</sub> ≥ 2.4V	1.44	1.8	2.16	1	

### DC ELECTRICAL CHARACTERISTICS (continued)

 $(\text{V}_{\text{CC}} = +4.75 \text{V to } +5.25 \text{V}, \text{V}_{\text{FB}} = +2.4 \text{V}, \text{C}_{\text{IOUT}} = \text{C}_{\text{QOUT}} = 10 \text{pF}, f_{\text{FLCLK}} = 2 \text{MHz}, \text{RFIN} = \text{unconnected}, \text{R}_{\text{IOUT}} = \text{R}_{\text{QOUT}} = 10 \text{k}\Omega, \\ \text{V}_{\text{LOBUFSEL}} = 0.5 \text{V}, \text{V}_{\text{RFBAND}} = \text{V}_{\text{INSEL}} = \text{V}_{\text{CPG1}} = \text{V}_{\text{CPG2}} = +2.4 \text{V}, \text{V}_{\text{PLIN}} = \text{V}_{\text{MOD}} = +1.3 \text{V}, \text{V}_{\text{PLLIN}} = \text{V}_{\text{MOD}} = +1.1 \text{V}, \text{T}_{\text{A}} = +25 ^{\circ}\text{C}, \\ \text{unless otherwise noted}.$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
		V <sub>CPG1</sub> ≤ 0.5V, V <sub>CPG2</sub> ≤ 0.5V	-0.12	-0.1	-0.08	
Charge-Pump Output Low		$V_{CPG1} \le 0.5V$ , $V_{CPG2} \ge 2.4V$	-0.36	-0.3	-0.24	mA
Measured at FB		V <sub>CPG1</sub> ≥ 2.4V, V <sub>CPG2</sub> ≤ 0.5V	-0.72	-0.6	-0.48	IIIA
		$V_{CPG1} \ge 2.4V$ , $V_{CPG2} \ge 2.4V$	-2.16	-1.8	-1.44	
Charge-Pump Output Current Matching Positive to Negative		Measured at FB	-5		5	%
Charge-Pump Output Leakage		Measured at FB	-25		25	nA
Charge-Pump Output Current Drive (Note 1)		Measured at CP	100			μA
ANALOG CONTROL INPUTS (G	C1 CC0\					
	C1, GC2)					
Input Current	IGC_	$V_{GC_{-}} = 1V \text{ to } 4V$	-50		50	μΑ
BASEBAND OUTPUTS (IOUT+,	IOUT-, QOL	JT+, QOUT-)				
Differential Output Voltage Swing		$R_L = 2k\Omega$ differential	1			Vp-p
Common-Mode Output Voltage (Note 1)			0.65		0.85	V
Offset Voltage (Note 1)			-50		50	mV

