## High Efficiency Small Packaged Step-up DC/DC Converter

No. EA-317-181205

## OUTLINE

The RP402x is a high efficiency step-up DC/DC converter with synchronous rectifier. The device can start up with low voltage of typically 0.7 V which is ideal for the applications powered by either one-cell or two-cell alkaline, nickel-metal-hydride (NiMH) or one-cell Lithium-ion (Li+) batteries.

Internally, the RP402x consists of an oscillator, a reference voltage unit with soft start, a chip enable circuit, an error amplifier, phase compensation circuits, a slope circuit, a PWM control circuit, a start-up circuit, a PWM/VFM mode control circuit, internal switches and protection circuits.

The RP402x is employing synchronous rectification for improving the efficiency or rectification by replacing diodes with built-in switching transistors. Using synchronous rectification not only increases circuit performance but also allows a design to reduce parts count.

The RP402x is available in either internally fixed output voltage type or adjustable output voltage type. The RP402xxxxx is the internally fixed output voltage type. The RP402x00xx is the adjustable output voltage type, which allows output voltages that range from 1.8 V to 5.5 V via an external divider resistor.

The RP402x provides the forced PWM control and the PWM/VFM auto switching control. Either one of these can be selected by inputting a signal to the MODE pin. The forced PWM control switches at fixed frequency rate in low output current in order to reduce noise. Likewise, the PWM/VFM auto switching control automatically switches from PWM mode to VFM mode in low output current in order to achieve high efficiency. The RP402N is available in the PWM/VFM auto switching control. However, the RP402N is also available in the forced PWM control as a custom-designed IC ${ }^{(1)}$.

The RP402x has a soft-start time of typically 0.5 ms .
The RP402x features the complete output disconnect shutdown option and the input-to-output bypass shutdown option. The RP402xxxxA/ B/ E/ F incorporates the complete output disconnect shutdown option, which allows the output to be disconnected from the input. The RP402xxxxC/ D/G/H incorporates the input-to-output bypass shutdown option, which allows the output to be connected to the input.

The RP402x is protected against damage by a short-current protection, an over-voltage lockout, an over voltage protection, an anti-ringing switch and a latch-type protection. An anti-ringing switch prevents the occurrence of noise when an inductor current reaches a discontinuous mode. The RP402x provides optional Latch function with current limit detection which can turn off the power in case the limit values are detected for a fixed time and current limit circuit controls peak inductor currents in every clock. The latch-type protection can be released by switching the CE pin from high to low while the power is turned on.

The RP402x is offered in a compact 5-pin SOT23-5 package or a 8-pin DFN(PLP)2020-8 package.

[^0]
## RP402x

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## FEATURES

- Low Voltage Start-up Typ. 0.7 V
- Input Voltage Range

Fixed Output Voltage Type: 0.6 V to 4.8 V
Adjustable Output Voltage Type: 0.6 V to 4.6 V
 $90 \%\left(1 \mathrm{~mA} / 5.0 \mathrm{~V}, \mathrm{~V}_{\mathrm{IN}}=3.6 \mathrm{~V}, 25^{\circ} \mathrm{C}\right)$


- Lx Driver ON Resistance ........................................... 20 PMOS: $0.20 \Omega$ (Vout $=5.0 \mathrm{~V}, 25^{\circ} \mathrm{C}$ )

- Output Voltage Range................................................... Output Voltage Type: 1.8 V to $5.5 \mathrm{~V}, 0.1 \mathrm{~V}$ step

Adjustable Output Voltage Type: 1.8 V to 5.5 V (recommended)


- OVP Detector Threshold .......................................... 6.0 V
- Lx Peak Current Limit ................................................... A
- Latch Protection Delay Time..........................Typ. 3.3 ms (RP402Kxx1x, RP402Nxx1x)

Typ. 4.1 ms (RP402Kxx2x)

- Soft-start Time

Typ. 0.5 ms

- EMI Suppression (Built-in Anti-ringing Switch) (RP402Kxx1x, RP402Nxx1x)
- Voltage Regulation at VIN $>$ Vout
- Zero Input Complete Shutdown at $\mathrm{V}_{\mathrm{IN}}=0 \mathrm{~V}$
- Input-to-Output Bypass Shutdown Option at CE $=L(R P 402 x x x x C / D / G / H)$
- Ceramic Capacitor Capable
- Package .......................................................................


## APPLICATIONS

- MP3 Players, PDA
- Digital Still Cameras
- LCD Bias Supplies
- Portable Blood Pressure Meter
- Wireless Handset
- GPS
- USB-OTG
- HDMI


## SELECTION GUIDE

The package type, the set output voltage, the PWM control type, the shutdown option, the MODE pin option, and the latch function are user-selectable options.

## Selection Guide

| Product Name | Package | Quantity per Reel | Pb Free | Halogen Free |
| :--- | :---: | :---: | :---: | :---: |
| RP402Kxx\#\$-TR | DFN(PLP)2020-8 | $5,000 \mathrm{pcs}$ | Yes | Yes |
| RP402Nxx\#\$-TR-FE | SOT-23-5 | $3,000 \mathrm{pcs}$ | Yes | Yes |

$x x$ : Specify the set output voltage $\left(V_{S E T}\right)$.
00: Adjustable Output Voltage Type (1.8 V to 5.5 V , recommended voltage range)
xx: Fixed Output Voltage Type ( 1.8 V to 5.5 V , adjustable in 0.1 V step)
Please note: SOT-23-5 package is only available with fixed output voltage type.
\#: Specify the PWM control type.
1: Normal PWM operation
2: Forced PWM operation
\$: Specify the combination of the shutdown option, the MODE pin option and the latch function.

| Version | Shutdown Options at CE = L | MODE Pin | Latch Function |
| :---: | :---: | :---: | :---: |
| A | Complete Output Disconnect | Yes | Yes |
| B | Complete Output Disconnect | Yes | No |
| C | Input-to-Output Bypass | Yes | Yes |
| D | Input-to-Output Bypass | Yes | No |
| E | Complete Output Disconnect | No | Yes |
| F | Complete Output Disconnect | No | No |
| G | Input-to-Output Bypass | No | Yes |
| H | Input-to-Output Bypass | No | No |

Please refer to Selection Guide Table on the next page for detailed information.

## RP402x

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## Selection Guide Table

| Package | Output Voltage Type | \#\$ | Shutdown Option at $C E=L$ | MODE Pin Function |  | PWM Controlling Method | Latch Function |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | $\begin{gathered} \text { MODE } \\ \text { Pin } \end{gathered}$ | Power Controlling Method |  |  |
| DFN(PLP)2020-8 | Fixed Output Voltage Type | 1A | Complete Output Disconnect | Yes | "H": Normal PWM Control, <br> "L": PWM/VFM Auto Switching Control | Normal PWM | Yes |
|  |  | 2A |  |  | "H": Forced PWM Control Note: "H" recommended | Forced PWM |  |
|  |  | 1B |  |  | "H": Normal PWM Control, "L": PWM/VFM Auto Switching Control | Normal PWM | No |
|  |  | 2B |  |  | "H": Forced PWM Control Note: "H" recommended | Forced PWM |  |
|  |  | 1C | Input-toOutput Bypass |  | "H": Normal PWM Control, <br> "L": PWM/VFM Auto Switching Control | Normal PWM | Yes |
|  |  | 1D |  |  | "H": Normal PWM Control, <br> "L": PWM/VFM Auto Switching Control | Normal PWM | No |
|  | Adjustable <br> Output <br> Voltage <br> Type | 1A | Complete <br> Output <br> Disconnect |  | "H": Normal PWM Control, <br> "L": PWM/VFM Auto Switching Control | Normal PWM | Yes |
|  |  | 2A |  |  | "H": Forced PWM Control Note: "H" recommended | Forced PWM |  |
|  |  | 1B |  |  | "H": Normal PWM Control, <br> "L": PWM/VFM Auto Switching Control | Normal PWM | No |
|  |  | 2B |  |  | "H": Forced PWM Control Note: "H" recommended | Forced PWM |  |
|  |  | 1C | Input-to- <br> Output <br> Bypass |  | " H ": Normal PWM Control, <br> "L": PWM/VFM Auto Switching Control | Normal PWM | Yes |
|  |  | 1D |  |  | "H": Normal PWM Control, "L": PWM/VFM Auto Switching Control | Normal PWM | No |
| SOT-23-5 | Fixed Output Voltage Type | 1E | Complete Output Disconnect | No | PWM/VFM Auto Switching Control | Normal PWM | Yes |
|  |  | 1F |  |  | PWM/VFM Auto Switching Control | Normal PWM | No |
|  |  | 1G | Input-to- <br> Output <br> Bypass |  | PWM/VFM Auto Switching Control | Normal PWM | Yes |
|  |  | 1H |  |  | PWM/VFM Auto Switching Control | Normal PWM | No |

