

LT1070/LT1071

FEATURES

- Wide Input Voltage Range: 3V to 60V
- Low Quiescent Current: 6mA
- Internal 5A Switch (2.5A for LT1071)
- Very Few External Parts Required
- Self Protected Against Overloads
- Operates in Nearly All Switching Topologies
- Shutdown Mode Draws Only 50µA Supply Current
- Flyback Regulated Mode Has Fully Floating Outputs
- Comes in Standard 5-Pin TO-220 Package
- Can be Externally Synchronized (Consult Factory)

APPLICATIONS

- Logic Supply 5V at 10A
- 5V Logic to ±15V Op Amp Supply
- Off-Line Converter Up to 200W
- Battery Upconverter
- Power Inverter (+ to -) or (- to +)
- Fully Floating Multiple Outputs
- For Lower Current Applications, See the LT1072

USER NOTE:

This data sheet is only intended to provide specifications, graphs and a general functional description of the LT1070/LT1071. Application circuits are included to show the capability of the LT1070/LT1071. A complete design manual (AN19) should be obtained to assist in developing new designs. This manual contains a comprehensive discussion of both the LT1070 and the external components used with it, as well as complete formulas for calculating the values of these components. The manual can also be used for the LT1071 by factoring in the lower switch current rating. A second Application Note, AN25, which details off-line applications is available.

5A and 2.5A High Efficiency Switching Regulators

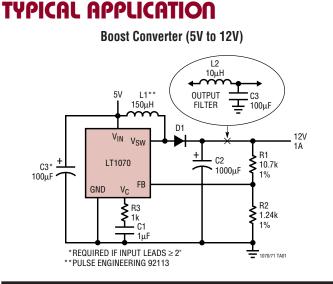
DESCRIPTION

The LT[®]1070/LT1071 are monolithic high power switching regulators. They can be operated in all standard switching configurations including buck, boost, flyback, forward, inverting and "Cuk". A high current, high efficiency switch is included on the die along with all oscillator, control and protection circuitry. Integration of all functions allows the LT1070/LT1071 to be built in a standard 5-pin T0-220 power package. This makes it extremely easy to use and provides "bust proof" operation similar to that obtained with 3-pin linear regulators.

The LT1070/LT1071 operate with supply voltages from 3V to 60V, and draw only 6mA quiescent current. They can deliver load power up to 100W with no external power devices. By utilizing current mode switching techniques, they provide excellent AC and DC load and line regulation.

The LT1070/LT1071 have many unique features not found even on the vastly more difficult to use low power control chips presently available. They use adaptive antisat switch drive to allow very wide ranging load currents with no loss in efficiency. An externally activated shutdown mode reduces total supply current to 50μ A typical for standby operation. Totally isolated and regulated outputs can be generated by using the optional "flyback regulation mode" built into the LT1070/LT1071, without the need for optocouplers or extra transformer windings.

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Maximum Output Power*

*ROUGH GUIDE ONLY. BUCK MODE POUT = 5A * VOUT. SPECIAL TOPOLOGIES DELIVER MORE POWER **DIVIDE VERTICAL POWER SCALE BY 2 FOR LT1071

INPUT VOLTAGE (V)

30

40

50

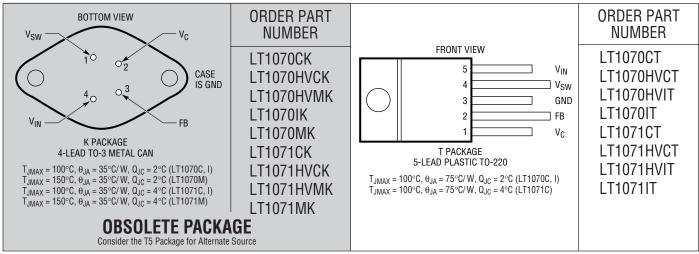
TECHNOLOGY

ABSOLUTE MAXIMUM RATINGS (Note 1)

