RICOH

0.5A/1A PWM/VFM Step-down DC/DC Converter with Synchronous Rectifier

No. EA-362-171208

OUTLINE

The RP509x is a low supply current CMOS-based PWM/VFM step-down DC/DC converter with synchronous rectifier featuring 0.5 A/1 A output current⁽¹⁾. Internally, a single converter consists of a reference voltage unit, an error amplifier, a switching control circuit, a mode control circuit, a soft-start circuit, an undervoltage lockout (UVLO) circuit, a thermal shutdown circuit, and switching transistors. The RP509x is employing synchronous rectification for improving the efficiency of rectification by replacing diodes with built-in switching transistors. Using synchronous rectification not only increases circuit performance but also allows a design to reduce parts count. Output voltage controlling method is selectable between a PWM/VFM auto-switching control type and a forced PWM control type, which further reduces noise than a normal PWM control under a light load, and these types can be set by the MODE pin. Output voltage type is selectable between an internally fixed output voltage type and an externally adjustable output voltage type. Protection circuits in the RP509x is current limit circuit and thermal shutdown circuit. LX current limit value (Typ.) is selectable between 1.6 A and 1.0 A. The RP509Z is available in WLCSP-6-P6 which achieves high-density mounting on boards. Using capacitor of 0402-/1005-size (inch/mm) and inductor of 0603-/1608-size (inch/mm) as external parts help to save space for devices. The RP509N is available in SOT-23-6.

FEATURES

 Input Voltage Range (Maximum Rating) 2.3 V to 5.5 V (6.5 V) Output Voltage Range (Fixed Output Voltage Type)0.6 V to 3.3 V, settable in 0.1 V steps (Adjustable Output Voltage Type) 0.6 V to 5.5 V Output Voltage Accuracy (Fixed Output Voltage Type) ·······±1.5% (V_{SET}⁽²⁾ ≥ 1.2 V), ±18 mV (V_{SET} < 1.2 V) Feedback Voltage Accuracy (Adjustable Output Voltage Type) ---- ±9 mV (V_{FB} = 0.6 V) Output Voltage/Feedback Voltage Temperature Coefficient ±100 ppm/°C Selectable Oscillator FrequencyTyp. 6.0 MHz Oscillator Maximum DutyMin. 100% • Built-in Driver ON Resistance (V_{IN} = 3.6 V)······ Typ. Pch. 0.175 Ω, Nch. 0.155 Ω (RP509Z) Typ. Pch. 0.195 Ω, Nch. 0.175 Ω (RP509N) UVLO Detector ThresholdTyp. 2.0 V Soft-start Time Typ. 0.15 ms Package WLCSP-6-P6 (1.28 mm x 0.88 mm x 0.69 mm) SOT-23-6

⁽¹⁾ This is an approximate value. The output current is dependent on conditions and external components. ⁽²⁾ V_{SET} = Set Output Voltage

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APPLICATIONS

- Portable Communication Equipment: Mobiles/Smartphones, Digital Cameras and Note-PCs
- Li-ion Battery-used Equipment

SELECTION GUIDE

The set output voltage, the output voltage type, the auto-discharge function⁽¹⁾, and the LX current limit for the ICs are user-selectable options.

Selection Guide

Product Name	Package	Quantity per Reel	Pb Free	Halogen Free
RP509ZxxX\$-E2-F	WLCSP-6-P6	5,000 pcs	Yes	Yes
RP509NxxX\$-TR-FE	SOT-23-6	3,000 pcs	Yes	Yes

xx: Specify the set output voltage (V_{\text{SET}})

Fixed Output Voltage Type: 06 to 33 (0.6 V to 3.3 V, 0.1 V steps)

The voltage in 0.05 V step is shown as follows.

1.05 V: RP509Z101B5

1.15 V: RP509N111x5

Adjustable Output Voltage Type: 00 only

X: Specify the LX Current Limit (I_{LXLIM})

Тур. 1.6 А: 1

Typ. 1.0 A: 2

\$: Specify the version

Version	Output Voltage Type	Auto-discharge	Oscillator Frequency	V _{SET}	
A	Fixed	No		0.6 V to 2.2 V	
В	Fixed	Yes		0.0 V 10 3.3 V	
С	Adjustable	No			
D	Aujustable	Yes		0.6 V to 5.5 V	

⁽¹⁾ Auto-discharge function quickly lowers the output voltage to 0 V, when the chip enable signal is switched from the active mode to the standby mode, by releasing the electrical charge accumulated in the external capacitor.



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BLOCK DIAGRAM

RP509ZxxXA/RP509ZxxXB, RP509NxxXA/RP509NxxXB (Fixed Output Voltage Type)



RP509xxxXA Block Diagram



RP509xxxXB Block Diagram

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RP509Z00XC/RP509Z00XD, RP509N00XC/RP509N00XD (Adjustable Output Voltage Type)

RP509x00XC Block Diagram



RP509x00XD Block Diagram

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PIN DESCRIPTION





WLCSP-6 Pin Configurations





SOT-23-6 Pin Configurations

WLCSP-6 Pin Description

Pin No.	Symbol	Description
A1	MODE	Mode Control Pin (High: Forced PWM Control, Low: PWM/VFM Auto-switching Control)
B1	LX	Switching Pin
C1	VOUT/VFB	Output/Feedback Voltage Pin
A2	VIN	Input Voltage Pin
B2	CE	Chip Enable Pin, Active-high
C2	GND	Ground Pin

SOT-23-6 Pin Description

Pin No.	Symbol	Description
1	CE	Chip Enable Pin, Active-high
2	GND	Ground Pin
3	VIN	Input Voltage Pin
4	MODE	Mode Control Pin (High: Forced PWM Control, Low: PWM/VFM Auto-switching Control)
5	LX	Switching Pin
6	VOUT/VFB	Output/Feedback Voltage Pin

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ABSOLUTE MAXIMUM RATINGS

Absolute Maximum Ratings

Absolute Ma	ximum Ratings		(GND = 0 V)		
Symbol		ltem		Rating	Unit
Vin	Input Voltage			-0.3 to 6.5	V
V_{LX}	LX Pin Voltage			–0.3 to V _{IN} +0.3	V
Vce	CE Pin Voltage		-0.3 to 6.5	V	
VMODE	MODE Pin Voltage			-0.3 to 6.5	V
Vout/Vfb	VOUT/VFB Pin Voltage			-0.3 to 6.5	V
I _{LX}	LX Pin Output Current			1.6	A
D-	Power	WLCSP6-P6	JEDEC STD. 51-9 Test Land Pattern	910	mW
PD	Dissipation ⁽¹⁾	SOT-23-6	Ultra-High Wattage Land Pattern	892	mW
Tj	Junction Tempe	rature		-40 to 125	°C
Tstg	Storage Temperature Range			-55 to 125	°C

ABSOLUTE MAXIMUM RATINGS

Electronic and mechanical stress momentarily exceeded absolute maximum ratings may cause the permanent damages and may degrade the lifetime and safety for both device and system using the device in the field. The functional operation at or over these absolute maximum ratings is not assured.