#### **AC ELECTRICAL CHARACTERISTICS**

(IC driven single-ended with RFIN- AC-terminated in 75 $\Omega$  to GND, V<sub>CC</sub> = +4.75V to +5.25V, V<sub>IOUT</sub> = V<sub>QOUT</sub> = 0.59Vp-p, C<sub>IOUT</sub> = C<sub>QOUT</sub> = 10pF,  $f_{LCLK}$  = 2MHz, R<sub>IOUT</sub> = R<sub>QOUT</sub> = 10k $\Omega$ , V<sub>LOBUFSEL</sub> = 0.5V, V<sub>RFBAND</sub> = V<sub>INSEL</sub> = V<sub>CPG1</sub> = V<sub>CPG2</sub> = +2.4V, V<sub>PLLIN+</sub> = V<sub>MOD+</sub> = +1.3V, V<sub>PLLIN-</sub> = V<sub>MOD-</sub> = +1.1V, T<sub>A</sub> = +25°C, unless otherwise noted. Typical values are at V<sub>CC</sub> = +5V.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS		
RF FRONT END	,								
RFIN_ Input Frequency Range	fRFIN_	Inferred by quadrature gain and phase-error test		925		2175	MHz		
RFIN_ Input Power for 0.59Vp-p		Single	VGC1 = V	GC2 = +4V (min gain)	-25			dBm	
Baseband Levels		carrier	VGC1 = V	'GC2 = +1V (max gain)			-68	dBm	
		Doen -	25dPm	$f_{LO} = 2175MHz$		10.5			
RFIN_ Input Third-Order Intercept Point (Note 3)		PRFIN_ = -25dBm per tone		$f_{LO} = 1550MHz$		11.5		dBm	
	IP3 <sub>RFIN</sub> _			$f_{LO} = 950MHz$		10.5			
	II SHFIN_	$\begin{array}{c} P_{RFIN\_} = -65 dBm \\ per tone \end{array} \qquad \begin{array}{c} f_{LO} = 2175 MHz \\ \hline f_{LO} = 1550 MHz \\ \hline f_{LO} = 950 MHz \end{array}$		$f_{LO} = 2175MHz$		-29			
				$f_{LO} = 1550MHz$		-26		dBm	
					-30				
RFIN_ Input Second-Order Intercept (Note 4)	IP2 <sub>RFIN</sub> _	_	$P_{RFIN} = -25dBm$ per tone, $f_{LO} = 951MHz$			17		dBm	
Output-Referred 1dB Compression Point (Note 5)	P1 <sub>dBOUT</sub>	P <sub>RFIN</sub> = -40dBm, signals within filter bandwidth			2		dBV		
Noise Figure	NF	f <sub>RFIN_</sub> = 1550MHz, V <sub>GC1</sub> = 1V, V <sub>GC2</sub> adjusted 0.59Vp-p baseband level		P <sub>RFIN</sub> _ = -65dBm		10.2		dB	
Troise rigule	INF			P <sub>RFIN</sub> _ = -25dBm		44.8		dB	

### **AC ELECTRICAL CHARACTERISTICS (continued)**

 $(\mathsf{RFIN} + \mathsf{IC} \text{ driven single-ended with RFIN- AC-terminated in } 75\Omega \text{ to GND, } V_{\mathsf{CC}} = +4.75V \text{ to } +5.25V, V_{\mathsf{IOUT}} = V_{\mathsf{QOUT}} = 0.59Vp-p, \\ \mathsf{CIOUT} = \mathsf{CQOUT} = 10pF, \mathsf{f}_{\mathsf{LCLK}} = 2\mathsf{MHz}, \mathsf{R}_{\mathsf{IOUT}} = \mathsf{RQOUT} = 10k\Omega, V_{\mathsf{LOBUFSEL}} = 0.5V, V_{\mathsf{RFBAND}} = V_{\mathsf{INSEL}} = V_{\mathsf{CPG1}} = V_{\mathsf{CPG2}} = +2.4V, \\ \mathsf{VPLLIN}_{\mathsf{H}} = \mathsf{V}_{\mathsf{MOD}_{\mathsf{H}}} = +1.3V, V_{\mathsf{PLLIN}_{\mathsf{H}}} = V_{\mathsf{MOD}_{\mathsf{H}}} = +1.1V, \mathsf{T}_{\mathsf{A}} = +25^{\circ}C, \mathsf{unless otherwise noted}. \mathsf{Typical values are at } V_{\mathsf{CC}} = +5V.)$ 