## BLOCK DIAGRAMS

*1 This Bvpass Switch is included in the RP402KxxxC / D only
${ }^{* 2}$ This Latch Timer is included in the RP402KxxxA / C only.


## RP402Kxxxx Block Diagram

RP402K00xx Block Diagram


## RP402x

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${ }^{* 1}$ This Bypass Switch is included in the RP402NxxxG/ H only
${ }^{*}$ 2 This Latch Timer is included in the RP402NxxxE/ G only.


RP402Nxxxx Block Diagram

## PIN DESCRIPTION



RP402K [DFN(PLP)2020-8] Pin Configurations


RP402N (SOT-23-5) Pin Configurations

* The tab on the bottom of the package enhances thermal performance and is electrically connected to GND (substrate level). It is recommended that the tab be connected to the ground plane on the board, or otherwise be left floating.


## RP402Kxxxx Pin Description

| Pin No. | Symbol | Description |
| :---: | :---: | :--- |
| 1 | MODE | Mode Pin $^{(1)}$ |
| 2 | NC | No Connection |
| 3 | GND | Ground Pin |
| 4 | Lx | Internal NMOS Switch Drain Pin |
| 5 | Vout | Output Pin |
| 6 | VIN $^{2}$ | Power Supply Pin |
| 7 | NC | No Connection |
| 8 | CE | Chip Enable Pin, Active-high |

## RP402K00xx Pin Description

| Pin No. | Symbol | Description |
| :---: | :---: | :--- |
| 1 | MODE | MODE Pin ${ }^{(1)}$ |
| 2 | NC | No Connection |
| 3 | GND | Ground Pin |
| 4 | Lx | Internal NMOS Switch Drain Pin |
| 5 | $V_{\text {out }}$ | Output Pin |
| 6 | VIN | Power Supply Pin |
| 7 | VFB | Feedback Input Pin for Setting Output Voltage |
| 8 | CE | Chip Enable Pin, Active-high |

RP402Nxx1x Pin Description

| Pin No. | Symbol | Description |
| :---: | :---: | :--- |
| 1 | Lx | Internal NMOS Switch Drain Pin |
| 2 | GND | Ground Pin |
| 3 | CE | Chip Enable Pin, Active-high |
| 4 | VIN | Power Supply Pin |
| 5 | Vout $^{2}$ | Output Pin |

(1) MODE Pin = "H" is recommended for RP402Kxx2x.

## ABSOLUTE MAXIMUM RATINGS

| Symbol | Parameter |  | Rating | Unit |
| :---: | :---: | :---: | :---: | :---: |
| Vin | Vin Pin Voltage |  | -0.3 to 6.5 | V |
| Vout | Vout Pin Voltage |  | -0.3 to 7.0 | V |
| VLx | Lx Pin Voltage |  | -0.3 to 6.5 | V |
| $\mathrm{V}_{\text {ce }}$ | CE Pin Voltage |  | -0.3 to 6.5 | V |
| $\mathrm{V}_{\text {FB }}$ | $\mathrm{V}_{\text {FB }}$ Pin Voltage (RP402K00xx only) |  | -0.3 to 6.5 | V |
| $\mathrm{V}_{\text {mode }}$ | MODE Pin Voltage (RP402Kxxxx only) |  | -0.3 to 6.5 | V |
| PD | Power Dissipation ${ }^{(1)}$ (JEDEC STD. 51-7) | DFN(PLP)2020-8 | 1800 | mW |
|  |  | SOT-23-5 | 660 |  |
| Tj | Junction Temperature Range |  | -40 to 125 | ${ }^{\circ} \mathrm{C}$ |
| Tstg | Storage Temperature Range |  | -55 to 125 | ${ }^{\circ} \mathrm{C}$ |

## ABSOLUTE MAXIMUM RATINGS

Electronic and mechanical stress momentarily exceeded absolute maximum ratings may cause the permanent damages and may degrade the life time and safety for both device and system using the device in the field. The functional operation at or over these absolute maximum ratings is not assured.

## RECOMMENDED OPERATING CONDITIONS

Recommended Operating Conditions

| Symbol | Parameter | Rating | Unit |
| :---: | :--- | :---: | :---: |
| $\mathrm{V}_{\mathrm{IN}}$ | Input Voltage | 0.6 to 4.8 | V |
| Ta | Operating Temperature | -40 to 85 | ${ }^{\circ} \mathrm{C}$ |

## RECOMMENDED OPERATING CONDITIONS

All of electronic equipment should be designed that the mounted semiconductor devices operate within the recommended operating conditions. The semiconductor devices cannot operate normally over the recommended operating conditions, even if when they are used over such conditions by momentary electronic noise or surge. And the semiconductor devices may receive serious damage when they continue to operate over the recommended operating conditions.

[^1]
## ELECTRICAL CHARACTERISTICS

The specifications surrounded by $\square$ are guaranteed by design engineering at $-40^{\circ} \mathrm{C} \leq \mathrm{Ta} \leq 85^{\circ} \mathrm{C}$.

RP402xxxxx Electrical Characteristics (Not applicable to RP402K00xx) ( $\mathrm{Ta}=25^{\circ} \mathrm{C}$ )