RECOMMENDED OPERATING CONDITIONS

Symbol	Item	Rating	Unit
VIN	Input Voltage	2.3 to 5.5	V
Та	Operating Temperature Range	-40 to 85	°C

RECOMMENDED OPERATING CONDITIONS

All of electronic equipment should be designed that the mounted semiconductor devices operate within the recommended operating conditions. The semiconductor devices cannot operate normally over the recommended operating conditions, even if when they are used over such conditions by momentary electronic noise or surge. And the semiconductor devices may receive serious damage when they continue to operate over the recommended operating conditions.

⁽¹⁾ Refer to POWER DISSIPATION for detailed information.

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ELECTRICAL CHARACTERISTICS

Test circuit is operated with "Open Loop Control" (GND = 0 V), unless otherwise specified.

RP509Zxx1A/RP509Zxx1B, RP509Nxx1A/RP509Nxx1B Electrical Characterisitcs (Ta = 25°C)								
Symbol	Item		Conditio	ons	Min.	Тур.	Max.	Unit
Vout	Output Voltage	$V_{IN} = V_{CE} = 3$ $(V_{SET} \le 2.6)$ $V_{IN} = V_{CE} = V_{T}$ $(V_{SET} > 2.6)$	3.6 V V), set +1 V V)	V _{SET} ≥ 1.2 V V _{SET} < 1.2 V	x 0.985 -0.018		x 1.015 +0.018	V
ΔV _{OUT} / ΔTa	Output Voltage Temperature Coefficient	-40 °C ≤ Ta	a ≤ 85 °C			±100		ppm/ °C
fosc	Oscillator Frequency	$V_{IN} = V_{CE} = 3$ Loop Contro	8.6 V, V _{SET} ol"	= 1.8 V, "Closed	4.8	6.0	7.2	MHz
IDD	Supply Current	$V_{IN} = V_{CE} = V_{CE}$	Vоит = 3.6	V, $V_{MODE} = 0 V$		15		μA
ISTANDBY	Standby Current	$V_{IN} = 5.5 V_{,V}$	$V_{CE} = 0 V$			0	5	μA
Ісен	CE "High" Input Current	$V_{IN} = V_{CE} = $	5.5 V		-1	0	1	μA
ICEL	CE "Low" Input Current	$V_{IN} = 5.5 V_{,V}$	$V_{CE} = 0 V$		-1	0	1	μA
IMODEH	MODE "High" Input Current	$V_{IN} = V_{MODE}$	= 5.5 V, V	_{CE} = 0 V	-1	0	1	μA
MODEL	MODE "Low" Input Current	$V_{IN} = 5.5 V$,	$V_{CE} = V_{MO}$	_{DE} = 0 V	-1	0	1	μA
Ivouth	Vout "High" Input Current	VIN = VOUT =	VIN = VOUT = 5.5 V, VCE = 0 V		-1	0	1	μA
IVOUTL	Vout "Low" Input Current	V _{IN} = 5.5 V,	VCE = VOU	т = 0 V	-1	0	1	μA
Rdistr	On-resistance for Auto Discharger ⁽¹⁾	$V_{IN} = 3.6 V, V_{CE} = 0 V$			40		Ω	
ILXLEAKH	LX "High" Leakage Current	$V_{IN} = V_{LX} = 5.5 \text{ V}, V_{CE} = 0 \text{ V}$		-1	0	5	μA	
ILXLEAKL	LX "Low" Leakage Current	$V_{IN} = 5.5 \text{ V}, V_{CE} = V_{LX} = 0 \text{ V}$		-5	0	1	μA	
VCEH	CE "High" Input Voltage	$V_{IN} = 5.5 V$			1.0			V
VCEL	CE "Low" Input Voltage	V _{IN} = 2.3 V					0.4	V
VMODEH	MODE "High" Input Voltage	$V_{IN} = V_{CE} = 1$	5.5 V		1.0			V
VMODEL	MODE "Low" Input Voltage	$V_{\text{IN}} = V_{\text{CE}} =$	2.3 V				0.4	V
RONR	On-resistance of Pch_transistor	RP509Z	$V_{IN} = 3.6$	V,		0.175		Ω
NONF		RP509N	$I_{LX} = -100$	0 mA		0.195		Ω
RONN	On-resistance of Nch. transistor	RP509Z	$V_{IN} = 3.6$	V,		0.155		Ω
		RP509N	$I_{LX} = -100$	0 mA		0.175		Ω
Maxduty	Maximum Duty Cycle			100			%	
t START	Soft-start Time	$V_{IN} = V_{CE} = 3.6 V (V_{SET} \le 2.6 V),$ $V_{IN} = V_{CE} = V_{SET} + 1 V (V_{SET} > 2.6 V)$			150	300	μS	
ILXLIM	LX Current Limit	$V_{IN} = V_{CE} = 3.6 \text{ V} (V_{SET} \le 2.6 \text{ V}),$ $V_{IN} = V_{CE} = V_{SET} + 1 \text{ V} (V_{SET} > 2.6 \text{ V})$		1200	1600		mA	
V _{UVLO1}		$V_{IN} = V_{CE}, F$	alling		1.85	2.00	2.20	V
VUVLO2		VIN = VCE, R	ising		1.90	2.05	2.25	V
T _{TSD}	Thermal Shutdown Threshold	Tj, Rising				140		°C
T _{TSR}	Temperature	Tj, Falling				100		°C

All test items listed under Electrical Characteristics are done under the pulse load condition (Tj \approx Ta = 25°C).

(1) RP509xxx1B only

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Test circuit is operated with "Open Loop Control" (GND = 0 V), unless otherwise specified.