Supply Voltage	
LT1070/LT1071 (Note 2)	40V
LT1070HV/LT1071HV (Note 2)	60V
Switch Output Voltage	
LT1070/LT1071	65V
LT1070HV/LT1071HV	75V
Feedback Pin Voltage (Transient, 1ms) ±	-15V

Operating Junction Temperature Range
Commercial (Operating) 0°C to 100°C
Commercial (Short Circuit)
Industrial40°C to 125°C
Military (OBSOLETE) – 55°C to 150°C
Storage Temperature Range –65°C to 150°C
Lead Temperature (Soldering, 10 sec) 300°C

PACKAGE/ORDER INFORMATION



Consult LTC Marketing for parts specified with wider operating temperature ranges.

ELECTRICAL CHARACTERISTICS The \bullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at T_A = 25°C. V_{IN} = 15V, V_C = 0.5V, V_{FB} = V_{REF}, output pin open unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
V _{REF}	Reference Voltage	Measured at Feedback Pin, $V_{C} = 0.8V$	•	1.224 1.214	1.244 1.244	1.264 1.274	V V
IB	Feedback Input Current	V _{FB} = V _{REF}	•		350	750 1100	nA nA
g _m	Error Amplifier Transconductance	$\Delta I_{C} = \pm 25 \mu A$	•	3000 2400	4400	6000 7000	µmho µmho
	Error Amplifier Source or Sink Current	V _C = 1.5V	•	150 120	200	350 400	μΑ μΑ
	Error Amplifier Clamp Voltage	Hi Clamp, V _{FB} = 1V Lo Clamp, V _{FB} = 1.5V		1.80 0.25	0.38	2.30 0.52	V V
	Reference Voltage Line Regulation	$3V \le V_{IN} \le V_{MAX}, V_C = 0.8V$	•			0.03	%/V
A _V	Error Amplifier Voltage Gain	$0.9V \le V_C \le 1.4V$		500	800		V/V
	Minimum Input Voltage		•		2.6	3.0	V
IQ	Supply Current	$3V \le V_{IN} \le V_{MAX}, V_C = 0.6V$			6	9	mA
	Control Pin Threshold	Duty Cycle = 0	•	0.8 0.6	0.9	1.08 1.25	V V
	Normal/Flyback Threshold on Feedback Pin			0.4	0.45	0.54	V
		1					10701fe



ELECTRICAL CHARACTERISTICS The \bullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at T_A = 25°C. V_{IN} = 15V, V_C = 0.5V, V_{FB} = V_{REF}, output pin open unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
V _{FB}	Flyback Reference Voltage	I _{FB} = 50μA	•	15 14	16.3	17.6 18.0	V V
	Change in Flyback Reference Voltage	$0.05 \le I_{FB} \le 1 \text{mA}$		4.5	6.8	8.5	V
	Flyback Reference Voltage Line Regulation	$I_{FB} = 50 \mu A, 3V \le V_{IN} \le V_{MAX}$ (Note 3)			0.01	0.03	%/V
	Flyback Amplifier Transconductance (g _m)	$\Delta I_{C} = \pm 10 \mu A$		150	300	650	µmho
	Flyback Amplifier Source and Sink Current	$V_{C} = 0.6V, I_{FB} = 50\mu A (Source)$ $V_{C} = 0.6V, I_{FB} = 50\mu A (Sink)$	•	15 25	32 40	70 70	μΑ μΑ
B _V	Output Switch Breakdown Voltage	$\begin{array}{l} 3V \leq V_{IN} \leq V_{MAX}, I_{SW} = 1.5 mA \\ (LT1070/LT1071) \\ (LT1070HV/LT1071HV) \end{array}$	•	65 75	90 90		V V
V _{SAT}	Output Switch "On" Resistance (Note 4)	LT1070 LT1071	•		0.15 0.30	0.24 0.50	Ω Ω
	Control Voltage to Switch Current Transconductance	LT1070 LT1071			8 4		A/V A/V
I _{LIM}	Switch Current Limit (LT1070)	$\begin{array}{l} \mbox{Duty Cycle} \leq 50\%, \ T_J \geq 25^\circ C \\ \mbox{Duty Cycle} \leq 50\%, \ T_J < 25^\circ C \\ \mbox{Duty Cycle} = 80\% \ (Note 5) \end{array}$	•	5 5 4		10 11 10	A A A
	Switch Current Limit (LT1071)	$\begin{array}{l} \mbox{Duty Cycle} \leq 50\%, \mbox{T}_J \geq 25^\circ\mbox{C}\\ \mbox{Duty Cycle} \leq 50\%, \mbox{T}_J < 25^\circ\mbox{C}\\ \mbox{Duty Cycle} = 80\% \end{tabular} \label{eq:tabular}$	•	2.5 2.5 2.0		5.0 5.5 5.0	A A A
$\frac{\Delta I_{\rm IN}}{\Delta I_{\rm SW}}$	Supply Current Increase During Switch "On" Time				25	35	mA/A
f	Switching Frequency		•	35 33	40	45 47	kHz kHz
DC (Max)	Maximum Switch Duty Cycle			90	92	97	%
	Flyback Sense Delay Time				1.5		μs
	Shutdown Mode Supply Current	$3V \le V_{IN} \le V_{MAX}, V_C = 0.05V$			100	250	μA
	Shutdown Mode Threshold Voltage	$3V \le V_{IN} \le V_{MAX}$	•	100 50	150	250 300	mV mV