| Symbol | Parameter |  | Condition | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Vstart | Start-up Voltage |  | $\mathrm{R} \mathrm{L}=5.5 \mathrm{k} \Omega$ |  | 0.7 | 0.8 | V |
| Vhold | Hold-on Voltage after start-up ${ }^{(1)}$ |  | $\mathrm{R} \mathrm{L}=5.5 \mathrm{k} \Omega$ | 0.6 |  |  | V |
| Vovıo | OVLO Voltage |  | - | 4.8 | 5.1 |  | V |
| Vovp | OVP Voltage |  | - |  | 6.0 |  | V |
| IDD1 | Quiescent Current 1 |  | $\begin{aligned} & \mathrm{V}_{\text {IN }}=\mathrm{V}_{\text {SET }}-0.4 \mathrm{~V}, \\ & \mathrm{~V}_{\text {OUT }}=0.95 \times \mathrm{V}_{\text {SET }} \end{aligned}$ |  | 1.6 |  | mA |
| IdD2 | Quiescent Current $\mathbf{2}^{(2)}$ |  | $\begin{aligned} & \hline \mathrm{V}_{\text {IN }}=\mathrm{V}_{\text {SET }}-0.4 \mathrm{~V}, \\ & \mathrm{~V}_{\text {OUT }}=\mathrm{V}_{\text {SET }}+0.2 \mathrm{~V} \end{aligned}$ |  | 21 | 37 | $\mu \mathrm{A}$ |
| Istandby | Standby Current | RP402xxxxA/ B/ E/ F | $\begin{aligned} & V_{I N}=4.8 \mathrm{~V}, \text { VOUT }=0 \mathrm{~V}, \\ & V_{C E}=0 \mathrm{~V} \end{aligned}$ |  | 0.2 | 1.0 | $\mu \mathrm{A}$ |
|  |  | RP402xxxxC/ D/ G/ H | $\mathrm{V}_{\text {IN }}=4.8 \mathrm{~V}, \mathrm{~V}_{\text {Ce }}=0 \mathrm{~V}$ |  | 1.2 | 2.5 |  |
| Vout | Output Voltage |  | $\mathrm{V}_{\mathrm{IN}}=\mathrm{V}_{\text {CE }}=1.5 \mathrm{~V}$ | x0.985 |  | x1.015 | V |
| $\Delta$ Vout $I \Delta T a$ | Output-Voltage <br> Temperature Coefficient |  | $-40^{\circ} \mathrm{C} \leq \mathrm{Ta} \leq 85^{\circ} \mathrm{C}$ |  | $\pm 50$ |  | $\underset{{ }^{\circ} \mathrm{C}}{\mathrm{p}}$ |
| fosc | Switching <br> Frequency | RP402xxx1x | $\begin{aligned} & V_{\text {IN }}=1.5 \mathrm{~V}, \\ & \text { Vout }=0.95 \times \mathrm{V}_{\text {SET }} \end{aligned}$ | 1080 <br> 1020 | 1200 | 1320 <br> 1380 <br> 1100 | kHz |
|  |  | RP402xxx2x |  | 900 <br> 850 | 1000 | $\begin{aligned} & 1100 \\ & 1150 \end{aligned}$ |  |
| Ronn | NMOS ON Resistance ${ }^{(1)}$ |  | Vout $=5.0 \mathrm{~V}$ |  | 0.20 |  | $\Omega$ |
| Ronp | PMOS ON Resistance ${ }^{(1)}$ |  | Vout $=5.0 \mathrm{~V}$ |  | 0.20 |  | $\Omega$ |
| Ісен | CE "H" Input Current |  | $\begin{aligned} & \text { VIN }=4.8 \mathrm{~V}, \\ & \mathrm{~V}_{\text {OUT }}=\mathrm{V}_{\text {CE }}=5 \mathrm{~V} \end{aligned}$ |  |  | 0.5 | $\mu \mathrm{A}$ |
| Icel | CE "L" Input Current |  | $\begin{aligned} & \mathrm{V}_{\text {IN }}=4.8 \mathrm{~V}, \\ & \mathrm{~V}_{\text {OUT }}=5 \mathrm{~V}, \mathrm{~V}_{\text {CE }}=0 \mathrm{~V} \end{aligned}$ | -0.5 |  |  | $\mu \mathrm{A}$ |
| Імоден | MODE "H" Input Current ${ }^{(3)}$ | RP402xxx1x | $\mathrm{V}_{\text {IN }}=4.8 \mathrm{~V}$, |  |  | 0.5 | $\mu \mathrm{A}$ |
|  |  | RP402xxx2x | $\begin{aligned} & \mathrm{V}_{\text {CE }}=0 \mathrm{~V}, \\ & \mathrm{~V}_{\text {MODE }}=5.5 \mathrm{~V} \end{aligned}$ |  |  | 72 |  |
| Imodel | MODE "L" Input Current ${ }^{(3)}$ |  | $\begin{aligned} & \mathrm{V}_{\mathrm{IN}}=4.8 \mathrm{~V}, \mathrm{~V}_{\mathrm{CE}}=\mathrm{V}_{\mathrm{MODE}}= \\ & 0 \mathrm{~V} \end{aligned}$ | -0.5 |  |  | $\mu \mathrm{A}$ |
| ILxh | Lx "H" Leakage Current |  | $\begin{aligned} & \hline V_{\text {IN }}=V_{\text {OUT }}=V_{L X}=4.8 \mathrm{~V}, \\ & V_{C E}=0 \mathrm{~V} \end{aligned}$ |  |  | 0.5 | $\mu \mathrm{A}$ |
| ILxL | Lx "L"Leakage Current |  | $\begin{aligned} & \text { Vоит }=5 \mathrm{~V}, \mathrm{~V}_{\mathrm{LX}}=0 \mathrm{~V}, \\ & \mathrm{~V}_{\mathrm{CE}}=0 \mathrm{~V} \end{aligned}$ |  |  | 0.5 | $\mu \mathrm{A}$ |
| IlXPeak | Lx Limit Current ${ }^{(4)}$ |  |  | 1.3 | 1.5 |  | A |

All test items listed under ELECTRICAL CHARACTERISTICS are done under the pulse load condition ( $\mathrm{Tj} \approx \mathrm{Ta}=25^{\circ} \mathrm{C}$ ).

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## ELECTRICAL CHARACTERISTICS (continued)

The specifications surrounded by $\square$ are guaranteed by design engineering at $-40^{\circ} \mathrm{C} \leq \mathrm{Ta} \leq 85^{\circ} \mathrm{C}$.

| $\left(\mathrm{Ta}=25^{\circ} \mathrm{C}\right)$ |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Symbol | Parameter |  |  | Condition | Min. | Typ. | Max. | Unit |
| $V_{\text {ceh }}$ | CE "H" Input Voltage |  |  |  | 0.7 |  |  | V |
| $V_{\text {cel }}$ | CE "L" Input Voltage |  |  |  |  |  | 0.3 | V |
| Vmodeh | MODE "H" Input Voltage ${ }^{(1)}$ |  |  |  | 1.0 |  |  | V |
| Vmodel | MODE "L" Input Voltage ${ }^{(1)}$ |  |  |  |  |  | 0.4 | V |
| Maxduty | Oscillator Maximum Duty Cycle |  |  | $\begin{aligned} & \mathrm{V}_{\text {IN }}=1.5 \mathrm{~V} \text {, Vout }=0.95 \\ & \mathrm{x} \mathrm{~V}_{\text {SET }} \end{aligned}$ | 80 | 88 | 95 | \% |
| tstart | Soft-start Time ${ }^{(2)}$ |  |  | Measures the time when <br> $\mathrm{V}_{\text {CE }}=0 \mathrm{~V}$ to 1.5 V , <br> $V_{\text {OUT }}=V_{\text {SET }} \times 0.95$ | 0.25 | 0.5 | 0.70 | ms |
| tprot | Protection Time ${ }^{(3)}$ | Delay | RP402xxx1x |  | 2.7 | 3.3 | 3.9 | ms |
|  |  |  | RP402xxx2x |  | 3.5 | 4.1 | 4.7 |  |
| Rona | Anti-ringing Switch ON Resistance ${ }^{(4)}$ |  |  | $\mathrm{V}_{\text {IN }}=2.5 \mathrm{~V}$, $\mathrm{V}_{\text {OUT }}=3.3 \mathrm{~V}$ |  | 100 |  | $\Omega$ |
| Ronb | Bypass Sw Resistance |  | RP402xxxxC/ D/ G/ H | $\begin{aligned} & \mathrm{V}_{\text {IN }}=3.0 \mathrm{~V}, \\ & \mathrm{~V}_{\text {OUt }}=0 \mathrm{~V} \end{aligned}$ |  | 160 |  | $\Omega$ |
| linzero | Vin Zero Current |  |  | $\mathrm{V}_{\text {IN }}=0 \mathrm{~V}$, $\mathrm{V}_{\text {out }}=5.5 \mathrm{~V}$ |  | 0.1 | 1.0 | $\mu \mathrm{A}$ |

All test items listed under ELECTRICAL CHARACTERISTICS are done under the pulse load condition ( $\mathrm{Tj} \approx \mathrm{Ta}=25^{\circ} \mathrm{C}$ ).