RP509Z00	1C/RP509Z001D, RP509N001C/RF	509N001D EI	ectrical Characterisitcs			(Ta	= 25°C)
Symbol	Item		Conditions	Min.	Тур.	Max.	Unit
V _{FB}	Feedback Voltage	$V_{IN} = V_{CE} =$	3.6 V	0.591	0.600	0.609	V
ΔV _{FB} / ΔTa	Feedback Voltage Temperature Coefficient	-40 °C ≤ T	a ≤ 85 °C		±100		ppm/ °C
fosc	Oscillator Frequency	$V_{IN} = V_{CE} =$ Loop Contr	3.6 V, V _{SET} = 1.8 V, "Closed rol"	4.8	6.0	7.2	MHz
IDD	Supply Current	$V_{IN} = V_{CE} =$	$V_{OUT} = 3.6V, V_{MODE} = 0V$		15		μA
ISTANDBY	Standby Current	V _{IN} = 5.5 V	,V _{CE} = 0 V		0	5	μA
ICEH	CE "High" Input Current	$V_{IN} = V_{CE} =$	5.5 V	-1	0	1	μA
ICEL	CE "Low" Input Current	V _{IN} = 5.5 V	,V _{CE} = 0 V	-1	0	1	μA
IMODEH	MODE "High" Input Current	VIN = VMODE	= 5.5 V, V _{CE} = 0 V	-1	0	1	μA
IMODEL	MODE "Low" Input Current	V _{IN} = 5.5 V	, Vce = Vmode = 0 V	-1	0	1	μA
Ілоптн	Vout "High" Input Current	VIN = VOUT	= 5.5 V, V _{CE} = 0 V	-1	0	1	μA
IVOUTL	Vout "Low" Input Current	V _{IN} = 5.5 V	, Vce = Vout = 0 V	-1	0	1	μA
Rdistr	On-resistance for Auto Discharge ⁽¹⁾	V _{IN} = 3.6 V	V _{IN} = 3.6 V, V _{CE} = 0 V		40		Ω
ILXLEAKH	LX "High" Leakage Current	$V_{IN} = V_{LX} =$	$V_{IN} = V_{LX} = 5.5 \text{ V}, V_{CE} = 0 \text{ V}$		0	5	μA
ILXLEAKL	LX "Low" Leakage Current	V _{IN} = 5.5 V	V _{IN} = 5.5 V, V _{CE} = V _{LX} = 0 V		0	1	μA
V _{CEH}	CE "High" Input Voltage	V _{IN} = 5.5 V	V _{IN} = 5.5 V				V
V _{CEL}	CE "Low" Input Voltage	V _{IN} = 2.3 V				0.4	V
VMODEH	MODE "High" Input Voltage	$V_{IN} = V_{CE} =$	5.5 V	1.0			V
VMODEL	MODE "Low" Input Voltage	$V_{IN} = V_{CE} =$	2.3 V			0.4	V
	On-resistance of	RP509Z	V _{IN} = 3.6 V,		0.175		Ω
KONP	Pch. Transistor	RP509N	I _{LX} = −100 mA		0.195		Ω
	On-resistance of	RP509Z	V _{IN} = 3.6 V,		0.155		Ω
KONN	Nch. Transistor	RP509N	I _{LX} = −100 mA		0.175		Ω
Maxduty	Maximum Duty Cycle			100			%
t start	Soft-start Time	$V_{IN} = V_{CE} =$ $V_{IN} = V_{CE} =$	$V_{IN} = V_{CE} = 3.6 \text{ V} (V_{SET} \le 2.6 \text{ V}),$ $V_{IN} = V_{CE} = V_{SET} + 1 \text{ V} (V_{SET} > 2.6 \text{ V})$		150	300	μs
Ilxlim	LX Current Limit	VIN = VCE = VIN =VCE = 7	$V_{IN} = V_{CE} = 3.6 \text{ V} (V_{SET} \le 2.6 \text{ V}),$ $V_{IN} = V_{CE} = V_{SET} + 1 \text{ V} (V_{SET} > 2.6 \text{ V})$		1600		mA
V _{UVLO1}		$V_{\text{IN}} = V_{\text{CE},} F$	Falling	1.85	2.00	2.20	V
V _{UVLO2}	UVLO ITITESTICIO VOltage	$V_{\text{IN}} = V_{\text{CE},} F$	Rising	1.90	2.05	2.25	V
T _{TSD}	Thermal Shutdown Threshold	Tj, Rising			140		°C
T _{TSR}	Temperature	Tj, Falling			100		°C

All test items listed under Electrical Characteristics are done under the pulse load condition (Tj \approx Ta = 25°C).

(1) RP509x001D only

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RP509Zxx2	2A/RP509Zxx2B, RP509Nxx2A/RI	P509Nxx2B E	lectrical C	haracterisitcs			(Ta	= 25°C)
Symbol	ltem		Conditi	ons	Min.	Тур.	Max.	Unit
Vout	Output Voltage	$V_{IN} = V_{CE} = 3$ $(V_{SET} \le 2.6)$ $V_{IN} = V_{CE} = 3$	3.6 V V), V _{SET} + 1 V	V _{SET} ≥ 1.2 V V _{SET} < 1.2 V	x 0.985		x 1.015 +0.018	V
ΔV _{OUT} / ΔTa	Output Voltage Temperature Coefficient	(VSET > 2.6 −40 °C ≤ Ta	v) i ≤ 85 °C			±100		ppm/ °C
fosc	Oscillator Frequency	V _{IN} = V _{CE} = 3 "Closed Lo	3.6 V, V _{SET} op Contro	= 1.8 V, I"	4.8	6.0	7.2	MHz
I _{DD}	Supply Current	$V_{IN} = V_{CE} = V_{CE}$	V _{OUT} = 3.6	V, V _{MODE} = 0V		15		μΑ
ISTANDBY	Standby Current	$V_{IN} = 5.5 V_{,}$	V _{CE} = 0 V			0	5	μA
ICEH	CE "High" Input Current	$V_{IN} = V_{CE} = 3$	5.5 V		-1	0	1	μΑ
ICEL	CE "Low" Input Current	V _{IN} = 5.5 V,	$V_{CE} = 0 V$		-1	0	1	μA
IMODEH	MODE "High" Input Current	VIN = VMODE	= 5.5 V, V	ce = 0 V	-1	0	1	μA
MODEL	MODE "Low" Input Current	$V_{IN} = 5.5 V$,	Vce = Vmo	de = 0 V	-1	0	1	μA
Ivouth	Vout "High" Input Current	VIN = VOUT =	= 5.5 V, Vc	e = 0 V	-1	0	1	μA
IVOUTL	VOUT "Low" Input Current	$V_{IN} = 5.5 V,$	$V_{IN} = 5.5 \text{ V}, V_{CE} = V_{OUT} = 0 \text{ V}$		-1	0	1	μA
R _{DISTR}	On-resistance for Auto Discharger ⁽¹⁾	$V_{IN} = 3.6 V, V_{CE} = 0 V$			40		Ω	
I _{LXLEAKH}	LX "High" Leakage Current	$V_{IN} = V_{LX} = 5.5 \text{ V}, V_{CE} = 0 \text{ V}$		-1	0	5	μA	
ILXLEAKL	LX "Low" Leakage Current	V _{IN} = 5.5 V,	$V_{IN} = 5.5 \text{ V}, V_{CE} = V_{LX} = 0 \text{ V}$		-5	0	1	μΑ
VCEH	CE "High" Input Voltage	V _{IN} = 5.5 V	V _{IN} = 5.5 V		1.0			V
VCEL	CE "Low" Input Voltage	V _{IN} = 2.3 V					0.4	V
VMODEH	MODE "High" Input Voltage	$V_{\text{IN}} = V_{\text{CE}} =$	5.5 V		1.0			V
V _{MODEL}	MODE "Low" Input Voltage	$V_{\text{IN}} = V_{\text{CE}} =$	2.3 V				0.4	V
Pour	On-resistance of	RP509Z	$V_{IN} = 3.6$	V,		0.175		Ω
NONP	Pch. transistor	RP509N	$I_{LX} = -100$	0 mA		0.195		Ω
Poul	On-resistance of	RP509Z	$V_{IN} = 3.6$	V,		0.155		Ω
NONN	Nch. transistor	RP509N	$I_{LX} = -100$	0 mA		0.175		Ω
Maxduty	Maximum Duty Cycle				100			%
tstart	Soft-start Time	$V_{IN} = V_{CE} = 3.6 \text{ V} (V_{SET} \le 2.6 \text{ V}),$ $V_{IN} = V_{CE} = V_{SET} + 1 \text{ V} (V_{SET} > 2.6 \text{ V})$			150	300	μS	
I _{LXLIM}	LX Current Limit	$V_{IN} = V_{CE} = 3.6 \text{ V} (V_{SET} \le 2.6 \text{ V}),$ $V_{IN} = V_{CE} = V_{SET} + 1 \text{ V} (V_{SET} > 2.6 \text{ V})$		600	1000		mA	
VUVLO1		$V_{IN} = V_{CE}, F$	alling		1.85	2.00	2.20	V
VUVLO2	OVEO miesnola voltage	VIN = VCE, R	ising		1.90	2.05	2.25	V
T _{TSD}	Thermal Shutdown Threshold	Tj, Rising				140		°C
T _{TSR}	Temperature	Tj, Falling				100		°C

Test circuit is operated with "Open Loop Control" (GND = 0 V), unless otherwise specified.