Note 1: Absolute Maximum Ratings are those values beyond which the life of a device may be impaired.

Note 2: Minimum switch "on" time for the LT1070/LT1071 in current limit is $\approx 1 \mu s$. This limits the maximum input voltage during short-circuit conditions, in the buck and inverting modes only, to \approx 35V. Normal (unshorted) conditions are not affected. Mask changes are being implemented which will reduce minimum "on" time to $\leq 1 \mu s$, increasing maximum short-circuit input voltage above 40V. If the present LT1070/LT1071 (contact factory for package date code) is being operated in the buck or inverting mode at high input voltages and short-circuit conditions are expected, a resistor must be placed in series with the inductor, as follows:

The value of the resistor is given by:

$$R = \frac{t \bullet f \bullet V_{IN} - V_F}{I_{LIMIT}} - R$$

t = Minimum "on" time of LT1070/LT1071 in current limit, $\approx 1 \mu s$ f = Operating frequency (40kHz)

V_F = Forward voltage of external catch diode at I_{LIMIT}

 I_{LIMIT} = Current limit of LT1070 (\approx 8A), LT1071 (\approx 4A)

R_L = Internal series resistance of inductor

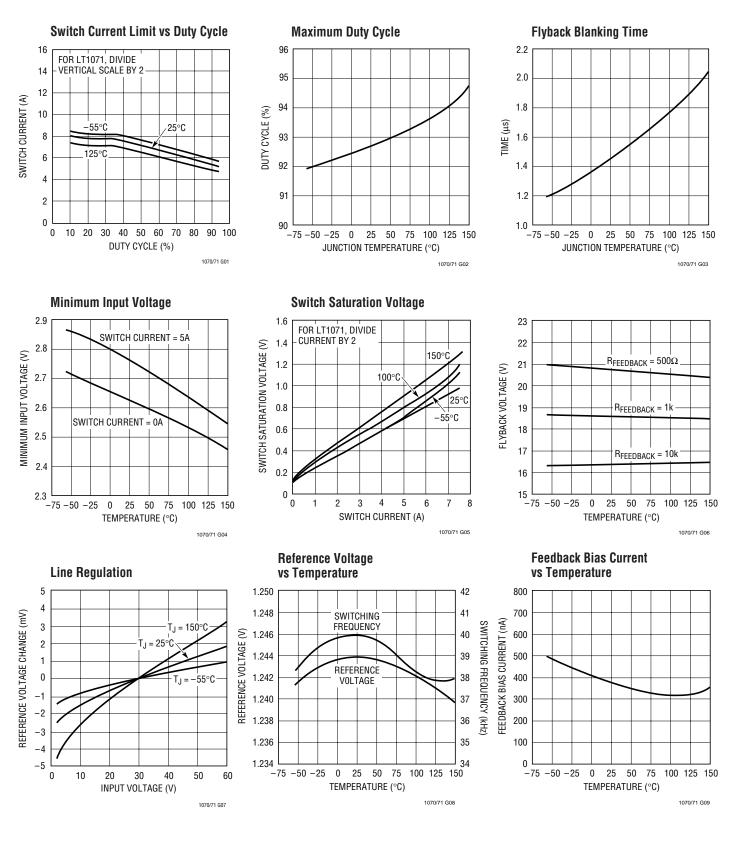
Note 3: V_{MAX} = 55V for LT1070HV and LT1071HV to avoid switch breakdown.