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
Dev Datawa Laga (Nata C)		$f_{RFIN}$ = 925MHz, $Z_{SOURCE}$ = 75 $\Omega$		+13		٩D	
R <sub>FIN</sub> + Return Loss (Note 6)		$f_{RFIN}$ = 2175MHz, $Z_{SOURCE}$ = 75 $\Omega$		+14		- dB	
LO 2nd Harmonic Rejection (Note 7)		Average level of VIOUT_, VQOUT_		32		dB	
LO Half Harmonic Rejection (Note 8)		Average level of V <sub>IOUT</sub> , VQOUT_		41.5		dB	
LO Leakage Power (Notes 6, 9)		Measured at RFIN+		-66		dBm	
RFOUT PORT (LOOPTHROUGH)						,	
		f = 925MHz		0.5			
RFIN+ to RFOUT Gain (Note 10)		f = 1550MHz		1.0		dB	
		f = 2175MHz		2.0			
		f = 925MHz		9			
RFOUT Output Third-Order Intercept Point (Note 10)		f = 1550MHz		7		dBm	
Tollit (Note 10)		f = 2175MHz		5			
		f = 925MHz		12.5			
RFOUT Noise Figure (Note 10)		f = 1550MHz		11		dB	
		f = 2175MHz		11			
RFOUT Return Loss (Notes 6, 10)		925MHz < f < 2175MHz, $Z_{LOAD} = 75Ω$		12		dB	
BASEBAND CIRCUITS	•						
Output Real Impedance (Note 1)		IOUT_, QOUT_			50	Ω	
Baseband Highpass -3dB Frequency (Note 1)		C <sub>IDC</sub> _ = C <sub>QDC</sub> _ = 0.22µF			750	Hz	
LPF -3dB Cutoff-Frequency Range (Note 1)		Controlled by FLCLK signal	8		33	MHz	
Baseband Frequency Response (Note 1)		Deviation from ideal 7th order, Butterworth, up to 0.7 × fc	-0.5		0.5	dB	
		f <sub>FLCLK</sub> = 0.5MHz, f <sub>C</sub> = 8MHz	-5.5		5.5		
LPF -3dB Cutoff-Frequency Accuracy (Note 1)		f <sub>FLCLK</sub> = 1.25MHz, f <sub>C</sub> = 19.3MHz	-10		10	%	
Accuracy (Note 1)		fFLCLK = 2.0625MHz, fC = 31.4MHz	10		10		
Ratio of In-Filter-Band to Out-of-Filter-Band Noise		f <sub>IN_BAND</sub> = 100Hz to 22.5MHz, f <sub>OUT_BAND</sub> = 67.5MHz to 112.5MHz		23		dB	
Quadrature Gain Error		Includes effects from baseband filters, measured at 125kHz baseband			1.2	dB	
Quadrature Phase Error		Includes effects from baseband filters, measured at 125kHz baseband			4	degrees	

### **AC ELECTRICAL CHARACTERISTICS (continued)**

(IC driven single-ended with RFIN- AC-terminated in 75 $\Omega$  to GND, V<sub>CC</sub> = +4.75V to +5.25V, V<sub>IOUT</sub> = V<sub>QOUT</sub> = 0.59Vp-p, C<sub>IOUT</sub> = C<sub>QOUT</sub> = 10pF,  $f_{LCLK}$  = 2MHz, R<sub>IOUT</sub> = R<sub>QOUT</sub> = 10k $\Omega$ , V<sub>LOBUFSEL</sub> = 0.5V, V<sub>RFBAND</sub> = V<sub>INSEL</sub> = V<sub>CPG1</sub> = V<sub>CPG2</sub> = +2.4V, V<sub>PLLIN+</sub> = V<sub>MOD+</sub> = +1.3V, V<sub>PLLIN-</sub> = V<sub>MOD-</sub> = +1.1V, T<sub>A</sub> = +25°C, unless otherwise noted. Typical values are at V<sub>CC</sub> = +5V.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
SYNTHESIZER						
XTLOUT Output Voltage Swing		Load = 10pF II 10k $\Omega$ , fxTLOUT = 6MHz	0.75	1	1.5	Vp-p
Crystal Frequency Range (Note 1)			4		7.26	MHz
MOD+, MOD- Setup Time (Note 1)	tsum	Figure 1	7			ns
MOD+, MOD- Hold Time (Note 1)	tHM	Figure 1	0			ns
LOCAL OSCILLATOR						
LO Tuning Range (Note 11)			590		1180	MHz
LO Buffer Output Voltage (Note 1)		V <sub>LOBUFSEL</sub> ≥ 2.4V, f <sub>LO</sub> = 925 MHz + 2175MHz	70			V <sub>RMS</sub>
		At 1kHz offset, f <sub>LO</sub> = 2175MHz		-60		
LO Phase Noise (Notes 6, 12)		At 10kHz offset, f <sub>LO</sub> = 2175MHz		-75		dBc/Hz
		At 100kHz offset, fLO = 2175MHz		-96		
RFIN+ to LO Input Isolation (Note 9)		f <sub>RFIN</sub> = 2175MHz		58		dB