[^3]Electrical Characteristics by Differenct Output Voltage

| Product Name | $\mathrm{V}_{\text {out }}\left(\mathrm{Ta}=25^{\circ} \mathrm{C}\right.$ ) |  |  |
| :---: | :---: | :---: | :---: |
|  | Min. | Typ. | Max. |
| RP402x18xx | 1.773 | 1.800 | 1.827 |
| RP402x19xx | 1.872 | 1.900 | 1.929 |
| RP402x20xx | 1.970 | 2.000 | 2.030 |
| RP402x21xx | 2.069 | 2.100 | 2.132 |
| RP402x22xx | 2.167 | 2.200 | 2.233 |
| RP402x23xx | 2.266 | 2.300 | 2.335 |
| RP402x24xx | 2.364 | 2.400 | 2.436 |
| RP402x25xx | 2.463 | 2.500 | 2.538 |
| RP402x26xx | 2.561 | 2.600 | 2.639 |
| RP402x27xx | 2.660 | 2.700 | 2.741 |
| RP402x28xx | 2.758 | 2.800 | 2.842 |
| RP402x29xx | 2.857 | 2.900 | 2.944 |
| RP402x30xx | 2.955 | 3.000 | 3.045 |
| RP402x31xx | 3.054 | 3.100 | 3.147 |
| RP402x32xx | 3.152 | 3.200 | 3.248 |
| RP402x33xx | 3.251 | 3.300 | 3.350 |
| RP402x34xx | 3.349 | 3.400 | 3.451 |
| RP402x35xx | 3.448 | 3.500 | 3.553 |
| RP402x36xx | 3.546 | 3.600 | 3.654 |
| RP402x37xx | 3.645 | 3.700 | 3.756 |
| RP402x38xx | 3.743 | 3.800 | 3.857 |
| RP402x39xx | 3.842 | 3.900 | 3.959 |
| RP402x40xx | 3.940 | 4.000 | 4.060 |
| RP402x41xx | 4.039 | 4.100 | 4.162 |
| RP402x42xx | 4.137 | 4.200 | 4.263 |
| RP402x43xx | 4.236 | 4.300 | 4.365 |
| RP402x44xx | 4.334 | 4.400 | 4.466 |
| RP402x45xx | 4.433 | 4.500 | 4.568 |
| RP402x46xx | 4.531 | 4.600 | 4.669 |
| RP402x47xx | 4.630 | 4.700 | 4.771 |
| RP402x48xx | 4.728 | 4.800 | 4.872 |
| RP402x49xx | 4.827 | 4.900 | 4.974 |
| RP402x50xx | 4.925 | 5.000 | 5.075 |
| RP402x51xx | 5.024 | 5.100 | 5.177 |
| RP402x52xx | 5.122 | 5.200 | 5.278 |
| RP402x53xx | 5.221 | 5.300 | 5.380 |
| RP402x54xx | 5.319 | 5.400 | 5.481 |
| RP402x55xx | 5.417 | 5.500 | 5.582 |

## RP402x

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## ELECTRICAL CHARACTERISTICS (continued)

The specifications surrounded by $\square$ are guaranteed by design engineering at $-40^{\circ} \mathrm{C} \leq \mathrm{Ta} \leq 85^{\circ} \mathrm{C}$.

RP402K00xx Electrical Characteristics
( $\mathrm{Ta}=25^{\circ} \mathrm{C}$ )

| Symbol | Parameter |  | Condition | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| VIN | Input Voltage |  |  |  |  | 4.6 | V |
| Vstart | Start-up Voltage |  | $\mathrm{RL}=5.5 \mathrm{k} \Omega$ |  | 0.7 | 0.8 | V |
| Vhold | Hold-on Voltage after start-up ${ }^{(1)}$ |  | $\mathrm{RL}=5.5 \mathrm{k} \Omega$ | 0.6 |  |  | V |
| Vovio | OVLO Voltage |  |  | 4.6 | 5.1 |  | V |
| Vovp | OVP Voltage |  |  |  | 6.0 |  | V |
| IdD1 | Quiescent Current 1 |  | $\begin{aligned} & \mathrm{V}_{\mathrm{IN}}=3 \mathrm{~V}, \mathrm{~V}_{\text {out }}=5 \mathrm{~V}, \\ & \mathrm{~V}_{\mathrm{FB}}=0.6 \mathrm{~V} \end{aligned}$ |  | 1.6 |  | mA |
| IDD2 | Quiescent Current ${ }^{(2)}$ |  | $\begin{aligned} & \mathrm{V}_{\mathrm{IN}}=4.8 \mathrm{~V}, \mathrm{~V}_{\text {OUT }}=5.5 \mathrm{~V}, \\ & \mathrm{~V}_{F B}=2.0 \mathrm{~V}, \mathrm{~V}_{\text {MODE }}=0 \mathrm{~V} \end{aligned}$ |  | 21 | 37 | $\mu \mathrm{A}$ |
| Istandby | Standby Current | RP402KxxxA/ B | $\begin{aligned} & \mathrm{V}_{\text {IN }}=4.8 \mathrm{~V}, \mathrm{~V}_{\text {OUT }}=0 \mathrm{~V}, \\ & \mathrm{~V}_{\mathrm{CE}}=0 \mathrm{~V} \end{aligned}$ |  | 0.2 | 1.0 | $\mu \mathrm{A}$ |
|  |  | RP402KxxxC/ D | $\mathrm{V}_{\text {IN }}=4.8 \mathrm{~V}, \mathrm{~V}_{\text {ce }}=0 \mathrm{~V}$ |  | 1.2 | 2.5 |  |
| $V_{\text {FB }}$ | Feedback Voltage |  | $\mathrm{V}_{\text {IN }}=3.0 \mathrm{~V}, \mathrm{~V}_{\text {OUt }}=5 \mathrm{~V}$ | 0.985 | 1.00 | 1.015 | V |
| $\begin{aligned} & \hline \Delta \mathrm{V}_{\mathrm{FB}} \\ & / \Delta \mathrm{Ta} \end{aligned}$ | Output Voltage Temperature Coefficient |  | $-40^{\circ} \mathrm{C} \leq \mathrm{Ta} \leq 85^{\circ} \mathrm{C}$ |  | $\pm 50$ |  | $\begin{aligned} & \mathrm{ppm} \\ & { }^{\circ} \mathrm{C} \end{aligned}$ |
| fosc | Switching Frequency | RP402K001x | $\begin{aligned} & V_{\text {IN }}=3.0 \mathrm{~V}, \mathrm{~V}_{\text {OUT }}=3.3 \mathrm{~V}, \\ & \mathrm{~V}_{\text {FB }}=0.6 \mathrm{~V} \end{aligned}$ | 1080 <br> 1020 | 1200 | 1320 | kHz |
|  |  | RP402K002x |  | 900 <br> 850 | 1000 | 1100 |  |
| Ronn | NMOS ON Resistance ${ }^{(1)}$ |  | $\mathrm{V}_{\text {OUT }}=5.0 \mathrm{~V}$ |  | 0.20 |  | $\Omega$ |
| Ronp | PMOS ON Resistance ${ }^{(1)}$ |  | Vout $=5.0 \mathrm{~V}$ |  | 0.20 |  | $\Omega$ |
| Icen | CE "H" Input Current |  | $\begin{aligned} & \mathrm{V}_{\text {IN }}=4.8 \mathrm{~V}, \mathrm{~V}_{\text {OUT }}=\mathrm{V}_{\mathrm{CE}}= \\ & 5.5 \mathrm{~V} \end{aligned}$ |  |  | 0.5 | $\mu \mathrm{A}$ |
| Icel | CE "L" Input Current |  | $\begin{aligned} & \mathrm{V}_{\text {IN }}=4.8 \mathrm{~V} \text {, } \mathrm{V}_{\text {OUT }}=5 \mathrm{~V}, \\ & \mathrm{~V}_{\text {CE }}=0 \mathrm{~V} \end{aligned}$ | -0.5 |  |  | $\mu \mathrm{A}$ |
| Імоden | MODE "H" Input Current | RP402K001x | $\begin{aligned} & \mathrm{V}_{\text {IN }}=4.8 \mathrm{~V}, \mathrm{~V}_{\text {MODE }}=5.5 \mathrm{~V}, \\ & \mathrm{~V}_{\mathrm{CE}}=0 \mathrm{~V} \end{aligned}$ |  |  | 0.5 | A |
|  |  | RP402K002x |  |  |  | 72 | $\mu \mathrm{A}$ |
| Imodel | MODE "H" Input Current |  | $\begin{aligned} & \mathrm{V}_{\text {IN }}=4.8 \mathrm{~V}, \mathrm{~V}_{\mathrm{CE}}=\mathrm{V}_{\mathrm{MODE}}= \\ & 0 \mathrm{~V} \end{aligned}$ | -0.5 |  |  | $\mu \mathrm{A}$ |
| ILxH | Lx "H" Leakage Current |  | $\begin{aligned} & \mathrm{V}_{\text {IN }}=\mathrm{V}_{\text {OUT }}=\mathrm{V}_{\mathrm{LX}}=4.8 \mathrm{~V}, \\ & \mathrm{~V}_{\mathrm{CE}}=0 \mathrm{~V} \end{aligned}$ |  |  | 0.5 | $\mu \mathrm{A}$ |
| ILxL | Lx "L"Leakage Current |  | $\begin{aligned} & V_{\text {OUT }}=5.0 \mathrm{~V}, \mathrm{~V}_{\mathrm{LX}}=0 \mathrm{~V}, \\ & \mathrm{~V}_{\mathrm{CE}}=0 \mathrm{~V} \end{aligned}$ |  |  | 0.5 | $\mu \mathrm{A}$ |
| ILXPEAK | Lx Limit Current ${ }^{(3)}$ |  |  | 1.3 | 1.5 |  | A |