Note 4: Measured with V_{C} in hi clamp, $V_{FB} = 0.8V$. $I_{SW} = 4A$ for LT1070 and 2A for LT1071.

Note 5: For duty cycles (DC) between 50% and 80%, minimum guaranteed switch current is given by $I_{LIM} = 3.33 (2 - DC)$ for the LT1070 and $I_{LIM} = 1.67 (2 - DC)$ for the LT1071.



TYPICAL PERFORMANCE CHARACTERISTICS





TYPICAL PERFORMANCE CHARACTERISTICS

-10

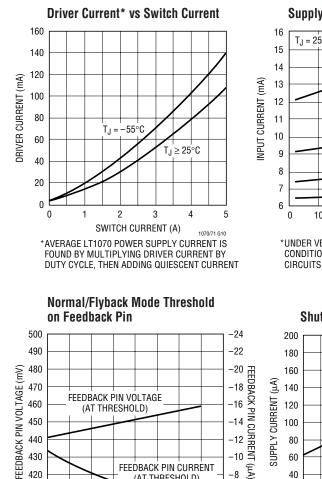
-6

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125 150

1070/71 G13

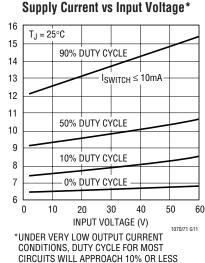
(μA) -8



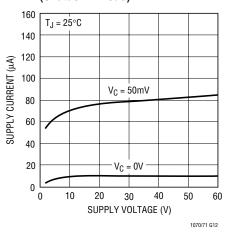
FEEDBACK PIN CURRENT

(AT THRESHOLD)

TEMPERATURE (°C)



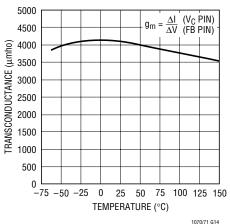
Supply Current vs Supply Voltage (Shutdown Mode)



Shutdown Mode Supply Current

T_J = 150°C 60 $-55^{\circ}C \le T_{J} \le 125^{\circ}C$ 40 20 0 0 10 20 30 40 50 60 70 80 90 100 V_C PIN VOLTAGE (mV) 1070/71 G14