- Note 1: Minimum and maximum values are guaranteed by design and characterization over supply voltage.
- **Note 2** Driving differential load of  $10k\Omega$  II 15pF.
- Note 3: Two signals are applied to RFIN\_ at f<sub>LO</sub> 100MHz and f<sub>LO</sub> 199MHz. V<sub>GC2</sub> = 1V, V<sub>GC1</sub> is set so that the baseband outputs are at 590mVp-p. IM products are measured at baseband outputs but are referred to RF inputs.
- **Note 4:** Two signals are applied to RFIN\_ at 1200MHz and 2150MHz. V<sub>GC2</sub> = 1V, V<sub>GC1</sub> is set so that the baseband outputs are at 590mVp-p. IM products are measured at baseband outputs but are referred to RF inputs.
- **Note 5:** P<sub>RFIN</sub> = -40dBm so that front-end IM contributions are minimized.
- Note 6: Using L64733/L64734 demo board from LSI Logic.
- Note 7: Downconverted level, in dBc, of carrier present at  $f_{LO} \times 2$ ,  $f_{LO} = 1180$ MHz,  $f_{VCO} = 590$ MHz,  $f_{VCO} = 5900$ MHz,  $f_{VCO} =$
- Note 8: Downconverted level, in dBc, of carrier present at fo / 2, fLO = 2175MHz, fvcO = 1087.5MHz, V<sub>RFBAND</sub> = 2.4V.
- Note 9: Leakage is dominated by board parasitics.
- Note 10:  $V_{CPG1} = V_{CPG2} = V_{RFBAND} = V_{INSEL} = 0.5V$ ,  $f_{LCLK} = 0.5MHz$ .
- **Note 11:** Guaranteed by design and characterization over supply and temperature.
- Note 12: Measured at tuned frequency with PLL locked. PLL loop bandwidth = 3kHz. All phase noise measurements assume tank components have a Q > 50.