All test items listed under Electrical Characteristics are done under the pulse load condition (Tj \approx Ta = 25°C).

(1) RP509xxx2B only

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Test circuit is operated with "Open Loop Control" (GND = 0 V), unless otherwise specified.

RP509Z002	2C/RP509Z002D, RP509N002C/RF	P509N002D E	lectrical Characterisitcs			(Ta	= 25°C)
Symbol	ltem		Conditions	Min.	Тур.	Max.	Unit
V _{FB}	Feedback Voltage	$V_{\text{IN}} = V_{\text{CE}} =$	3.6 V	0.591	0.600	0.609	V
ΔV _{FB} / ΔTa	Feedback Voltage Temperature Coefficient	-40 °C ≤ Ta	a ≤ 85 °C		±100		ppm/ °C
fosc	Oscillator Frequency	V _{IN} = V _{CE} = Loop Contro	3.6 V, V _{SET} = 1.8 V, "Closed ol"	4.8	6.0	7.2	MHz
I _{DD}	Supply Current	$V_{IN} = V_{CE} = 1$	$V_{OUT} = 3.6V, V_{MODE} = 0V$		15		μA
ISTANDBY	Standby Current	$V_{IN} = 5.5 V,$	$V_{CE} = 0 V$		0	5	μA
Ісен	CE "High" Input Current	$V_{IN} = V_{CE} =$	5.5 V	-1	0	1	μA
ICEL	CE "Low" Input Current	$V_{IN} = 5.5 V_{,}$	V _{CE} = 0 V	-1	0	1	μA
IMODEH	MODE "High" Input Current	VIN = VMODE	= 5.5 V, V _{CE} = 0 V	-1	0	1	μA
IMODEL	MODE "Low" Input Current	$V_{IN} = 5.5 V$,	V _{CE} = V _{MODE} = 0 V	-1	0	1	μA
Ivouth	Vout "High" Input Current	VIN = VOUT =	= 5.5 V, V _{CE} = 0 V	-1	0	1	μA
IVOUTL	VOUT "Low" Input Current	$V_{IN} = 5.5 V,$	$V_{CE} = V_{OUT} = 0 V$	-1	0	1	μA
Rdistr	On-resistance for Auto Discharge ⁽¹⁾	V _{IN} = 3.6 V,	V _{IN} = 3.6 V, V _{CE} = 0 V		40		Ω
I _{LXLEAKH}	LX "High" Leakage Current	$V_{\text{IN}} = V_{\text{LX}} =$	5.5 V, V _{CE} = 0 V	-1	0	5	μA
ILXLEAKL	LX "Low" Leakage Current	V _{IN} = 5.5 V,	V _{IN} = 5.5 V, V _{CE} = V _{LX} = 0 V		0	1	μA
VCEH	CE "High" Input Voltage	V _{IN} = 5.5 V	V _{IN} = 5.5 V				V
VCEL	CE "Low" Input Voltage	V _{IN} = 2.3 V				0.4	V
VMODEH	MODE "High" Input Voltage	$V_{\text{IN}} = V_{\text{CE}} =$	5.5 V	1.0			V
VMODEL	MODE "Low" Input Voltage	$V_{\text{IN}} = V_{\text{CE}} =$	2.3 V			0.4	V
	On-resistance of	RP509Z	$V_{IN} = 3.6 V_{,}$		0.175		Ω
RONP	Pch. Transistor	RP509N	I _{LX} = −100 mA		0.195		Ω
	On-resistance of	RP509Z	$V_{IN} = 3.6 V_{,}$		0.155		Ω
Ronn	Nch. Transistor	RP509N	I _{LX} = −100 mA		0.175		Ω
Maxduty	Maximum Duty Cycle			100			%
İ START	Soft-start Time	$V_{IN} = V_{CE} = 3.6 \text{ V} (V_{SET} \le 2.6 \text{ V}),$ $V_{IN} = V_{CE} = V_{SET} + 1 \text{ V} (V_{SET} > 2.6 \text{ V})$			150	300	μs
ILXLIM	LX Current Limit	$V_{IN} = V_{CE} = 3.6 V (V_{SET} \le 2.6 V),$ $V_{IN} = V_{CE} = V_{SET} + 1 V (V_{SET} > 2.6 V)$		600	1000		mA
VUVLO1		VIN = VCE, F	alling	1.85	2.00	2.20	V
VUVLO2	UVLO I nresnoid voitage	VIN = VCE, R	lising	1.90	2.05	2.25	V
T _{TSD}	Thermal Shutdown Threshold	Tj, Rising			140		°C
T _{TSR}	Temperature	Tj, Falling			100		°C

All test items listed under Electrical Characteristics are done under the pulse load condition (Tj \approx Ta = 25°C).

(1) RP509x002D only

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RP509ZxxXA/RP509ZxxX	(Ta = 25°C			
Dredu	at Nama			
Produ	ct Name	Min.	Тур.	Max.
RP509x06XA	RP509x06XB	0.582	0.600	0.618
RP509x07XA	RP509x07XB	0.682	0.700	0.718
RP509x08XA	RP509x08XB	0.782	0.800	0.818
RP509x09XA	RP509x09XB	0.882	0.900	0.918
RP509x10XA	RP509x10XB	0.982	1.000	1.018
RP509x11XA	RP509x11XB	1.082	1.100	1.118
RP509x12XA	RP509x12XB	1.182	1.200	1.218
RP509x13XA	RP509x13XB	1.281	1.300	1.319
RP509x14XA	RP509x14XB	1.379	1.400	1.421
RP509x15XA	RP509x15XB	1.478	1.500	1.522
RP509x16XA	RP509x16XB	1.576	1.600	1.624
RP509x17XA	RP509x17XB	1.675	1.700	1.725
RP509x18XA	RP509x18XB	1.773	1.800	1.827
RP509x19XA	RP509x19XB	1.872	1.900	1.928
RP509x20XA	RP509x20XB	1.970	2.000	2.030
RP509x21XA	RP509x21XB	2.069	2.100	2.131
RP509x22XA	RP509x22XB	2.167	2.200	2.233
RP509x23XA	RP509x23XB	2.266	2.300	2.334
RP509x24XA	RP509x24XB	2.364	2.400	2.436
RP509x25XA	RP509x25XB	2.463	2.500	2.537
RP509x26XA	RP509x26XB	2.561	2.600	2.639
RP509x27XA	RP509x27XB	2.660	2.700	2.740
RP509x28XA	RP509x28XB	2.758	2.800	2.842
RP509x29XA	RP509x29XB	2.857	2.900	2.943
RP509x30XA	RP509x30XB	2.955	3.000	3.045
RP509x31XA	RP509x31XB	3.054	3.100	3.146
RP509x32XA	RP509x32XB	3.152	3.200	3.248
RP509x33XA	RP509x33XB	3.251	3.300	3.349
_	RP509Z101B5	1.032	1.050	1.068
RP509N111A5	RP509N111B5	1.132	1.150	1.168

Electrical Characteristics by Different Output Voltage

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OPERATING DESCRIPTIONS

Soft-start Time

Starting-up with CE Pin

The IC starts to operate when the CE pin voltage (V_{CE}) exceeds the threshold voltage. The threshold voltage is preset between CE "H" input voltage (V_{CEH}) and CE "Low" input voltage (V_{CEL}).