Error Amplifier Transconductance





Shutdown Thresholds

0 25 50 75 100

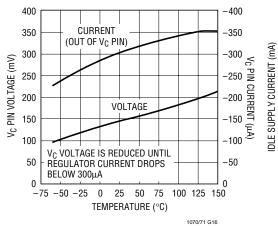
430

420

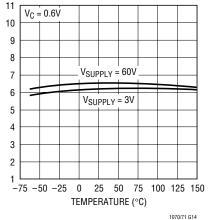
410

400

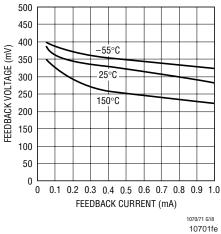
-50 -25



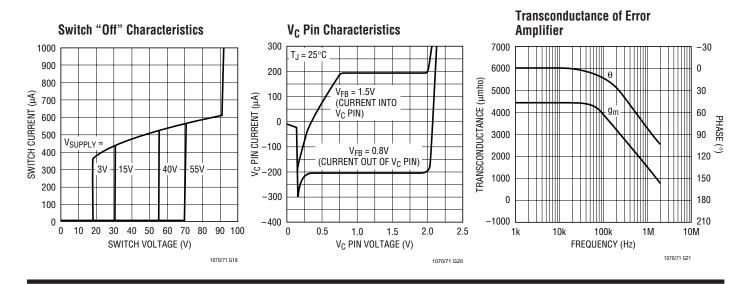




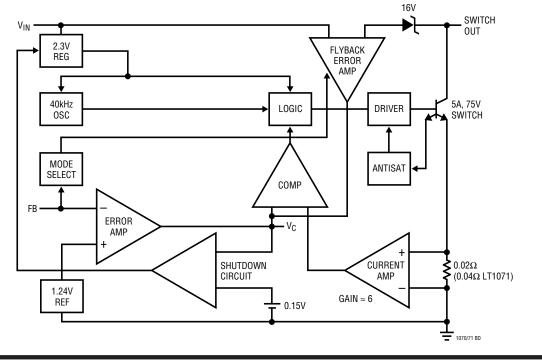
Feedback Pin Clamp Voltage



TYPICAL PERFORMANCE CHARACTERISTICS



BLOCK DIAGRAM



OPERATION

The LT1070/LT1071 is a current mode switcher. This means that switch duty cycle is directly controlled by switch current rather than by output voltage. Referring to the Block Diagram, the switch is turned "on" at the start of each oscillator cycle. It is turned "off" when switch current reaches a predetermined level. Control of output

voltage is obtained by using the output of a voltage sensing error amplifier to set current trip level. This technique has several advantages. First, it has immediate response to input voltage variations, unlike ordinary switchers which have notoriously poor line transient response. Second, it reduces the 90° phase shift at 10701fe



OPERATION

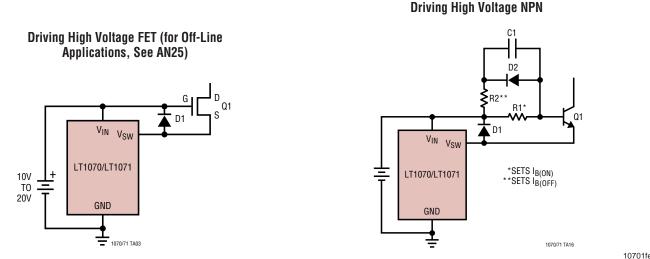
midfrequencies in the energy storage inductor. This greatly simplifies closed-loop frequency compensation under widely varying input voltage or output load conditions. Finally, it allows simple pulse-by-pulse current limiting to provide maximum switch protection under output overload or short-circuit conditions. A low dropout internal regulator provides a 2.3V supply for all internal circuitry of the LT1070/LT1071. This low dropout design allows input voltage to vary from 3V to 60V with virtually no change in device performance. A 40kHz oscillator is the basic clock for all internal timing. It turns "on" the output switch via the logic and driver circuitry. Special adaptive antisat circuitry detects onset of saturation in the power switch and adjusts driver current instantaneously to limit switch saturation. This minimizes driver dissipation and provides very rapid turn-off of the switch.

A 1.2V bandgap reference biases the positive input of the error amplifier. The negative input is brought out for output voltage sensing. This feedback pin has a second function; when pulled low with an external resistor, it programs the LT1070/LT1071 to disconnect the main error amplifier output and connects the output of the flyback amplifier to the comparator input. The LT1070/LT1071 will then regulate the value of the flyback pulse with respect to the supply voltage. This flyback pulse is directly proportional to output voltage in the traditional

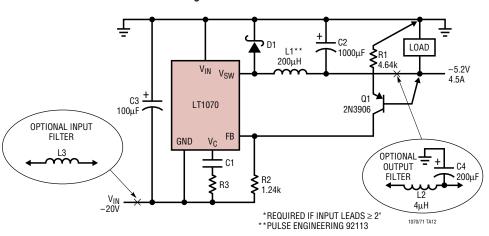
transformer coupled flyback topology regulator. By regulating the amplitude of the flyback pulse, the output voltage can be regulated with no direct connection between input and output. The output is fully floating up to the breakdown voltage of the transformer windings. Multiple floating outputs are easily obtained with additional windings. A special delay network inside the LT1070/ LT1071 ignores the leakage inductance spike at the leading edge of the flyback pulse to improve output regulation.