# \_\_\_\_\_Pin Description

PIN	NAME	FUNCTION
1, 6, 19, 29, 39, 45	Vcc	$V_{CC}$ Power-Supply Input. Connect each pin to a +5V ±5% low-noise supply. Bypass each $V_{CC}$ pin to the nearest GND with a ceramic chip capacitor.
2	CFLT	External Bypass for Internal Bias. Bypass this pin with a 0.1µF ceramic chip capacitor to GND.
3	XTL-	Inverting Input to Crystal Oscillator. Consult crystal manufacturer for circuit loading requirements.
4	XTL+	Noninverting Input to Crystal Oscillator. Consult crystal manufacturer for circuit loading requirements.
5, 9, 10, 16, 40, 41, 46	GND	Ground. Connect each of these pins to a solid ground plane. Use multiple vias to reduce inductance where possible.
7	RFIN-	RF Inverting Input. Bypass RFIN- with 47pF capacitor in series with a 75 $\Omega$ resistor to GND.
8	RFIN+	RF Noninverting Input. Connect to $75\Omega$ source with a 47pF ceramic chip capacitor.
11	QDC-	Baseband Offset Correction. Connect a 0.22µF ceramic chip capacitor from QDC- to QDC+ (pin 12).
12	QDC+	Baseband Offset Correction. Connect a 0.22µF ceramic chip capacitor from QDC+ to QDC- (pin 11).
13	IDC-	Baseband Offset Correction. Connect a 0.22µF ceramic chip capacitor from IDC- to IDC+ (pin 14).
14	IDC+	Baseband Offset Correction. Connect a 0.22µF ceramic chip capacitor from IDC+ to IDC- (pin 13).
15	LOBUFSEL	Local Oscillator Buffer Select. Connect to GND to select DIV32/33 prescaler output; connect V <sub>CC</sub> to DIV1 to select DIV2 LO buffer output.
17	RFOUT	Buffered RF Output. Enabled when INSEL is low.
18	CPG1	Charge-Pump Gain Select. High-impedance digital input. Sets the charge-pump output scaling. See DC Electrical Characteristics for available gain settings.
20	XTLOUT	Buffered Crystal Oscillator Output
21	CPG2	Charge-Pump Gain Select. High-impedance digital input. Sets the charge-pump output scaling. See DC Electrical Characteristics for available gain settings.
22	GC1	Gain Control Input for RF Front End. High-impedance analog input, with an input range of +1V to +4V. See AC Electrical Characteristics for transfer function.
23	GC2	Gain Control Input for Baseband Signals. High-impedance analog input, with an input range of +1V to +4V. See AC Electrical Characteristics for transfer function.
24	INSEL	Loopthrough Mode Enable. High-impedance digital input. Drive low to enable the RFOUT buffer and disable the LO converters. Drive high for normal tuner operation.
25	FLCLK	Baseband Filter Cutoff Adjust. Connect to a slew-rate-limited clock source. See AC Electrical Characteristics for transfer function.
26	RFBAND	RF Input Band Select Input. Drive high to enable 1680 MHz to 2175 MHz band. Leave unconnected to enable 1180 MHz to 1680 MHz band. Connect to GND to enable 925 MHz to 1180 MHz band.
27	QOUT-	Baseband Quadrature Output. Connect to inverting input of high-speed ADC.
28	QOUT+	Baseband Quadrature Output. Connect to noninverting input of high-speed ADC.
30	IOUT-	Baseband In-Phase Output. Connect to inverting input of high-speed ADC.
31	IOUT+	Baseband In-Phase Output. Connect to noninverting input of high-speed ADC.
32	LODIVSEL	LO Buffer Divider Ratio Input. Drive high to enable divide-by-one LO buffer output. Connect to GND to enable divide-by-two buffer output.
33	MOD+	PECL Modulus Control. A PECL high on MOD+ sets the dual-modulus prescaler to divide by 32. A PECL logic low sets the divide ratio to 33. Drive with a differential PECL signal in conjunction with MOD- (pin 34).

# Pin Description (continued)

PIN	NAME	FUNCTION
34	MOD-	PECL Modulus Control. A PECL low on MOD- sets the dual-modulus prescaler to divide by 32. A PECL logic high sets the divide ratio to 33. Drive with a differential PECL signal in conjunction with MOD+ (pin 33).
35	PLLIN+	PECL Phase-Locked Loop Input. Drive with a differential PECL signal in conjunction with PLLIN- (pin 36).
36	PLLIN-	PECL Phase-Locked Loop Input. Drive with a differential PECL signal in conjunction with PLLIN+ (pin 35)
37	LOBUF+/ PSOUT+	LOBUFSEL = GND: PECL Prescaler Output. Differential output of the dual-modulus prescaler. Used in conjunction with PSOUT Requires PECL-compatible termination. LOBUFSEL= $V_{CC}$ : $50\Omega$ LO buffer noninverting output.
38	LOBUF-/ PSOUT-	LOBUFSEL = GND: PECL Prescaler Output. Differential output of the dual-modulus prescaler. Used in conjunction with PSOUT+. Requires PECL-compatible termination. LOBUFSEL = $V_{CC}$ : 50 $\Omega$ LO buffer inverting output.
42	TANK-	LO Tank Oscillator Input. Connect to an external LC tank with varactor tuning.
43	VRLO	LO Internal Regulator. Bypass with a 1000pF ceramic chip capacitor to GND.
44	TANK+	LO Tank Oscillator Input. Connect to an external LC tank with varactor tuning.
47	FB	Feedback Input for Loop Filter
48	CP	Voltage Drive Output. Control of external charge-pump transistor.