After the start-of the start-up of the IC, soft-start circuit starts to operate. Then, after a certain period of time, the reference voltage (V_{REF}) in the IC gradually increases up to the specified value.

Notes: Soft start time $(t_{START})^{(1)}$ is not always equal to the turn-on speed of the step-down DC/DC converter. Please note that the turn-on speed could be affected by the power supply capacity, the output current, the inductance value and the C_{OUT} value.



Timing Chart when Starting-up with CE Pin

Starting-up with Power Supply

After the power-on, when V_{IN} exceeds the UVLO released voltage (V_{UVLO2}), the IC starts to operate. Then, softstart circuit starts to operate and after a certain period of time, V_{REF} gradually increases up to the specified value.

Notes: Please note that the turn-on speed of V_{OUT} could be affected by the power supply capacity, the output current, the inductance value, the C_{OUT} value and the turn-on speed of V_{IN} determined by C_{IN} .



Timing Chart when Starting-up with Power Supply

⁽¹⁾ Soft-start time (t_{START}) indicates the duration until the reference voltage (V_{REF}) reaches the specified voltage after softstart circuit's activation.



<u>RP509x</u>

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Undervoltage Lockout (UVLO) Circuit

If V_{IN} becomes lower than V_{SET} , the step-down DC/DC converter stops the switching operation and ON duty becomes 100%, and then V_{OUT} gradually drops according to V_{IN} .

If the V_{IN} drops more and becomes lower than the UVLO detector threshold (V_{UVLO1}), the UVLO circuit starts to operate, V_{REF} stops, and Pch. and Nch. built-in switch transistors turn "OFF". As a result, V_{OUT} drops according to the C_{OUT} capacitance value and the load.

To restart the operation, V_{IN} needs to be higher than V_{UVLO2} . The timing chart below shows the voltage shifts of V_{REF} , V_{LX} and V_{OUT} when V_{IN} value is varied.

Notes: Falling edge (operating) and rising edge (releasing) waveforms of V_{OUT} could be affected by the initial voltage of C_{OUT} and the output current of V_{OUT} .



Timing Chart with Variations in Input Voltage (VIN)

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Current Limit Circuit

Current limit circuit supervises the inductor peak current (the peak current flowing through Pch. Tr.) in each switching cycle, and if the current exceeds the LX current limit (I_{LXLIM}), it turns off Pch. Tr. I_{LXLIM} of the RP509x is set to Typ.1.6 A or Typ.1.0 A.

Notes: I_{LXLIM} could be easily affected by self-heating or ambient environment. If the V_{IN} drops dramatically or becomes unstable due to short-circuit, protection operation could be affected.



Over-Current Protection Operation

RICOH

<u>RP509x</u>

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Operation of Step-down DC/DC Converter and Output Current

The step-down DC/DC converter charges energy in the inductor when LX Tr. turns "ON", and discharges the energy from the inductor when LX Tr. turns "OFF" and controls with less energy loss, so that a lower output voltage (V_{OUT}) than the input voltage (V_{IN}) can be obtained. The operation of the step-down DC/DC converter is explained in the following figures.





Basic Circuit

Inductor Current (IL) flowing through Inductor (L)

- **Step1.** Pch. Tr. turns "ON" and I_L (i1) flows, L is charged with energy. At this moment, i1 increases from the minimum inductor current (I_{LMIN}), which is 0 A, and reaches the maximum inductor current (I_{LMAX}) in proportion to the on-time period (to_N) of Pch. Tr.
- **Step2.** When Pch. Tr. turns "OFF", L tries to maintain I_L at I_{LMAX} , so L turns Nch Tr. "ON" and I_L (i2) flows into L.
- **Step3.** i2 decreases gradually and reaches I_{LMIN} after the open-time period (t_{OPEN}) of Nch. Tr., and then Nch. Tr. turns "OFF". This is called discontinuous current mode.

As the output current (I_{OUT}) increases, the off-time period (t_{OFF}) of Pch. Tr. runs out before I_{L} reaches I_{LMIN} . The next cycle starts, and Pch. Tr. turns "ON" and Nch. Tr. turns "OFF", which means I_{L} starts increasing from I_{LMIN} . This is called continuous current mode.

In PWM mode, V_{OUT} is maintained by controlling ton. The oscillator frequency (fosc) is maintained constant during PWM mode.

When the step-down DC/DC operation is constant, I_{LMIN} and I_{LMAX} during ton of Pch. Tr. would be same as during toFF of Pch. Tr. The current differential between I_{LMAX} and I_{LMIN} is described as ΔI , as the following equation 1.

The above equation is predicated on the following requirements.

 $T = 1 / f_{OSC} = t_{ON} + t_{OFF}$ duty (%) = t_{ON} / T × 100 = t_{ON} × f_{OSC} × 100 t_{OPEN} \le t_{OFF}

In Equation 1, "V_{OUT} × t_{OPEN} / L" shows the amount of current change in "OFF" state. Also, "(V_{IN} – V_{OUT}) × t_{ON} / L" shows the amount of current change at "ON" state.

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Discontinuous Mode and Continuous Mode

As illustrated in Figure A., when I_{OUT} is relatively small, $t_{OPEN} < t_{OFF}$. In this case, the energy charged into L during t_{ON} will be completely discharged during t_{OFF} , as a result, $I_{LMIN} = 0$. This is called discontinuous mode. When I_{OUT} is gradually increased, eventually $t_{OPEN} = t_{OFF}$ and when I_{OUT} is increased further, eventually $I_{LMIN} > 0$ as illustrated in Figure B. This is called continuous mode.



Figure A. Discontinuous Mode

Figure B. Continuous Mode

In the continuous mode, the solution of Equation 1 is described as tonc.

tonc = T × Vout / Vin------ Equation 2

When $t_{ON} < t_{ONC}$, it is discontinuous mode, and when $t_{ON} = t_{ONC}$, it is continuous mode.

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Forced PWM Mode and VFM Mode

Output voltage controlling method is selectable between a forced PWM control type and a PWM/VFM autoswitching control type, and can be set by the MODE pin. The forced PWM control switches at fixed frequency rate in order to reduce noise in low output current. The PWM/VFM auto-switching control automatically switches from PWM mode to VFM mode in order to achieve high efficiency in low output current.

Forced PWM Mode

By setting the MODE pin to "H", the IC switches the frequency at the fixed rate to reduce noise even when the output load is light. Therefore, when I_{OUT} is $\Delta I_L/2$ or less, I_{LMIN} becomes less than "0". That is, the accumulated electricity in CL is discharged through the IC side while I_L is increasing from I_{LMIN} to "0" during ton, and also while I_L is decreasing from "0" to I_{LMIN} during toFF.