The error signal developed at the comparator input is brought out externally. This pin $(V_{\rm C})$ has four different functions. It is used for frequency compensation, current limit adjustment, soft starting and total regulator shutdown. During normal regulator operation this pin sits at a voltage between 0.9V (low output current) and 2.0V (high output current). The error amplifiers are current output (g_m) types, so this voltage can be externally clamped for adjusting current limit. Likewise, a capacitor coupled external clamp will provide soft start. Switch duty cycle goes to zero if the V_C pin is pulled to ground through a diode, placing the LT1070/LT1071 in an idle mode. Pulling the V_C pin below 0.15V causes total regulator shutdown, with only 50µA supply current for shutdown circuitry biasing. See AN19 for full application details.

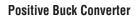
TYPICAL APPLICATIONS (Note that maximum output currents are divided by 2 for the LT1071)

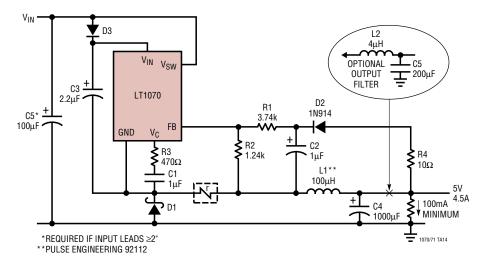




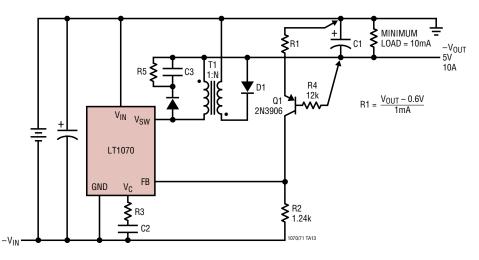


Negative Buck Converter

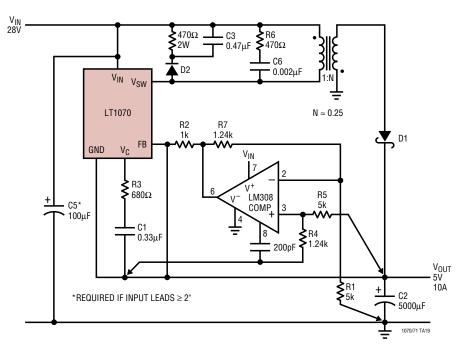








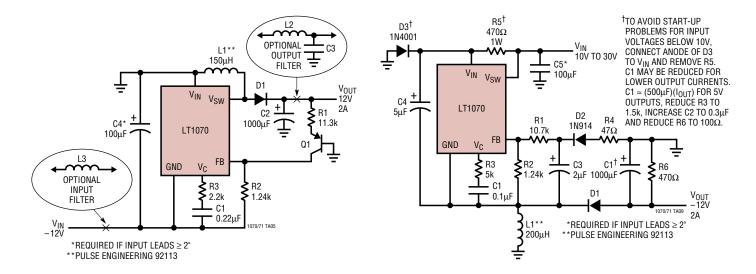




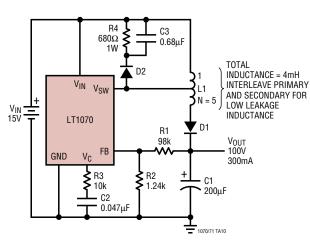