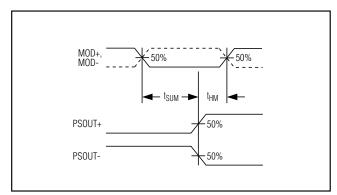
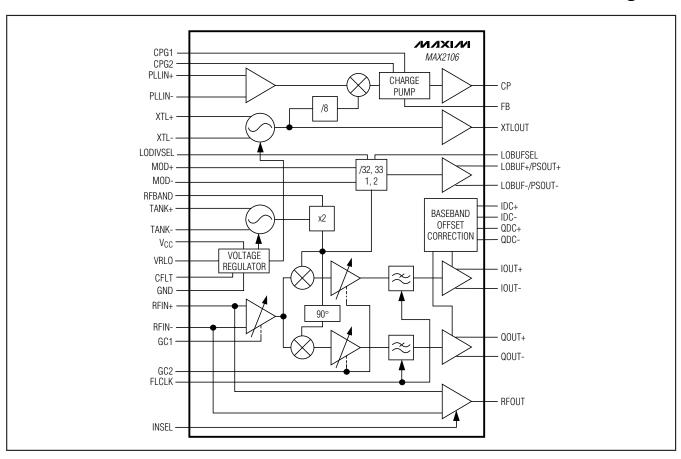


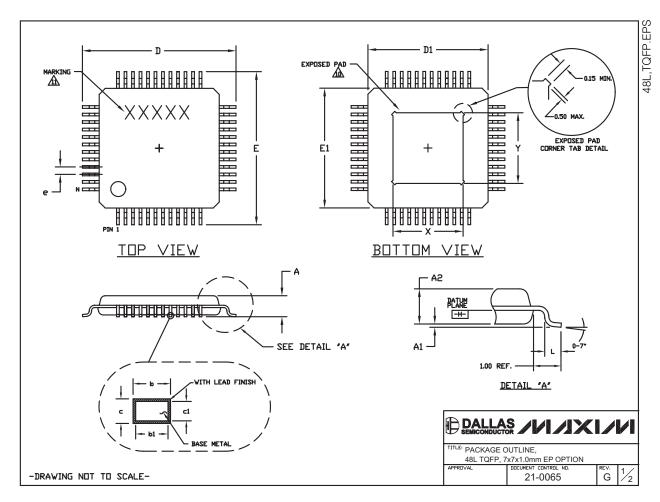
Figure 1. Modulus Control Timing Diagram

# **Functional Diagram**



### **Package Information**

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to <a href="https://www.maxim-ic.com/packages">www.maxim-ic.com/packages</a>.)



### Package Information (continued)

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to www.maxim-ic.com/packages.)

- NOTES:

  1. ALL DIMENSIONS AND TOLERANCING CONFORM TO ANSI Y14.5-1982.

  2. DATUM PLANE [-H-] IS LOCATED AT MOLD PARTING LINE AND COINCIDENT WITH LEAD, WHERE LEAD EXITS PLASTIC BODY AT BOTTOM OF PARTING LINE.