VFM Mode

By setting the MODE pin to "Low", in low output current, the IC automatically switches into VFM mode in order to achieve high efficiency. In VFM mode, ton is determined depending on V_{IN} and V_{OUT}.



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APPLICATION INFORMATION

Application Circuits

MODE = High: Forced PWM Control, MODE = Low: PWM/VFM Auto-switching Control



RP509xxxXA/RP509xxxXB (Fixed Output Voltage Type)

MODE = High: Forced PWM Control, MODE = Low: PWM/VFM Auto-switching Control



RP509x00XC/RP509x00XD (Adjustable Output Voltage Type)

Recommended External Components

Symbol	Descriptions
CIN	4.7 μF and more, Ceramic Capacitor,
	See the table of "Input Voltage vs. Capacitance" in the following page.
C	10 µF, Ceramic Capacitor,
COUT	See the table of "Set Output Voltage (V_{SET}) vs. Capacitance" in the following page.
	0.47 μH to 0.56 μH,
L	See the table of "Inductance Range vs. PWM Frequency" in the following page.

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V _{IN} [V]	Size [mm]	C _{ιℕ} [μF]	Rated Voltage [V]	Model
	4005	4.7	6.3	JMK105BBJ475MV (Taiyo Yuden)
	1005	10	6.3	C1005X5R0J106M050BC (TDK)
Up to 4.5	1608	4.7	6.3	GRM188R60J475ME84 (Murata) GRM188R60J475ME19 (Murata) C1608X5R0J475M080AB (TDK) JMK107BJ475MA (Taivo Yuden)
		10	6.3	GRM188R60J106ME47 (Murata) C1608X5R0J106M080AB (TDK) JMK107ABJ106MA (Taiyo Yuden)
	1005	10	6.3	C1005X5R0J106M050BC (TDK)
Up to 5.5	1608	4.7	6.3	GRM188R60J475ME84 (Murata) GRM188R60J475ME19 (Murata) JMK107BJ475MA (Taiyo Yuden)
-		10	6.3	GRM188R60J106ME47 (Murata) C1608X5R0J106M080AB (TDK) JMK107ABJ106MA (Taiyo Yuden)

Input Voltage vs. Capacitance

Set Output Voltage (V_{SET}) vs. Capacitance

Version	V _{SET} [V]	Size [mm]	С _{оυт} [µF]	Rated Voltage [V]	Model
		1005	10	4	GRM155R60G106ME44 (Murata) C1005X5R0G106M050BB (TDK) AMK105CBJ106MV (Taiyo Yuden)
	0.6 to 1.8		10	6.3	C1005X5R0J106M050BC (TDK)
RP509xxxXA RP509xxxXB or RP509x00XC RP509x00XD		1608	10	6.3	GRM188R60J106ME47 (Murata) C1608X5R0J106M080AB (TDK) JMK107ABJ106MA (Taiyo Yuden)
	1.9 to 3.3	1005	10	4	GRM155R60G106ME44(Murata) C1005X5R0G106M050BB (TDK) AMK105CBJ106MV (Taiyo Yuden)
			10	6.3	C1005X5R0J106M050BC (TDK)
			1608	10	6.3
RP509x00XC RP509x00XD	3.4 to 4.5	1608	10	6.3	GRM188R60J106ME47 (Murata) C1608X5R0J106M080AB (TDK) JMK107ABJ106MA (Taiyo Yuden)

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Version	PWM Frequency [MHz]	Size [mm]	Height(Max) [mm]	L [µH]	Rdc (Typ) [mΩ]	Model
		1000	0.05	0.47	110	MDT1608-CHR47M (TOKO)
		1008	0.95	0.47	90	MDT1608-CRR47M (TOKO)
RP509xxxXA RP509xxxXB or RP509x00XC RP509x00XD	6.0			0.5	60	MIPSZ2012D0R5 (FDK)
				0.56	65	MDT2012-CRR56N (TOKO)
				0.47	70	MLP2012HR47MT (TDK)
		2012	1.0	0.54	65	MLP2012HR54MT (TDK)
				0.47	60	CKP2012NR47M-T (Taiyo Yuden)
				0.47	48	BRL2012TR47M6 (Taiyo Yuden)
				0.47	75	LQM21PNR47MG0 (Murata)

Inductance Range vs. PWM Frequency

Precautions for the Selection of External Parts

- Choose a low ESR ceramic capacitor. The capacitance of C_{IN} between V_{IN} and GND should be more than or equal to 4.7 μF. The capacitance of a ceramic capacitor (C_{OUT}) should be 10 μF. Also, choose the capacitor with consideration for bias characteristics and input/output voltages. See the above tables of "Input Voltage vs. Capacitance" and "Set Output Voltage vs. Capacitance".
- The phase compensation of this device is designed according to the C_{OUT} and L values. The inductance range of an inductor should be between 0.47µH to 0.56 µH in order to gain stability. See the above table of "Inductance Range vs. PWM Frequency".
- Choose an inductor that has small DC resistance, has enough permissible current and is hard to cause magnetic saturation. If the inductance value of the inductor becomes extremely small under the load conditions, the peak current of LX may increase along with the load current. As a result, over current protection circuit may start to operate when the peak current of LX reaches to LX limit current. Therefore, choose an inductor with consideration for the value of I_{LXMAX}. See the following page of "Calculation Conditions of LX Pin Maximum Output Current (I_{LXMAX})".
- As for the adjustable output voltage type (RP509x00XC/RP509x00XD), the set output voltage (V_{SET}) can be arbitrarily set by changing the vales of R1 and R2 using the following equation: V_{SET} = V_{FB} × (R1 + R2) / R2

Refer to the following table for the recommended values for R1, R2 and C1.

V _{SET} [V]	R1 [kΩ]	R2 [kΩ]	C1 [pF]		
0.6	0	220	Open		
$0.6 < V_{SET} \le 0.9$		220	47		
$0.9 < V_{SET} \le 1.8$		220	33		
1.8 < V _{SET} ≤ 2.1		150	10		
2.1 < V _{SET} ≤ 2.4	R1 = (V _{SET} / V _{FB} -1) x R2	100	10		
$2.4 < V_{SET} \le 2.7$		68	10		
$2.7 < V_{SET} \le 3.0$		47	10		
$3.0 < V_{SET} \leq V_{IN}$		47	6.8		

Set Output Voltage (V_{SET}) vs. R1/R2/C1 (Adjustable Output Voltage Type)

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Calculation Conditions of LX Pin Maximum Output Current (ILXMAX)

The following equations explain the relationship to determine I_{LXMAX} at the ideal operation of the ICs in continuous mode.

Ripple Current P-P value is described as I_{RP} , ON resistance of Pch. Tr. is described as R_{ONP} , ON resistance of Nch. Tr. is described as R_{ONN} , and DC resistor of the inductor is described as R_L .

First, when Pch. Tr. is "ON", Equation 1 is satisfied.