Positive Current Boosted Buck Converter

Negative to Positive Buck/Boost Converter

Positive to Negative Buck/Boost Converter

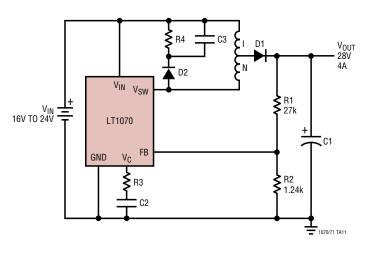




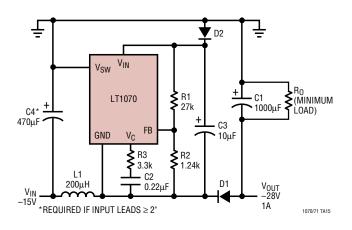


Voltage Boosted Boost Converter

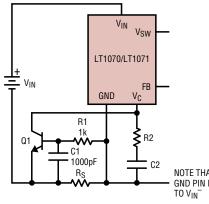
Current Boosted Boost Converter



Negative Boost Regulator



External Current Limit



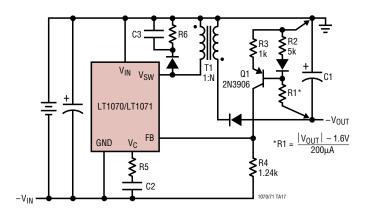
 NOTE THAT THE LT1070/LT1071

 GND PIN IS NO LONGER COMMON

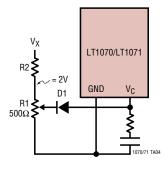
 TO VIN⁻

 1070/1 TA06

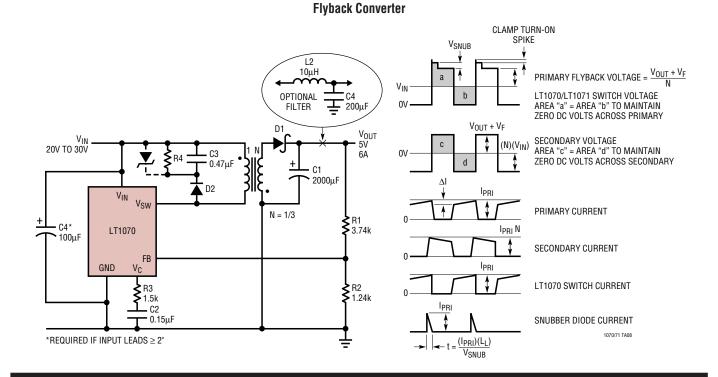
Negative Input/Negative Output Flyback Converter



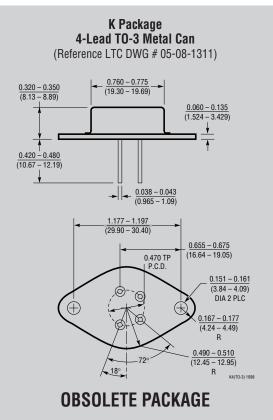
External Current Limit



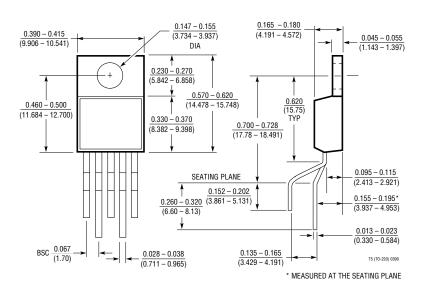




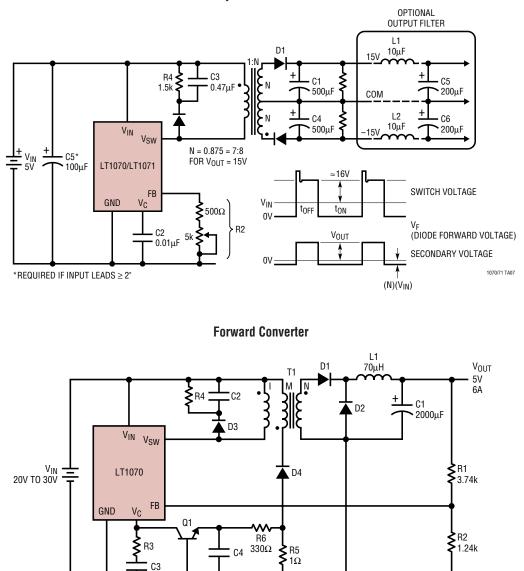
PACKAGE DESCRIPTION



T Package 5-Lead Plastic TO-220 (Standard) (Reference LTC DWG # 05-08-1421)



Information furnished by Linear Technology Corporation is believed to be accurate and reliable. However, no responsibility is assumed for its use. Linear Technology Corporation makes no representation that the interconnection of its circuits as described herein will not infringe on existing patent rights.