All test items listed under ELECTRICAL CHARACTERISTICS are done under the pulse load condition ( $\mathrm{Tj} \approx \mathrm{Ta}=25^{\circ} \mathrm{C}$ ).

[^4]
## ELECTRICAL CHARACTERISTICS (continued)

The specifications surrounded by $\qquad$ are guaranteed by design engineering $-40^{\circ} \mathrm{C} \leq \mathrm{Ta} \leq 85^{\circ} \mathrm{C}$.

RP402K00xx Electrical Characteristics
( $\mathrm{Ta}=25^{\circ} \mathrm{C}$ )

| Symbol | Parameter |  | Condition | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $V_{\text {Ceh }}$ | CE "H" Input Voltage |  |  | 0.7 |  |  | V |
| $V_{\text {cel }}$ | CE "L" Input Voltage |  |  |  |  | 0.3 | V |
| Vmodeh | MODE "H" Input Voltage |  |  | 1.0 |  |  | V |
| Vmodel | MODE "L" Input Voltage |  |  |  |  | 0.4 | V |
| Maxduty | Oscillator Maximum Duty Cycle |  | $\begin{aligned} & \mathrm{V}_{\text {IN }}=3.0 \mathrm{~V}, \mathrm{~V}_{\text {out }}=3.3 \mathrm{~V}, \\ & \mathrm{~V}_{\text {FB }}=0.6 \mathrm{~V} \end{aligned}$ | 80 | 88 | 95 | \% |
| tstart | Soft-start Time ${ }^{(1)}$ |  | Measures the time when <br> $V_{\text {out }}=3.3 \mathrm{~V}$, <br> $\mathrm{V}_{\mathrm{CE}}=0 \mathrm{~V}$ to 1.5 V , <br> Vout $=3.13 \mathrm{~V}$ | 0.25 | 0.5 | 0.70 | ms |
| tprot | Protection Delay Time ${ }^{(2)}$ | RP402K001x | - | 2.7 | 3.3 | 3.9 | ms |
|  |  | RP402K002x | - | 3.5 | 4.1 | 4.7 | ms |
| Rona | Anti-ringing Switch ON Resistance ${ }^{(3)}$ |  | $\mathrm{V}_{\text {IN }}=2.5 \mathrm{~V}$, $\mathrm{V}_{\text {OUT }}=3.3 \mathrm{~V}$ |  | 100 |  | $\Omega$ |
| Ronb | Bypass Switch ON Resistance ${ }^{(4)}$ | RP402KxxxC/ D | $\begin{aligned} & \mathrm{V}_{\text {IN }}=3.0 \mathrm{~V}, \\ & \mathrm{~V}_{\text {OUT }}=0 \mathrm{~V} \end{aligned}$ |  | 160 |  | $\Omega$ |
| linzero | VIN Zero Current |  | $\mathrm{V}_{\text {IN }}=0 \mathrm{~V}$, $\mathrm{V}_{\text {Out }}=5.5 \mathrm{~V}$ |  | 0.1 | 1.0 | $\mu \mathrm{A}$ |

All test items listed under ELECTRICAL CHARACTERISTICS are done under the pulse load condition ( $\mathrm{Tj} \approx \mathrm{Ta}=25^{\circ} \mathrm{C}$ ).

[^5]
## RP402x

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## THEORY OF OPERATION

## Forced PWM Control Type (RP402xx2A/ B)

While normal PWM control type prevents the reverse inductor current at light load, forced PWM control type makes the inductor current reverse in order to eliminate the discontinuous current period. Therefore, even at light load or when the voltage difference between input and output is less, forced PWM control type can provide PWM operation without bursting.
Normal PWM
IL

Forced PWM


## Operating Waveform of Normal PWM/ Forced PWM Control Type

There is a case that forced PWM control performs burst operation without PWM operation because of the conditions of use. The conditions which cause burst operation are various and differ in set output voltage, input voltage, ambient temperature and load current.

Please note that forced PWM control type decreases the efficiency at light load and does not include antiringing switch. The graph below indicates the typical operational maximum input voltage of forced PWM control type.
RP402Kxx1x: MODE = "H" (Normal PWM), RP402Kxx2x: (Forced PWM)




## MODE Pin (RP402K only)

When setting the MODE pin "high" of RP402K, it is recommended to connect the MODE pin "high" with the Vout pin to ensure stability. Please note that a current flows through the pull-down resistor to consume power even in a standby state (CE="low") as the MODE pin is pulled down by an internal resistor when the Mode pin "high" is connected with VIN pin. Since RP402Kxx2A/B have only Forced PWM control type, the MODE pin "high" fixing is recommended to ensure safety, but also connecting the MODE pin to GND or using it in the open state are other options.

## Bypass Mode Application Example (RP402xxxxC/ D/ G/ H)

The RP402xxxxC/ D/G/H is available in bypass mode when CE = L. The shown below is the application example of the device in bypass mode. In this application, when the main system is not in sleep, the RP402xxxxC/ D/G/H is set to active state to supply power to the main system and RTC. When the main system is in sleep, the RP 402xxxxC/ D/G/H is set to standby state to supply power to RTC in bypass mode. Using the device in the bypass mode can reduce the power loss and the consumption of battery. Also, using the device in bypass mode can eliminate external components for short-circuit protection.