$V_{IN} = V_{OUT} + (R_{ONP} + R_L) \times I_{OUT} + L \times I_{RP} / t_{ON}$ Equation 1
Second, when Pch. Tr. is "OFF" (Nch. Tr. is "ON"), Equation 2 is satisfied.
$L \times I_{RP} / t_{OFF} = R_{ONN} \times I_{OUT} + V_{OUT} + R_L \times I_{OUT}$ Equation 2
Put Equation 2 into Equation 1 to solve ON duty of Pch. Tr. $(D_{ON} = t_{ON} / (t_{OFF} + t_{ON}))$:
$D_{ON} = (V_{OUT} + R_{ONN} \times I_{OUT} + R_{L} \times I_{OUT}) / (V_{IN} + R_{ONN} \times I_{OUT} - R_{ONP} \times I_{OUT}) \cdots Equation 3$
Ripple Current is described as follows:
$I_{RP} = (V_{IN} - V_{OUT} - R_{ONP} \times I_{OUT} - R_{L} \times I_{OUT}) \times D_{ON} / f_{OSC} / L_{\cdots} Equation 4$
Peak current that flows through L, and LX Tr. is described as follows:
I _{LXMAX} = I _{OUT} + I _{RP} / 2····· Equation 5

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TECHNICAL NOTES

The performance of a power source circuit using this device is highly dependent on a peripheral circuit. A peripheral component or the device mounted on PCB should not exceed its rated voltage, rated current or rated power. When designing a peripheral circuit, please be fully aware of the following points.

- Set the external components as close as possible to the IC and minimize the wiring between the components and the IC. Especially, place a capacitor (C_{IN}) as close as possible to the VIN pin and GND.
- Ensure the VIN and GND lines are sufficiently robust. If their impedance is too high, noise pickup or unstable operation may result.
- The VIN line, the GND line, the VOUT line, an inductor, and LX should make special considerations for the large switching current flows.
- The wiring between the VOUT pin and an inductor (L) (RP509xxxXA/RP509xxxXB) or between a resistor for setting output voltage (R1) and L (RP509x00XC/RP509x00XD) should be separated from the wiring between L and Load.
- Over current protection circuit may be affected by self-heating or power dissipation environment.
- For any setting type of output voltage, the input/output voltage ratio must meet the following requirement to achieve a stable VFM mode at light load when the MODE pin is "Low" (at PWM/VFM Auto Switching):

 $V_{OUT} / V_{IN} < 0.7$





Available Voltage Area with Stable VFM Mode

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PCB LAYOUT



Adjustable Output Voltage Type (RP509Z00XC/D) **Top View**



GND ΙN CI ND MODE vou



Bottom View

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Adjustable Output Voltage Type (RP509N00XC/D)





RICOH

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TYPICAL CHARACTERISTICS

Note: Typical Characteristics are intended to be used as reference data; they are not guaranteed.

1) Efficiency vs. Output Current (RP509Z)

 $V_{OUT} = 1.0 V$ $V_{MODE} = "L" PWM/VFM Auto Switching$ L = MIPSZ2012D0R5



$$\label{eq:Vout} \begin{split} V_{OUT} &= 1.8 \ V \\ V_{MODE} &= "L" \ PWM/VFM \ Auto \ Switching \\ L &= MIPSZ2012D0R5 \end{split}$$



 $\label{eq:Vout} \begin{array}{l} \mathsf{V}_{\text{OUT}} = 1.8 \ \mathsf{V} \\ \mathsf{V}_{\text{MODE}} = "\mathsf{H}" \ \text{Forced} \ \mathsf{PWM} \ \text{Mode} \\ \mathsf{L} = \mathsf{MIPSZ2012D0R5} \end{array}$



 $V_{OUT} = 1.2 V$ $V_{MODE} = "L" PWM/VFM Auto Switching$ L = MIPSZ2012D0R5



 $V_{OUT} = 3.3 V$ (Fixed Output Voltage Type) $V_{MODE} = "L" PWM/VFM Auto Switching$ L = MIPSZ2012D0R5



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Efficiency vs. Output Current (RP509N)

 $V_{OUT} = 1.0 V$ $V_{MODE} = "L" PWM/VFM Auto Switching$ L = MIPSZ2012D0R5



 $V_{OUT} = 1.8 V$ $V_{MODE} = "L" PWM/VFM Auto Switching$





 $V_{OUT} = 1.8 V$ $V_{MODE} = "H"$ Forced PWM Mode L = MIPSZ2012D0R5



$\label{eq:Vout} \begin{array}{l} \mathsf{V}_{\mathsf{OUT}} = 1.2 \ \mathsf{V} \\ \mathsf{V}_{\mathsf{MODE}} = "L" \ \mathsf{PWM/VFM} \ \mathsf{Auto} \ \mathsf{Switching} \\ \mathsf{L} = \mathsf{MIPSZ2012D0R5} \end{array}$



 $\label{eq:Vout} \begin{array}{l} V_{OUT} = 3.3 \ V \ (Fixed \ Output \ Voltage \ Type) \\ V_{MODE} = "L" \ PWM/VFM \ Auto \ Switching \\ L = MIPSZ2012D0R5 \end{array}$



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Small Mount Solution (RP509Z)

 $\label{eq:Vout} \begin{array}{l} V_{OUT} = 1.0 \ V \\ V_{MODE} = "L" \ PWM/VFM \ Auto \ Switching \\ L = MDT1608\text{-}CRR47M \end{array}$



 $\label{eq:Vout} \begin{array}{l} V_{OUT} = 1.8 \ V \\ V_{MODE} = "L" \ PWM/VFM \ Auto \ Switching \\ L = MDT1608\text{-}CRR47M \end{array}$



 $\label{eq:Vout} \begin{array}{l} \mathsf{V}_{\text{OUT}} = 1.8 \ \mathsf{V} \\ \mathsf{V}_{\text{MODE}} = "\mathsf{H}" \ \text{Forced} \ \mathsf{PWM} \ \text{Mode} \\ \mathsf{L} = \mathsf{MDT1608}\text{-}\mathsf{CRR47M} \end{array}$



 $\label{eq:Vout} \begin{array}{l} V_{\text{OUT}} = 1.2 \ V \\ V_{\text{MODE}} = "L" \ PWM/VFM \ Auto \ Switching \\ L = MDT1608\text{-}CRR47M \end{array}$



 $V_{OUT} = 3.3 V$ (Fixed Output Voltage Type) $V_{MODE} = "L" PWM/VFM Auto Switching$ L = MDT1608-CRR47M



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2) Output Voltage vs. Output Current (RP509Z)

 $V_{IN} = 3.6 \text{ V}, V_{OUT} = 1.8 \text{ V}$ $V_{MODE} = "L" PWM/VFM Auto Switching$



Output Voltage vs. Output Current (RP509N) $V_{IN} = 3.6 V$, $V_{OUT} = 1.8 V$

V_{MODE} = "L" PWM/VFM Auto Switching



3) Oscillator Frequency vs. Input Voltage I_{OUT} = 1.0 mA V_{MODE} = "L" PWM/VFM Auto Switching











 $I_{OUT} = 1.0 \text{ mA}$ V_{MODE} = "H" Forced PWM Mode



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$$\label{eq:lout} \begin{split} I_{\text{OUT}} &= 500 \text{ mA} \\ V_{\text{MODE}} &= "H" \text{ Forced PWM Mode} \end{split}$$