  3. DIMENSIONS D1 AND E1 DO NOT INCLUDE MOLD PROTRUSION.
  ALLOWABLE MOLD PROTRUSION IS 0.25 MM ON D1 AND E1 DIMENSIONS.
  4. THE TOP OF PACKAGE IS SMALLER THAN THE BOTTOM OF PACKAGE BY 0.15 MILLIMETERS.
  5. DIMENSION & DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.08 MM TOTAL IN EXCESS OF THE & DIMENSION AT MAXIMUM MATERIAL CONDITION.
  6. ALL DIMENSIONS ARE IN MILLIMETERS.
  7. THIS OUTLINE CONFORMS TO JEDEC PUBLICATION 95 REGISTRATION MS-026, VARIATION ABA-HD.
  8. LEADS SHALL BE COPLANAR WITHIN 0.08 MM.
  9. EXPOSED DIE PAD SHALL BE COPLANAR WITH BOTTOM OF PACKAGE WITHIN 2 MILS (.05 MM).
  10. DIMENSIONS X & Y APPLY TO EXPOSED PAD (EP) VERSIONS ONLY. SEE INDIVIDUAL PRODUCT DATASHEET TO DETERMINE IF A PRODUCT USES EXPOSED PAD PACKAGE.

  12. NUMBER OF LEADS SHOWN ARE FOR REFERENCE ONLY.

S	JEDEC VARIATION						
M B C L	ABC-HD						
ן נ	MIN.	NDM.	MAX.				
Α	N/L	~~	1.20				
A <sub>1</sub>	0.05	0.10	0.15				
Az	0.95	1.00	1.05				
D	8.90	9.00	9.10				
D <sub>1</sub>	6.90	7.00	7.10				
Ε	8.90	9.00	9.10				
E <sub>1</sub>	6.90	7.00	7.10				
L	0.45	0.60	0.75				
N		48					
e		0.50 BSC.					
b	0.17	0.22	0.27				
b1	0.17	0.20	0.23				
С	0.09		0.20				
<b>c</b> 1	0.09		0.16				

		EXPOSED PAD VARIATIONS						
DICE		Х			Υ			
PKG. CODE	MIN.	NDM.	MAX.	MIN.	NDM.	MAX.		
C48E-7	3.70	4.00	4.30	3.70	4.00	4.30		
C48E-8	4.70	5.00	5.30	4.70	5.00	5.30		
C48E-10	3.70	4.00	4.30	3.70	4.00	4.30		

DALLAS ////XI//I

21-0065

LEI PACKAGE OUTLINE.

48L TQFP, 7x7x1.0mm EP OPTION
VAL DOCUMENT CONTROL NO.

G

-DRAWING NOT TO SCALE-

Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.

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# **MAX2106**

# **Part Number Table**

#### Notes:

- 1. See the MAX2106 QuickView Data Sheet for further information on this product family or download the MAX2106 full data sheet (PDF, 192kB).
- 2. Other options and links for purchasing parts are listed at: <a href="http://www.maxim-ic.com/sales">http://www.maxim-ic.com/sales</a>.
- 3. <u>Didn't Find What You Need?</u> Ask our applications engineers. Expert assistance in finding parts, usually within one business day.
- 4. Part number suffixes: T or T&R = tape and reel; += RoHS/lead-free; #= RoHS/lead-exempt. More: See <u>full data sheet</u> or <u>Part Naming Conventions</u>.
- 5. \* Some packages have variations, listed on the drawing. "PkgCode/Variation" tells which variation the product uses.

Part Number	Free Sample	Buy Direct	Package: TYPE PINS SIZE  DRAWING CODE/VAR *	Temp	RoHS/Lead-Free? Materials Analysis
MAX2106UCM-D			TQFP;48 pin;7x7x1 mm Dwg: 21-0065G (PDF) Use pkgcode/variation: C48E-10*	0C to +85C	RoHS/Lead-Free: No Materials Analysis
MAX2106UCM-TD			TQFP;48 pin;7x7x1 mm Dwg: 21-0065G (PDF) Use pkgcode/variation: C48E-10*	0C to +85C	RoHS/Lead-Free: No Materials Analysis
MAX2106UCM+D			TQFP;48 pin;7x7x1 mm Dwg: 21-0065G (PDF) Use pkgcode/variation: C48E+10*	0C to +85C	RoHS/Lead-Free: Yes Materials Analysis

MAX2106UCM+TD

TQFP;48 pin;7x7x1 mm

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