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## Regulation Operation at $\mathbf{V}_{\text {IN }}>\mathrm{V}_{\text {OUT }}$

The RP402x regulates the output voltage to the set output voltage even when the input voltage is higher than the set output voltage. Please note that this regulation operation decreases the efficiency and the maximum output current driving ability. The maximum output current driving ability can be different due to the set output voltage, the input voltage and the ambient temperature.
The following is the switching condition (Typ.) from step-up operation to the step-down regulation.
$\mathrm{V}_{\text {IN }} \leq$ Vout 150 mV : Step-down regulation $\rightarrow$ Step-up operation
VIN $>$ Vout-100 mV: Step-up operation $\rightarrow$ Step-down regulation


## Output Voltage Setting for RP402K00xx

The RP402K00xx can set the output voltage freely by the external divider resistors using the following equation.

Output Voltage $=\mathrm{V}_{\mathrm{FB}} \times(\mathrm{R} 1+\mathrm{R} 2) / \mathrm{R} 2 \quad\left(\mathrm{~V}_{\mathrm{FB}}=1.0 \mathrm{~V}\right)$

## Zero Input Complete Shutdown at $\mathrm{V}_{\mathrm{IN}}=0 \mathrm{~V}$

The RP402x provides a zero input complete shutdown function that allows the device to shut down the output when $\mathrm{V}_{\mathrm{IN}}=0 \mathrm{~V}$ or $\mathrm{V}_{\mathrm{IN}}=$ open. This function protects against reverse current flow from Vout to $\mathrm{V}_{\mathrm{IN}}$ when a voltage is applied to the Vout pin while $\mathrm{V}_{\mathrm{IN}}=0 \mathrm{~V}$ or $\mathrm{V}_{\mathrm{IN}}=$ open.

## Overcurrent Protection

The RP402x incorporates a Lx peak current limit circuit as the overcurrent protection circuit which controls the duty of $L_{x}$ when the $L_{x}$ peak current (ILхРЕАк) reaches typically 1.5 A.

## Latch Type Protection (RP402xxxxA/ C/E/G)

The RP402xxxxA/ C/ E/ G provides a latch type protection circuit to latch the power MOSFET to the off state in order to stop the DC/DC operation. To release the latch type protection, switch the CE pin from high to low once and switch it back to high while the power is turned on. Please note that the Lx peak current (lıxpeak) and the protection delay time (tprot) are easily affected by the self-heating or heat radiation efficiency. The large reduction in input voltage ( $\mathrm{V}_{\mathrm{IN}}$ ) or the unstable input voltage caused by short-circuit may affect the protection operation or protection delay time.

## Short-circuit Protection

The RP402x provides a short-circuit protection which stops the switching operation when a short circuit is detected. After a consecutive fixed period of the short-circuit state, the device performs a restart with soft-start operation. RP402xxxxA/ C/E/G latches the power in a stop state when the input voltage becomes lower than typically 1.6 V and it is short-circuited.

## Overvoltage Protection

The RP402x provides an overvoltage lockout (OVLO) circuit for monitoring the input pin voltage and an overvoltage protection (OVP) circuit for monitoring the output pin voltage. These circuits stops the switching operation when an overvoltage is detected. If the output voltage is dropped below the set output voltage when OVLO is released, the output voltage will be boosted to the set output voltage.

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## OUTPUT CURRENT AND SELECTION OF EXTERNAL COMPONENTS

## Operation of Step-up DC/DC Converter and Output Current

## Basic Circuit



## Discontinuous Mode



Continuous Mode


A PWM control type step-up DC/DC converter has two operation modes characterized by the continuity of inductor current: discontinuous current mode and continuous current mode.

The voltage applied to the inductor L , when transistor is ON , is described as " V IN". So, the current is described as "VIN $\mathrm{Xt} / \mathrm{L}$ ".

Therefore, the electric power (Pon) supplied from the input side, while transistor is ON , is described as follows:
$P_{\text {ON }}=\int_{0}^{\text {ton }} V \mathbb{N}^{2} \times t / L d t$
Equation 1

In step-up circuit, power source supplies the electric power (Poff) even while transistor is OFF. The input current supplied by power source while transistor is OFF is described as "(Vout - Vin) x $t / L$ ". Therefore, the electric power Poff is described as follows:
$P_{\text {off }}=\int_{0}^{\mathrm{tf}} \mathrm{V}$ IN $\times\left(\right.$ Vout $\left.-\mathrm{V}_{\text {IN }}\right) \times \mathrm{t} / \mathrm{L} d t$
Equation 2
The time of which the inductance $L$ releases the saved energy is described as " tf ". Therefore, the average electric power ( Pav ) in a cycle is described as follows:
$\mathrm{P}_{\mathrm{AV}}=1 /($ ton + toff $) \times\left\{\int_{0}^{\text {ton }} V_{\mathbb{N}}{ }^{2} \times t / L d t+\int_{0}^{\mathrm{tf}} \mathrm{V}_{\mathbb{N}} \times\left(\mathrm{Vout}^{2}-\mathrm{V}_{\mathbb{I}}\right) \times \mathrm{t} / \mathrm{L} d t\right\} \ldots \ldots . . . . . . . . . . .$. Equation 3

In PWM control, when "tf = toff", the inductor current becomes continuous, so the switching regulator operation turns into continuous current mode. The current deviation between On time and Off time is equal under steadystate condition of continuous current mode as follows:
$V_{\text {IN }} x$ ton $/ L=\left(V_{\text {OUT }}-V_{\text {IN }}\right) x$ toff $/ L$
Equation 4

The electric power $\left(\mathrm{P}_{\mathrm{AV}}\right)$ is equal to the output voltage ( $\mathrm{V}_{\text {Out }} \xi$ lout $)$. Therefore, lout is as follows:
lout $=$ fosc $\times$ VIN $^{2} \times \operatorname{ton}^{2} /\left\{2 \times L\left(\right.\right.$ Vout $\left.\left.-V_{\text {IN }}\right)\right\}=\operatorname{VIN}^{2} x$ ton $/(2 \times L \times$ Vout $)$
Equation 5

When lout becomes more than $\mathrm{V}_{\mathrm{IN}} \times$ ton $\times$ toff $/(2 \times \mathrm{L} \times$ (ton + toff $)$ ), the inductor current becomes continuous, so the switching regulator operation turns into continuous current mode. The continuous inductor current is described as Iconst, so lout is described as follows:
lout $=$ fosc $\times \mathrm{V}_{\text {IN }}{ }^{2} \times$ ton $^{2} /\left(2 \times L\left(V_{\text {OUT }}-V_{\text {IN }}\right)\right)+\mathrm{V}_{\text {IN }} \times$ Iconst $/ \mathrm{V}_{\text {OUT }}$
Equation 6

The peak current (ILmax) flowing through the inductor is described as follows:

ILmax $=$ Iconst $+\mathrm{V}_{\mathrm{IN}} \mathrm{x}$ ton $/ \mathrm{L}$
Equation 7

Put Equation 4 into Equation 6 to solve ILmax. ILmax is described as follows:

ILmax $=\mathrm{V}_{\text {out }} / \mathrm{V}_{\text {IN }} \times$ lout $+\mathrm{V}_{\text {IN }} x$ ton $/(2 \mathrm{xL})$
Equation 8

However, ton $=\left(1-V_{\text {IN }} / V_{\text {OUt }}\right) /$ fosc. The peak current is more than lout.

Please consider ILmax when setting conditions of input and output, as well as selecting the external components. The peak current in the discontinuous current mode in Equation 7 can be calculated by Iconst $=0$.

Please note: The above calculation formulas are based on the ideal operation of the device in continuous mode. The loss caused by the external components and the built-in Lx switch are not included. Please use the peak current in Equation 8 as a reference when selecting an inductor.