4) Load Transient Response Waveform

 $\label{eq:VIN} \begin{array}{l} V_{\text{IN}} = 3.6 \text{ V}, V_{\text{OUT}} = 1.8 \text{ V} \\ V_{\text{MODE}} = "L" \text{ PWM/VFM Auto Switching} \\ I_{\text{OUT}} = 1.0 \text{ -> } 500 \text{ mA} \end{array}$



 $\label{eq:VIN} \begin{array}{l} V_{\text{IN}} = 3.6 \ \text{V}, \ V_{\text{OUT}} = 1.8 \ \text{V} \\ V_{\text{MODE}} = "\text{H"} \ \text{Forced} \ \text{PWM} \ \text{Mode} \\ I_{\text{OUT}} = 1.0 \ \text{->} 500 \ \text{mA} \end{array}$



 $V_{\text{IN}} = 3.6 \ \text{V}, \ \text{V}_{\text{OUT}} = 1.8 \ \text{V} \\ V_{\text{MODE}} = "L" \ \text{PWM/VFM} \ \text{Auto} \ \text{Switching} \\ l_{\text{OUT}} = 500 \ \text{->} 1.0 \ \text{mA}$



 $\label{eq:VIN} \begin{array}{l} V_{\text{IN}} = 3.6 \ \text{V}, \ V_{\text{OUT}} = 1.8 \ \text{V} \\ V_{\text{MODE}} = "\text{H"} \ \text{Forced} \ \text{PWM} \ \text{Mode} \\ I_{\text{OUT}} = 500 \ \text{->} 1.0 \ \text{mA} \end{array}$



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 $V_{IN} = 3.6 V, V_{OUT} = 1.8 V$



 $\label{eq:VIN} \begin{array}{l} V_{\text{IN}} = 3.6 \ \text{V}, \ V_{\text{OUT}} = 1.8 \ \text{V} \\ V_{\text{MODE}} = "\text{H" Forced PWM Mode} \\ \text{I}_{\text{OUT}} = 300 \ \text{->} 600 \ \text{mA} \end{array}$







 $V_{\text{IN}} = 3.6 \text{ V}, \ V_{\text{OUT}} = 1.8 \text{ V} \\ V_{\text{MODE}} = "L" \ PWM/VFM \ Auto \ Switching \\ l_{\text{OUT}} = 600 \ -> \ 300 \ mA$



 $\label{eq:VIN} \begin{array}{l} V_{\text{IN}} = 3.6 \ \text{V}, \ V_{\text{OUT}} = 1.8 \ \text{V} \\ V_{\text{MODE}} = "\text{H" Forced PWM Mode} \\ \text{I}_{\text{OUT}} = 600 \ \text{->} 300 \ \text{mA} \end{array}$





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V_{MODE} = "L" PWM/VFM Auto Swit Iout = 1.0 mA



 $\label{eq:VIN} \begin{array}{l} V_{\text{IN}} = 3.6 \ \text{V}, \ V_{\text{OUT}} = 1.8 \ \text{V} \\ V_{\text{MODE}} = "L" \ PWM/VFM \ Auto \ Switching \\ I_{\text{OUT}} = 500 \ \text{mA} \end{array}$





 $\label{eq:VIN} \begin{array}{l} \mathsf{V}_{\mathsf{IN}} = 3.6 \; \mathsf{V}, \; \mathsf{V}_{\mathsf{OUT}} = 1.8 \; \mathsf{V} \\ \mathsf{V}_{\mathsf{MODE}} = "\mathsf{H}" \; \mathsf{Forced} \; \mathsf{PWM} \; \mathsf{Mode} \\ \mathsf{I}_{\mathsf{OUT}} = 500 \; \mathsf{mA} \end{array}$



POWER DISSIPATION

SOT-23-6

 $(Ta = 25^{\circ}C, Tjmax = 125^{\circ}C)$

Ver. A

The power dissipation of the package is dependent on PCB material, layout, and environmental conditions. The following measurement conditions are based on JEDEC STD. 51-7.

ltem	Measurement Conditions
Environment	Mounting on Board (Wind Velocity = 0 m/s)
Board Material	Glass Cloth Epoxy Plastic (Four-Layer Board)
Board Dimensions	76.2 mm × 114.3 mm × 0.8 mm
Copper Ratio	Outer Layer (First Layer): Less than 95% of 50 mm Square Inner Layers (Second and Third Layers): Approx. 100% of 50 mm Square Outer Layer (Fourth Layer): Approx. 100% of 50 mm Square
Through-holes	φ 0.3 mm × 7 pcs

Measurement Conditions

Measurement Result

Item	Measurement Result
Power Dissipation	660 mW
Thermal Resistance (θja)	θja = 150°C/W
Thermal Characterization Parameter (ψjt)	ψjt = 51°C/W

θja: Junction-to-Ambient Thermal Resistance

wit: Junction-to-Top Thermal Characterization Parameter





Power Dissipation vs. Ambient Temperature

Measurement Board Pattern

PACKAGE DIMENSIONS

SOT-23-6

Ver. A





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POWER DISSIPATION

WLCSP-6-P6

Ver. C

The power dissipation of the package is dependent on PCB material, layout, and environmental conditions. The following measurement conditions are based on JEDEC STD. 51-9.

Measurement Conditions

Item	Measurement Conditions	
Environment Mounting on Board (Wind Velocity = 0 m/s)		
Board Material	Glass Cloth Epoxy Plastic (Four-Layer Board)	
Board Dimensions	101.5 mm x 114.5 mm x 1.6 mm	
Copper Ratio	Outer Layers (First and Fourth Layers): 60%	
	Inner Layers (Second and Third Layers): 100%	

Measurement Result

(Ta = 25°C, Tjmax = 125°C)

Item	Measurement Result		
Power Dissipation	910 mW		
Thermal Resistance (θja)	θja = 109°C/W		

θja: Junction-to-Ambient Thermal Resistance







Measurement Board Pattern

PACKAGE DIMENSIONS

WLCSP-6-P6

(0.24)

0.40

(0.24)

Ver. B



WLCSP-6-P6 Package Dimensions (Unit: mm)

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VISUAL INSPECTION CRITERIA

WLCSP

VI-160823

No.	Inspection Items	Inspection Criteria	Figure
1	Package chipping	A≥0.2mm is rejected B≥0.2mm is rejected C≥0.2mm is rejected And, Package chipping to Si surface and to bump is rejected.	B ↓ C
2	Si surface chipping	A≥0.2mm is rejected B≥0.2mm is rejected C≥0.2mm is rejected But, even if A≥0.2mm, B≤0.1mm is acceptable.	B + C
3	No bump	No bump is rejected.	
4	Marking miss	To reject incorrect marking, such as another product name marking or another lot No. marking.	
5	No marking	To reject no marking on the package.	
6	Reverse direction of marking	To reject reverse direction of marking character.	
7	Defective marking	To reject unreadable marking. (Microscope: X15/ White LED/ Viewed from vertical direction)	
8	Scratch	To reject unreadable marking character by scratch. (Microscope: X15/ White LED/ Viewed from vertical direction)	
9	Stain and Foreign material	To reject unreadable marking character by stain and foreign material. (Microscope: X15/ White LED/ Viewed from vertical direction)	

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