Totally Isolated Converter

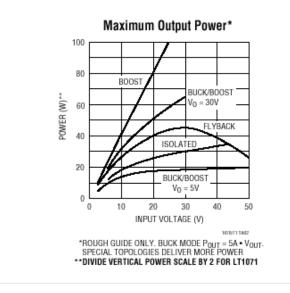
RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT1074/LT1076	High Voltage Switching Regulators	40V Input (60V for HV Versions), 100kHz, 5A and 2A
LT1170/LT1171/ LT1172	100kHz High Efficiency Switching Regulators	40V Input (65V for HV Versions), 5A/2.5A/1.25A Internal Switch
LT1370/LT1371	500kHz High Efficiency Switching Regulators	35V, 6A/3A Internal Switch
LT1374/LT1376	100kHz High Efficiency Switching Regulators	25V Input, 4.5A/1.5A Internal Switch
LT1765	1.25MHz, 3A, Step-Down Regulator	25V Input, TSSOP-16E, SO8 Package
	·	10701fe

LT/CPI 0202 1.5K REV E • PRINTED IN USA

1070/71 TA18





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DESCRIPTION

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PACKAGING

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ORDER INFO

Part numbers ending in PBF are <u>lead free</u>. Please contact LTC marketing for information on lead based finish parts.

Part numbers containing TR or TRM are shipped in tape and reel or 500 unit mini tape and reel, respectively

Please refer to our <u>general ordering information</u> or the product datasheet for more details

Package Variations and Pricing

Part Number	Package	Pins	Temp	Price (1- 99)	Price (1k) [*]	RoHS Data
LT1070CT	<u>TO-220</u>	5	С	\$7.08	\$5.60	View
LT1070CT#PBF	<u>TO-220</u>	5	С	\$7.08	\$5.60	View

LT1070HVCT	<u>TO-220</u>	5	С	\$9.17	\$7.25	View	
LT1070HVCT#06PBF	<u>TO-220</u>	5	С	\$9.17	\$7.25	View	
LT1070HVCT#PBF	<u>TO-220</u>	5	С	\$9.17	\$7.25	View	
LT1070HVIT	<u>TO-220</u>	5	I	\$11.92	\$9.65	View	
LT1070HVIT#PBF	<u>TO-220</u>	5	I	\$11.92	\$9.65	View	
LT1070IT	<u>TO-220</u>	5	I	\$9.25	\$7.60	View	
LT1070IT#PBF	<u>TO-220</u>	5	I	\$9.25	\$7.60	View	
Buy Now							
	-						

Request Samples

^{*} The USA list pricing shown is for BUDGETARY USE ONLY, shown in United States dollars (FOB USA per unit for the stated volume), and is subject to change. International prices may differ due to local duties, taxes, fees and exchange rates. For volume-specific price or delivery quotes, please contact your local Linear Technology <u>sales office or authorized</u> <u>distributor</u>.

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APPLICATIONS

Logic Supply 5V at 10A 5V Logic to ± 15V Op Amp Supply Offline Converter up to 200W Battery Upconverter Power Inverter (+ to -) or (- to +) Fully Floating Multiple Outputs For Lower Current Applications, See the LT1072

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SIMULATE

<u>LTspice IV</u> is a powerful, fast and free simulation tool, schematic capture and waveform viewer with enhancements and models for improving the simulation of switching regulators. To download LTspice IV and other FREE simulation tools, please visit our <u>Design Simulation</u> page.

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