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## TIMING CHART

## Soft-start Operation and Latch-type Protection Operation


*1 Only for RP402xxx1x (MODE = "L")
*2 Only for RP402xxxxA/ C/ E/G

## < Start-up >

When CE is changed from "L" to "H", DC/DC converter starts up the operation. The RP402x has Low-Boost mode which can start up with low voltage such as 0.7 V . The DC/DC boosts up with Low-Boost mode until the output voltage reaches to typically 1.6 V . When the output voltage becomes more than or equal to typically 1.6 V , the soft-start operation starts in order to control inrush current. The DC/DC boosts up the output voltage until it reaches to the setting output voltage.

Please note: During Low-Boost mode, the oscillator frequency is dropped, so the step-up ability is low compared to the normal operation mode. Please pay attention to the step-up ratio and the load current. Softstart time depends on "set output voltage", "input voltage", "ambient temperature", and "load current".


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## APPLICATION INFORMATION



RP402Kxxxx Typical Application (Fixed Output Voltage Type)


RP402K00xx Typical Application (Adjustable Output Voltage Type)


RP402Nxx1x Typical Application (Fixed Output Voltage Typ)

## Recommended Components

| Symbol | Descriptions |
| :---: | :---: |
| L | VLF403215MT-2R2M, $2.2 \mu \mathrm{H}$, TDK VLS3012HBX-2R2M, TDK NRS5020T2R2NMGJ, TAIYO YUDEN |
| C1 (CIN) | GRM188R60J106ME47, $10 \mu \mathrm{~F}$, Murata |
| C2 (Cout) | GRM188R60J106ME47, $10 \mu \mathrm{~F} \times 2$, Murata <br> As for the fixed output voltage type (RP402x50xx), $10 \mu \mathrm{~F} \times 1$ can be used if the mounting area is limited. |
| Cspd | The speedup capacitor ( $\mathrm{C}_{\text {SPD }}$ ) is required for the adjustable output voltage type. <br> Connect Cspd in parallel with the output resistor (R1). <br> To calculate the CspD value, the following equation can be used: $f=1 /\left(2 \pi \times C_{s P D} \times R 1\right)$ <br> Adjust the CsPD value to make the oscillator frequency (f) approximately 20 kHz . <br> For example, Vout $=5.0 \mathrm{~V}, \mathrm{R} 1=2 \mathrm{M} \Omega, \mathrm{R} 2=500 \mathrm{k} \Omega$ and $\mathrm{Cspd}=4 \mathrm{pF}$. <br> The R1 and R2 values are calculated based on the operation efficiency under a light load, therefore R1 and R2 are having high-resistance values. The feedback voltage ( $\mathrm{V}_{\mathrm{FB}}$ ) can be affected by noise. To stabilize the device operation, decrease the R1 and R2 values. |
| RSPD | The speedup resistor ( $\mathrm{R}_{\mathrm{SPD}}$ ) is required for the adjustable output voltage type. <br> Using RsPD can prevent the deterioration of the characteristics due to noise. <br> If there's a possibility of generation of a spike noise, use an approximately $1 \mathrm{k} \Omega \mathrm{R}_{\text {SPD }}$. |

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## TECHNICAL NOTES

The performance of a power source circuit using this device is highly dependent on a peripheral circuit. A peripheral component or the device mounted on PCB should not exceed its rated voltage, rated current or rated power. When designing a peripheral circuit, please be fully aware of the following points. (Refer to $P C B$ Layout Considerations below.)

- Ensure the $\mathrm{V}_{\mathbb{N}}$ and GND lines are firmly connected. A large switching current flows through the GND lines and the $\mathrm{V}_{\text {IN }}$ line. If their impedance is too high, noise pickup or unstable operation may result. When the built-in switch is turned off, the inductor may generate a spike-shaped high voltage. Use the highbreakdown voltage capacitor (Cout) which output voltage is 1.5 times or more than the set output voltage.
- After a boosting of the step-up converter, the converter uses Vout as a main power source. Therefore, the ceramic capacitor between the Vout pin and the GND pin acts as a bypass capacitor. Considering the bias dependence, place a $10 \mu \mathrm{~F}$ or more ceramic capacitor (Cоut) between the Vout pin - the GND pin as close as possible. Also, place an approximately $10 \mu \mathrm{~F}$ ceramic capacitor ( $\mathrm{C}_{\mathrm{IN}}$ ) between the $\mathrm{V}_{\mathrm{IN}}$ pin - the GND pin.
- Use a $2.2 \mu \mathrm{H}$ inductor (L) which is having a low equivalent series resistance, having enough tolerable current and which is less likely to cause magnetic saturation.
- The MODE pin is controlled with a logic voltage. To make it " H ", 1.0 V or more must be forced to the MODE pin. If power supply is less than 1.0 V , MODE pin must be pulled up to Vout.
- When using Forced PWM Control Type, the MODE pin should be "H".
- The RP402x can reset the latch protection circuit by setting the CE signal 'L' ( $\mathrm{V}_{\mathrm{CE}}<0.3 \mathrm{~V}$ ) once while the power is switched on $\left(\mathrm{V}_{\mathrm{IN}}>0.8 \mathrm{~V}\right)$. If setting the $C E$ pin when $\mathrm{V}_{\text {IN }}$ does not reach 0.8 V due to too large $\mathrm{C}_{\mathrm{IN}}$, the latch protection circuit cannot be reset correctly. Likewise, if starting the device up when the CE pin is shorted to the $\mathrm{V}_{\text {IN }}$ pin or $\mathrm{V}_{\text {out }}$ pin, the latch protection circuit cannot be reset.
- If controlling the CE pin by input voltage, the gradient of the power supply at rising must be considered. So, the CE pin must be connected via the delay circuit or the voltage detector to become the CE pin voltage less than 0.3 V until the $\mathrm{V}_{\mathrm{IN}}$ becomes more than 0.8 V .


## PCB Layout Considerations

## Current Path on PCB

Figure 1 and Figure 2 show the current pathways of application circuits when MOSFET is turned ON or when MOSFET is turned OFF, respectively. As shown in Figure 1 and Figure 2, the currents flow in the directions of blue or green arrows. The parasitic components (impedance, inductance or capacitance) formed in the pathways indicated by the red arrows affect the stability of the system and become the cause of noise. Reduce the parasitic components as much as possible. The current pathways should be made by short and thick wirings.


Figure 1. MOSFET-ON


Figure 2. MOSFET-OFF

## RP402x

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RP402Kxxxx (PKG: DFN(PLP)2020-8pin) Typical Board Layout


RP402K00xx (PKG: DFN(PLP)2020-8pin) Typical Board Layout


RP402Nxxxx (PKG: SOT-23-5pin) Typical Board Layout


## TYPICAL CHARACTERISTICS

Note: Typical Characteristics are intended to be used as reference data; they are not guaranteed.

1) Output Voltage vs. Output Current







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## 2) Efficiency vs. Output Current








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3) Standby Current vs. Ambient Temperature


4) Supply Current 1 vs. Ambient Temperature

5) Supply Current 2 vs. Ambient Temperature

6) Start-up vs. Ambient Temperature


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## 7) Hold-on Voltage vs. Ambient Temperature


8) Oscillator Frequency vs. Ambient Temperature


9) Maxduty vs. Ambient Temperature

10) Lx Current Limit vs. Duty

11) Lx Current Limit vs. Ambient Temperature

12) CE " H " Input Voltage vs. Ambient Temperature

13) MODE " H " Input Voltage vs. Ambient Temperature

14) Output Voltage vs. Ambient Temperature


## RP402x

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15) Feedback Voltage vs. Ambient Temperature

16) Start-up Waveform ( Cout $=20 \mu \mathrm{~F}$ )

17) Load Transient Response (Cout $=\mathbf{2 0} \mu \mathrm{F}$ )


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18) Output Voltage Waveform (Cout $=\mathbf{2 0} \mu \mathrm{F}$ )


19) Mode Switching Waveform

20) Bypass Switch ON Resistance





## RP402x

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21) PWM Operable Maximum Input Voltage vs. Ambient Temperature RP402Kxx2x: (Forced PWM)
RP402Kxx1x: MODE = "H" (Normal PWM)



22) Reverse Current at $\mathrm{V}_{\mathrm{IN}}=0$ vs. Ambient Temperature

23) Latch Protection Delay Time vs. Ambient Temperature



The power dissipation of the package is dependent on PCB material, layout, and environmental conditions. The following measurement conditions are based on JEDEC STD. 51-7.

Measurement Conditions

| Item | Measurement Conditions |
| :--- | :--- |
| Environment | Mounting on Board (Wind Velocity $=0 \mathrm{~m} / \mathrm{s}$ ) |
| Board Material | Glass Cloth Epoxy Plastic (Four-Layer Board) |
| Board Dimensions | $76.2 \mathrm{~mm} \times 114.3 \mathrm{~mm} \times 0.8 \mathrm{~mm}$ |
| Copper Ratio | Outer Layer (First Layer): Less than 95\% of 50 mm Square <br> Inner Layers (Second and Third Layers): Approx. 100\% of 50 mm Square <br> Outer Layer (Fourth Layer): Approx. 100\% of 50 mm Square |
| Through-holes | $\phi 0.3 \mathrm{~mm} \times 23$ pcs |

Measurement Result
$\left(\mathrm{Ta}=25^{\circ} \mathrm{C}, \mathrm{Tjmax}=125^{\circ} \mathrm{C}\right)$

| Item | Measurement Result |
| :--- | :---: |
| Power Dissipation | 1800 mW |
| Thermal Resistance ( $\theta \mathrm{ja})$ | $\theta \mathrm{ja}=53^{\circ} \mathrm{C} / \mathrm{W}$ |
| Thermal Characterization Parameter ( $\psi \mathrm{jj})$ | $\psi \mathrm{jj}=27^{\circ} \mathrm{C} / \mathrm{W}$ |

日ja: Junction-to-Ambient Thermal Resistance
$\psi j \mathrm{j}$ : Junction-to-Top Thermal Characterization Parameter


Power Dissipation vs. Ambient Temperature


Measurement Board Pattern


DFN (PLP) 2020-8 Package Dimensions

* The tab on the bottom of the package shown by blue circle is a substrate potential (GND). It is recommended that this tab be connected to the ground plane on the board but it is possible to leave the tab floating.

The power dissipation of the package is dependent on PCB material, layout, and environmental conditions. The following measurement conditions are based on JEDEC STD. 51-7.

## Measurement Conditions

| Item | Measurement Conditions |
| :---: | :---: |
| Environment | Mounting on Board (Wind Velocity $=0 \mathrm{~m} / \mathrm{s}$ ) |
| Board Material | Glass Cloth Epoxy Plastic (Four-Layer Board) |
| Board Dimensions | $76.2 \mathrm{~mm} \times 114.3 \mathrm{~mm} \times 0.8 \mathrm{~mm}$ |
| Copper Ratio | Outer Layer (First Layer): Less than 95\% of 50 mm Square Inner Layers (Second and Third Layers): Approx. 100\% of 50 mm Square Outer Layer (Fourth Layer): Approx. 100\% of 50 mm Square |
| Through-holes | $\phi 0.3 \mathrm{~mm} \times 7 \mathrm{pcs}$ |

Measurement Result
$\left(\mathrm{Ta}=25^{\circ} \mathrm{C}, \mathrm{Tjmax}=125^{\circ} \mathrm{C}\right)$

| Item | Measurement Result |
| :--- | :---: |
| Power Dissipation | 660 mW |
| Thermal Resistance (日ja) | $\theta \mathrm{ja}=150^{\circ} \mathrm{C} / \mathrm{W}$ |
| Thermal Characterization Parameter ( $\psi \mathrm{jt}$ ) | $\psi j \mathrm{j}=51^{\circ} \mathrm{C} / \mathrm{W}$ |

Өja: Junction-to-Ambient Thermal Resistance
$\psi j$ t: Junction-to-Top Thermal Characterization Parameter



SOT-23-5 Package Dimensions

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6. We are making our continuous effort to improve the quality and reliability of our products, but semiconductor products are likely to fail with certain probability. In order to prevent any injury to persons or damages to property resulting from such failure, customers should be careful enough to incorporate safety measures in their design, such as redundancy feature, fire containment feature and fail-safe feature. We do not assume any liability or responsibility for any loss or damage arising from misuse or inappropriate use of the products.
7. Anti-radiation design is not implemented in the products described in this document.
8. The X-ray exposure can influence functions and characteristics of the products. Confirm the product functions and characteristics in the evaluation stage.
9. WLCSP products should be used in light shielded environments. The light exposure can influence functions and characteristics of the products under operation or storage.
10. There can be variation in the marking when different AOI (Automated Optical Inspection) equipment is used. In the case of recognizing the marking characteristic with AOI, please contact Ricoh sales or our distributor before attempting to use AOI.
11. Please contact Ricoh sales representatives should you have any questions or comments concerning the products or the technical information.

Ricoh is committed to reducing the environmental loading materials in electrical devices with a view to contributing to the protection of human health and the environment.
Ricoh has been providing RoHS compliant products since April 1, 2006 and Halogen-free products since April 1, 2012.

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## https://www.e-devices.ricoh.co.jp/en/

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[^0]:    ${ }^{(1)}$ As for the custom-designed IC, please contact our sales representatives

[^1]:    ${ }^{(1)}$ Refer to POWER DISSIPATION for detailed information.

[^2]:    ${ }^{(1)}$ Hold-on Voltage and NMOS/ PMOS ON Resistance are dependent on Vout.
    ${ }^{(2)}$ Quiescent Current 2 is not applicable to RP402xxx2x.
    ${ }^{(3)}$ MODE "H"/ "L" Input Current/ Voltage is only applicable to RP402Kxxxx.
    ${ }^{(4)}$ Lx Limit Current fluctuates depending on Duty.

[^3]:    ${ }^{(1)}$ MODE "H"/ "L" Input Current/ Voltage is only applicable to RP402Kxxxx.
    (2) $\mathrm{V}_{\mathrm{IN}} \geq 1.7 \mathrm{~V}$
    ${ }^{(3)}$ Protection Delay Time is not included in RP402xxxxB/D/F/H.
    ${ }^{(4)}$ Anti-ringing Switch ON Resistance is dependent on Vout. Not applicable to RP402xxx2x.
    ${ }^{(5)}$ Bypass Switch ON Resistance is dependent on VIN.

[^4]:    ${ }^{(1)}$ Hold-on Voltage and NMOS/ PMOS ON Resistance are dependent on Vout
    ${ }^{(2)}$ Quiescent Current 2 is not applicable to RP402K002x.
    ${ }^{(3)}$ Lx Limit Current fluctuates depending on Duty.

[^5]:    ${ }^{(1)}$ Soft-start Time is $\mathrm{V}_{\mathrm{IN}} \geq 1.7 \mathrm{~V}$.
    ${ }^{(2)}$ Quiescent Current 2 is not applicable to RP402K002x.
    ${ }^{(3)}$ Lx Limit Current fluctuates depending on Duty.
    ${ }^{(4)}$ Bypass Switch ON Resistance is dependent on Vin.

